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Parallel and multistep simulation of power system transients

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Département de génie électrique

Thèse présentée en vue de l'obtention du diplôme de *Philosophiæ Doctor*

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présentée par **Ming CAI**

en vue de l'obtention du diplôme de *Philosophiæ Doctor*

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DEDICATION

To all that came into my life and remain therein

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RÉSUMÉ

La simulation des régimes transitoires électromagnétiques (EMT) est devenue indispensable aux ingénieurs dans de nombreuses études des réseaux électriques. L'approche EMT a une nature de large bande et est applicable aux études des transitoires lents (électromécaniques) et rapides (électromagnétiques). Cependant, la complexité des réseaux électriques modernes qui ne cesse de s'accroître, particulièrement des réseaux avec des interconnexions HVDC et des éoliennes, augmente considérablement le temps de résolution dans les études des transitoires électromagnétiques qui exigent la résolution précise des systèmes d'équations différentielles et algébriques avec un pas de calcul pré-déterminé. En tant que sujet de recherche, la réduction du temps de résolution des grands réseaux électriques complexes a donc attiré beaucoup d'attention et d'intérêt. Cette thèse a pour objectif de proposer de nouvelles méthodes numériques qui sont efficaces, flexibles et précises pour la simulation des régimes transitoires électromagnétiques des réseaux électriques.

Dans un premier temps, une approche parallèle et à pas multiples basée sur la norme *Functional Mock-up Interface* (FMI) pour la simulation transitoire des réseaux électriques avec systèmes de contrôle complexes est développée. La forme de co-simulation de la norme FMI dont l'objectif est de faciliter l'échange de données entre des modèles développés avec différents logiciels est implémentée dans EMTP. Tout en profitant de cette implémentation, les différents systèmes de contrôle complexes peuvent être découplés du réseau principal en mémoire et résolus de façon indépendante sur des processeurs séparés. Ils communiquent avec le réseau principal à travers une interface de co-simulation pendant une simulation. Cette méthodologie non seulement réduit la charge de calcul total sur un seul processeur, mais elle permet aussi de simuler les systèmes de contrôle découplés de façon parallèle et à pas multiples. Deux modes de co-simulation sont proposés dans la première étape du développement, qui sont les modes asynchrone et synchrone. Dans le mode asynchrone, tous les systèmes de contrôle découplés (esclaves) sont simulés en parallèle avec le réseau principal (maître) en utilisant un seul pas de calcul tandis que le mode synchrone permet une simulation séquentielle en utilisant différents pas de calcul dans le maître et les esclaves. La communication entre le maître et les esclaves est réalisée et coordonnée par des fonctions qui implémentent le primitif de synchronisation de bas niveau sémaphore. Dans la deuxième étape du développement, afin d'améliorer la méthodologie originale, un nouveau mode asynchrone parallèle à pas multiples (*parallel multistep asynchronous mode*) et une procédure de

correction de signaux sont proposés. Le premier combine les avantages numériques (respectivement la simulation en parallèle et à pas multiples) des deux modes développés précédemment et la seconde, basée sur l'extrapolation linéaire, sert à améliorer la précision de simulation. Cette approche est testée sur différents benchmarks des réseaux électriques avec systèmes de contrôle complexes tels que les contrôles éoliens et les relais de protection. Ses avantages de pouvoir entretenir et améliorer la précision, accélérer la simulation temporelle tout en étant évolutive et flexible sont démontrés par les résultats de simulation en comparaison avec ceux obtenus des simulations EMT sur un seul processeur.

Dans un deuxième temps, une approche de parallélisation-dans-le-temps (*parallel-in-time*) basée sur la technique parallèle de regroupement et permutation d'équations (*equation grouping and reordering technique, PEGR*) pour la simulation transitoire des réseaux électriques est développée dans la thèse. La proposition originale de la technique PEGR, formulée dans une forme d'état, a pour objectif de réduire le nombre total de pas de calcul de façon logarithmique en regroupant et permutant les équations du réseau à plusieurs points de calcul. L'approche de parallélisation-dans-le-temps développée dans cette thèse est adaptée à l'analyse nodale modifiée augmentée (*modified-augmented-nodal analysis, MANA*). Elle est implémentée en C++ avec l'API OpenMP Multithreading pour la parallélisation de résolution en utilisant différents nombres de fils. Le solveur linéaire hautement efficace à matrice creuse KLU est utilisé comme le solveur pour les matrices de réseau. Certaines stratégies de programmation et solutions aux problèmes rencontrés dans les simulations transitoires réalistes des réseaux électriques sont incluses dans le développement de l'approche. Cette approche de parallélisation-dans-le-temps permet une formulation flexible en regroupant les équations du réseau à différents nombres de points de calcul, qui est ajustable par rapport au nombre de processeurs de l'ordinateur utilisé. Afin de valider la précision et l'efficacité de l'approche développée, elle est testée sur des benchmarks réalistes des réseaux électriques et d'autres construits avec de nombreuses répliques de ces mêmes benchmarks représentant des réseaux électriques à plus grandes échelles. Les résultats obtenus de cette approche démontrent qu'elle peut accélérer la simulation temporelle des réseaux électriques à différentes échelles en utilisant différents nombres de processeurs parallèles, en comparaison avec la méthode de résolution séquentielle traditionnelle.

Enfin, certains travaux futurs sont proposés dans le cadre du projet de cette thèse.

ABSTRACT

The simulation of electromagnetic transients (EMT) has become indispensable to utility engineers in a multitude of studies in power systems. The EMT approach is of wideband nature and applicable to both slower electromechanical as well as faster electromagnetic transients. However, the ever-growing complexity of modern-day power systems, especially those with HVDC interconnections and wind generations, considerably increases computational time in EMT studies which require the accurate solution of usually large sets of differential and algebraic equations (DAEs) with a pre-determined time-step. Therefore, computing time reduction for solving complex, practical and large-scale power system networks has become a hot research topic. This thesis proposes new fast, flexible and accurate numerical methods for the simulation of power system electromagnetic transients.

As a first step in this thesis, a parallel and multistep approach based on the Functional Mock-up Interface (FMI) standard for power system EMT simulations with complex control systems is developed. The co-simulation form of the FMI standard, a tool independent interface standard aiming to facilitate data exchange between dynamic models developed in different simulation environments, is implemented in EMTP. Taking advantage of the compatibility established between the FMI standard and EMTP, various computationally demanding control systems can be decoupled from the power network in memory, solved independently on separate processors, and communicate with the power network through a co-simulation interface during a simulation. This not only reduces the total computation burden on a single processor, but also allows parallel and multistep simulation for the decoupled control systems. Following a master-slave co-simulation scheme (with the master representing the power network and the slaves denoting the decoupled control systems), two co-simulation modes, which are respectively the asynchronous and synchronous modes, are proposed in the first stage of the development. In the asynchronous mode, all decoupled subsystems are simulated in parallel with a single numerical integration time-step whereas the synchronous mode allows the use of different numerical time-steps in a sequential co-simulation environment. The communication between master and slaves is coordinated by functions employing the low-level synchronization primitive semaphore. The second stage of the development improves the original methodology by proposing the parallel multistep asynchronous mode that combines the advantages (parallel and multistep simulation) of the previously developed two modes together and a linear extrapolation-based signal correction procedure for enhanced

simulation accuracy. All developments in this approach are tested on realistic power system benchmarks with complex control systems such as wind generator controls and protection relays. The advantages of this approach in maintaining and improving accuracy, simulation speedup, scalability as well as flexibility are demonstrated by simulation results in comparison to conventional single-core-based EMT simulations.

Next, a parallel-in-time approach based on the parallel-in-time equation grouping and reordering (PEGR) technique is developed in the thesis for power system EMT simulations. The original proposition of the PEGR technique aims to logarithmically reduce the number of time-domain solution steps by grouping network equations at several solution time-points together and taking advantage of independency between certain solution steps found after a series of recursive row and column reordering. The parallel-in-time approach developed in this thesis adapts the original formulation of the PEGR technique into the modified-augmented-nodal analysis (MANA) from state-space and is implemented using C++ with API OpenMP Multithreading for parallelization using different numbers of threads. The highly efficient sparse linear solver KLU is used as the network matrix solver. Several programming concerns and issues encountered in realistic power system EMT simulations are also considered in the development of this approach. This parallel-in-time approach allows flexible formulation in grouping the network equations at different numbers of solution time-points, catering to the computing capacity (in terms of the number of logical processors) of the PC used in the simulation. Realistic power system benchmarks as well as larger networks constructed with multiple replicas of these benchmarks are used to test the accuracy and efficiency of the proposed approach. Its capability of speeding up a time-domain simulation using different numbers of parallel processors for power systems of different levels of complexity is demonstrated through simulation results in comparison to the conventional stepwise solution scheme.

Finally, several possible future developments based on the completed work proposed in the thesis are presented.

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LIST OF SYMBOLS AND ABBREVIATIONS

DAE	Differential and algebraic equation
DLL	Dynamic-link library
EMT	Electromagnetic transients
EMTP	Electromagnetic transients program
FA	Frequency adaptive
FDNE	Frequency-dependent network equivalent
FMI	Functional Mock-up Interface
FMU	Functional Mock-up Unit
FPGA	Field-programmable gate array
GPU	Graphics processing unit
HVDC	High voltage direct current
MANA	Modified-augmented-nodal analysis
PEGR	Parallel-in-time equation grouping and reordering
TS	Transient stability

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CHAPTER 1 INTRODUCTION

1.1 Motivation

Modern-day power systems are increasingly complex, thanks to more and more challenging cases created by state-of-the-art developments in the research fields such as HVDC systems and wind generation [1]. Such ever-growing complexity of modern power grids requires high-performance computing resources (preferably inexpensive) as well as advanced computer-based simulation methods for design, operation and post-mortem analysis stages.

Numerical solution methods are crucial in the evolution and technological advancements in modern power systems, in which circuit-based numerical methods, such as those used in the high accuracy time-domain electromagnetic transients (EMT) simulations, are of particular interest to researchers and utility engineers. The EMT approach [2]-[6] targets operation problems in power system analyses. It helps utility engineers perform highly accurate studies including fault analysis, power flow analysis, stability analysis as well as EMT analysis, demonstrating great advantages over phasor-domain approaches [7], [8]. It is of wideband nature and applicable to both slower electromechanical transients of lower frequencies as well as high frequency electromagnetic transients, such as switching and lightning. Typical numerical integration time-steps are in the range of μs , which creates major obstacles in terms of computing times in the accurate solution of large sets of differential and algebraic equations (DAE) for large-scale systems. Therefore, computing time reduction for solving complex, practical and large-scale power system networks has become a hot research topic.

1.1.1 Accelerating EMT simulations—numerical solution algorithms

Over the years, many efforts have been dedicated to reducing computing times in EMT-type simulation methods, in which development of models and new numerical solution algorithms have drawn a fair amount of interest among researchers. The modelling aspect focuses mainly on enhancing computational performance and optimizing accuracy for a given numerical integration time-step in the implementation of model equations [9], [10]. Although this could, in some cases, allow the adoption of larger numerical integration time-steps to improve simulation efficiency, properly maintaining the accuracy, generalization and application in large-scale networks need to

be further investigated. Therefore, more efforts have been dedicated to the research on numerical solution algorithms for enhancing the performance of EMT-type solution schemes, which fall principally into the following categories: adoption of sparse matrix solvers, parallelization techniques, hybrid techniques, network equivalents and frequency adaptive methods.

1.1.1.1 Adoption of sparse matrix solvers

Among all the available sparse linear solvers for efficient large sparse matrix solutions, LU factorization is a popular approach and is used as a direct solver in circuit simulation [11]. In particular, a Gilbert-Peierls' left-looking algorithm based sparse high-performance linear solver named KLU stands out in the solution of circuit matrices thanks to the efficient ordering mechanisms and high-performance factorization and solve algorithms it adopts [12], [13]. Moreover, KLU uses a hybrid ordering strategy consisting of an unsymmetrical permutation to ensure a zero-free diagonal, a symmetrical permutation to Block Triangular Form (BTF) and a fill-in reducing ordering scheme to achieve a superior fill-in quality and good performance speedup when compared to other existing sparse linear solvers for circuit simulation [12], [13]. A KLU-based simulator developed in [13] has demonstrated its numerical advantages for solving circuit matrices. Other LU factorization-based software packages can be found in [11].

1.1.1.2 Parallelization techniques

Parallelization is another area that has been researched in depth in computing time reduction. Throughout the years, a great number of techniques and methods have been proposed and developed in which parallelization is achieved by partitioning a network using the intrinsic propagation delay at distributed parameter transmission line and cable models for natural decoupling of networks while adopting a shared-memory computing scheme [14]-[18]. It is worth mentioning that the co-simulation based parallel approach proposed in [16] partially implemented the Functional Mock-up Interface (FMI) standard [19] with the adoption of a master-slave co-simulation scheme. Notwithstanding, the steady-state solution needs to be obtained at the master-slave interface and stored prior to each co-simulation start, and several steps in [16] are also complex to automate and require user intervention. Other implementation of the co-simulation approach in power system studies can be found in [20], [21].

Several more flexible partitioning techniques are also proposed in the literature, such as the state-space nodal method in [22] and techniques based on the Multi-Area Thevenin Equivalents (MATE) [23]-[27]. However, the MATE-based techniques can be related to more fundamental circuit analysis theories, such as diakoptics [28], hybrid analysis [29] as well as the compensation method to solve non-linearities using network decoupling in EMT-type solvers [29].

Compared to more commonly seen network decoupling based parallel-in-space approaches, parallel-in-time approaches have not been widely researched due to the difficulty in identifying the intrinsic independency of solutions at different time-steps. A parallel-in-time technique is proposed in [30] that identifies truly independent tasks which could be performed simultaneously with an ideal number of parallel processors after certain pre-processing procedures consisting of the building of a huge matrix containing the solutions at all time-steps as well as a series of recursive row and column reordering.

Additionally, an iterative method named waveform relaxation in which parallelism can be exploited in space and time through partitioning is reported in [31]-[37]. Even though it has seen applications in large-scale integrated circuits [32],[34] and transient stability analysis [35], its feasibility and reliability in power system EMT simulations need to be further investigated.

Continuously stable developments in the semiconductor industry and advances in architecture design have prompted the use of computational accelerators in order to enable higher performance when exploiting fine-grained parallelism. The most popular accelerators are Field Programmable Gate Arrays (FPGA) and graphical processing units (GPUs). Since nowadays new generations of FPGAs have the ability of integrating a great number of computational modules and built-in parallel machines, in recent years many attempts have been taken in exploring the parallel-in-space characteristic of this promising computing paradigm [38]-[41]. On the other hand, many efforts have been put in developing massive-threading CPU-GPU based and entirely GPU based fine-grained parallel simulation approaches using the parallel computing platform and API CUDA created by NVIDIA, as can be seen in [42]-[47]. These hardware accelerators can provide fast computations for certain classes of problems and are used as an alternative to accelerate the circuit simulation besides distributed and shared memory architectures.

Furthermore, multithreading can also be used to express parallelism in simulation programs on shared-memory systems, in which OpenMP has become an industry standard API. An OpenMP

Multithreading based parallelization technique on multi-core systems is reported in [17], [18] for the simulation of large power grids, with the network partitioned at distributed parameter transmission lines, as was discussed earlier.

Apart from what has been discussed so far, the development of high-performance real-time EMT-type simulation tools sparked a great deal of interest in the 1970s and has since become an important research field in power system EMT simulations. Major achievements in the past few decades are presented in [48]-[54], which include notably the full-digital real-time simulator Hypersim [48], RTDS [50], a cluster-based real-time simulator [51] from the University of British Columbia, and the processor cluster based real-time simulator eMEGAsim [52]. The basic principle in real-time solvers is the separation of network equations onto several processing units through the natural time-delay decoupling introduced by distributed parameter transmission line and cable models. Despite their rising popularity in the industry, off-line simulators are still preferred for their advantages in accuracy and ability of solving networks without dimension and numerical integration time-step limits.

1.1.1.3 Hybrid techniques

Hybrid simulation techniques, the combination of an EMT-type algorithm simulating a section of the network in detail whose dynamics are of crucial importance and a fundamental-frequency-single-phase-based transient stability program (TS) for the rest of the network, have been extensively studied in the literature [55]-[73]. It was originally proposed to achieve the simulation speed of a TS-type solver that uses numerical integration time-steps in the order of milliseconds while maintaining similar accuracy of an EMT-type solver that usually requires a much smaller numerical integration time-step. Therefore, the conventional hybrid simulation approaches are usually comprised of time-domain solution using the EMT-type solver and phase-domain solution using the TS-type solver, associated through the choice of interface location, equivalencing of different sections of the network, data conversion methods between time- and phasor-domain as well as interaction protocol. Although such an approach can gain from the high efficiency of TS-type solvers, the lack of concrete theoretical foundation, insufficient generalization and accuracy control to deal with arbitrary network topologies prevent the hybrid simulation approaches from gaining widespread applications in practice.

1.1.1.4 Network equivalents and frequency adaptive methods

Research on computing time reduction in frequency-domain is comprised chiefly of the use of frequency dependent network equivalents (FDNEs) [74]-[76] and frequency adaptive (FA) approaches [77], [78]. The former can be used to reduce the size of studied network and consequently accelerate computations of localized transients, whereas the latter allows accelerating computations by applying large time-steps when required for letting EMT-type methods to efficiently perform TS-type computations. Nevertheless, the FDNE-based approaches are currently limited to linear networks, and the selection of network equivalent ports calls for user intervention and empirical judgement. Limitations of the FA-based techniques include higher computational complexity compared to EMT-type approaches and possible errors brought by the aliasing effect when simulating very fast transients.

1.1.2 Accelerating EMT simulations—multistep solution techniques

Another aspect of computing time reduction is the employment of multiple time-steps in different subsystems of the network in the same simulation environment. This multistep simulation scheme, similar to the conventional hybrid simulation approaches, is based on the generalized relaxation technique in which subsystems employing smaller numerical integration time-steps are solved independently while considering those with a larger numerical integration time-step remain unchanged. Solutions are consequently interfaced and updated at the interfacing locations at certain fixed time-points. Using such a simulation scheme, a “data-smoothing” technique at line-bus interfaces is proposed in [79], and a frequency-domain multistep approach is proposed in [80]. Although computational speedup has been observed in some cases, the implementation of such multistep techniques on large-scale networks requires user intervention and remains complex to automate [81].

On a side note, currently all EMT-type solvers for large power system simulations are based on the fixed time-step trapezoidal integration methods. Variable time-step numerical integration methods for the solution of differential equations have seen applications in SPICE tools. Notwithstanding that overall improved numerical performance can be expected with these methods, step size management for better accuracy control arises as a complex problem [82], [83]. In particular, two time-step control mechanisms, iteration-count time-step control and truncation-error time-step control, for variable time-step numerical integration methods were discussed in [82]. It is

demonstrated that the former control mechanism exhibits problems in solution errors, whereas the latter has issues in simulation efficiency with certain types of circuits. Overall, both control mechanisms are circuit-dependent and require a certain level of intervention and empirical judgement from the user.

This thesis is focused on the design and implementation of novel parallel and multistep numerical techniques based on the current network partitioning, generalized relaxation and co-simulation paradigms for fast large-scale power system EMT simulations while accurately accounting for various dynamic phenomena. Parallelism in the developed techniques is exploited both in space and in time. The newly developed techniques are either implemented directly on an EMT-type solver [2] or in an EMT solution scheme. Unique advantages of these techniques in terms of efficiency, flexibility, scalability, level of automation and accuracy control are demonstrated through tests results on power system benchmarks of different levels of complexity, in comparison to conventional EMT-type solution methods.

1.2 Contributions

The achievements and contributions of this thesis are summarized here.

— IMPLEMENTATION OF THE FUNCTIONAL MOCK-UP INTERFACE (FMI) STANDARD IN EMTP

The Functional Mock-up Interface (FMI) standard is a tool independent interface standard developed in the context of the European project Modelisar, providing support for standardized data exchange between dynamic models designed under different simulation environments with the help of xml files and C-code [19]. It aims at exchanging dynamic models between tools with ease, obviating the need to convert one dynamic model developed in a different simulation tool to one adapted to the host simulation tool, and providing an interface standard for coupling of simulation tools in a co-simulation environment. The FMI standard comes in two forms: model exchange and co-simulation. In this thesis, the co-simulation form of the FMI standard is fully implemented in an EMT-type solver [2] and all FMI functions are concretized for the purpose of power system EMT simulations, particularly for power grids with control systems while respecting standard simulation procedure defined in the standard [19]. Such an implementation remarkably

enhances the flexibility of power system EMT-type solvers in their ability of accommodating a broad range of dynamic simulation needs and environments and presents great prospects for parallel and distributed computation by exploiting the modular nature of decoupled dynamic models.

— DEVELOPMENT OF A NEW OFF-LINE SIMULATION METHOD BASED ON THE CO-SIMULATION APPROACH AND USING THE FMI STANDARD FOR PARALLEL AND MULTISTEP EMT SIMULATIONS WITH CONTROL SYSTEMS

One contribution in this thesis is the design and implementation of a new off-line parallel and multistep simulation method based on the co-simulation approach using the FMI standard for power system EMT simulations with control systems on conventional multicore computers. Instead of bridging different solvers, this newly developed approach interfaces different instances of the same solver using decoupling in memory between power network and control systems in which the control systems are decoupled from the power networks into subsystems and the calculation of all subsystems are distributed among separate processors.

The development of this new approach is realized in two stages. In the first stage, full compatibility between EMTP and the FMI standard is established, a master-slave co-simulation scheme is adopted with the master representing the power network and the slaves denoting the decoupled control systems, and two co-simulation modes are designed (asynchronous and synchronous). In the asynchronous mode, the decoupled subsystems are simulated in parallel using a single numerical integration time-step, whereas in the synchronous mode the simulation of each subsystem is executed in a sequential multistep environment. Considerable computational speedup in both modes is observed in power system protection studies on large-scale networks with accuracy properly maintained.

In the second stage, the computational capacity of the asynchronous mode is extended into accommodating the use of different numerical integration time-steps in different subsystems decoupled in memory, greatly improving simulation flexibility and efficiency. Furthermore, a signal correction procedure based on linear extrapolation is introduced to achieve higher accuracy in a multistep simulation environment.

The developed off-line parallel and multistep EMT simulation approach, with its complete memory decoupling between power network and control systems, pure EMT nature, EMT accuracy and

targeted decoupling interface, presents great prospects of higher simulation scalability, flexibility and level of automation in the massive use of control diagram blocks in power system simulations. Unique advantages are demonstrated for large protection system studies.

— DEVELOPMENT OF A NEW OFF-LINE PARALLEL-IN-TIME APPROACH BASED
ON THE PARALLEL-IN-TIME EQUATION GROUPING AND REORDERING (PEGR)
TECHNIQUE FOR POWER SYSTEM EMT SIMULATIONS

Another contribution in this thesis is the development of a new parallel-in-time off-line approach based on the parallel-in-time equation grouping and reordering (PEGR) technique for power system EMT simulations. In the development of this new approach, the fundamental theories of the PEGR technique are revisited, and its original formulation in state-space is extended to the later more popular and advantageous modified-augmented-nodal analysis (MANA) formulation method [2]. This approach concretizes the PEGR technique to constructing a large DAE system that incorporates the solutions at multiple time-points, thereby logarithmically reducing the actual number of solution steps in the forward and backward substitution procedures with the help of recursive row and column reordering.

The algorithm of this approach is realized in C++ with API OpenMP Multithreading, catering to the current computing capability of most off-the-shelf PCs, and the highly efficient sparse linear solver KLU is used as the network matrix solver. Measures are taken to tackle concurrency issues encountered in a parallel computing environment. Treatment of topological changes and arbitrary numbers of simulation points is also included in the algorithm. It is worth mentioning that the developed parallel-in-time approach offers flexible formulation of grouping the network equations at various solution time-points, suitable for PCs with any number of logical processors to attain their full potential of parallelism.

This newly developed parallel-in-time approach, due to its special network equation formulation scheme that differs significantly from the conventional EMT step-wise approaches, sheds remarkable insight on fast power system EMT simulation from a different perspective. It serves as a prototype in a new parallel simulation scheme and facilitates possible integration with other techniques (e.g., hybrid techniques) to solve for power system dynamic problems on a much larger scale.

1.3 Thesis outline

This thesis is composed of six chapters and four appendices.

— CHAPTER 1 INTRODUCTION

This chapter explains the background motivating this PhD project, highlights its objectives and contributions and summarizes the contents of each chapter.

— CHAPTER 2 A PARALLEL MULTISTEP APPROACH BASED ON FUNCTIONAL MOCK-UP INTERFACE (FMI)

It presents the details of implementation of the Functional Mock-up Interface (FMI) standard in EMTP. This includes the adoption of a master-slave co-simulation scheme, the design of two co-simulation modes (parallel single-step asynchronous and sequential multistep synchronous), and the master-slave synchronization process in both modes with the use of low-level synchronization primitive semaphore. Two improvements on the original methodology are then introduced: the development of the parallel multistep asynchronous mode and the linear extrapolation-based signal correction procedure.

— CHAPTER 3 TEST CASES AND SIMULATION RESULTS OF THE FMI-BASED APPROACH

This chapter contains simulation results using the approach developed in Chapter 2 on power system benchmarks of different levels of complexity, compared with those obtained from a conventional EMT-type solver. Error analysis for the accuracy validation tests and discussion on the observed computational speedup are also provided.

— CHAPTER 4 A PARALLEL-IN-TIME APPROACH BASED ON PARALLEL-IN-TIME EQUATION GROUPING AND REORDERING (PEGR)

The development of the off-line parallel-in-time approach based on the PEGR technique is explained in detail in this chapter. It starts off with the theoretical background of the modified augmented nodal analysis (MANA) and the PEGR technique, then moves on to the adaptation of the PEGR technique from state-space to MANA formulation, its implementation in C++ with OpenMP Multithreading. Detailed programming consideration to minimize concurrency issues as

well as the treatment of topological changes and arbitrary numbers of simulation points are also elaborated.

— CHAPTER 5 TEST CASES AND SIMULATION RESULTS OF THE PEGR-BASED PARALLEL-IN-TIME APPROACH

This chapter presents simulation results of the parallel-in-time approach developed in Chapter 4 on various test cases using different numbers of threads, compared to those obtained from the conventional EMT-type step-wise solution scheme.

— CHAPTER 6 CONCLUSION

The main conclusions of the thesis and possible future developments based on this work are presented in this chapter.

— APPENDIX A EXAMPLE XML FILE IN AN FUNCTIONAL MOCK-UP UNIT (FMU)

An example xml file in an FMU generated from a wind generator control system is presented in this appendix.

— APPENDIX B DETAILED IMPLEMENTATION OF FUNCTIONS COORDINATING MASTER-SLAVE CO-SIMULATION

This appendix presents detailed implementation of functions used in the master-slave communication and data exchange process.

— APPENDIX C ALGORITHM FOR THE EXTRACTION OF BLOCK ELEMENTS IN THE \mathbf{L} AND \mathbf{U} FACTORS

The algorithm to extract various block elements in the \mathbf{L} and \mathbf{U} factors for forward and backward substitutions in the PEGR-based parallel-in-time approach is presented in this appendix.

— APPENDIX D LIST OF PUBLICATIONS

All journal and conference publications derived from the work of this thesis are presented here.

CHAPTER 2 A PARALLEL MULTISTEP APPROACH BASED ON FUNCTIONAL MOCK-UP INTERFACE (FMI)

In this chapter, the implementation of the FMI standard in EMTP is explained in detail with elaboration on the contents and functionalities of the executable Functional Mock-up Unit (FMU) as well as the concretization of the FMI functions in the context of co-simulation. It proceeds with co-simulation management including the adoption of a master-slave co-simulation scheme, the design of different co-simulation modes, synchronization and data exchange between master and slaves. Two improvements on such a methodology are then proposed for the enhancement of simulation flexibility, efficiency and accuracy.

2.1 Implementation of FMI in EMTP

This section provides a brief introduction on the FMI standard (more details can be found in [19]), with focus on the definition and components of the executable FMU implementing the interface defined by the standard and detailed implementation of FMI functions for co-simulation.

2.1.1 Functional Mock-up Interface (FMI) standard

As was mentioned in Chapter 1, the Functional Mock-up Interface (FMI) standard is a tool independent interface standard developed in the context of the European project Modelisar, providing support for standardized data exchange between dynamic models designed under different simulation environments with the help of a combination of xml files and C-code (either compiled in DLL/shared libraries or in source code). The first version, FMI 1.0, was finalized and published in 2010, and the design of the second version, FMI 2.0, came into fruition in 2014. The standard is freely accessible under the licence CC-BY-SA [84].

A number of potential advantages prompted the development of such a standard:

- Firstly, it obviates the need to convert a dynamic model developed in a different simulation tool to one adapted to the host simulation tool as details of such a model, usually enclosed in a “blackbox”, are often unknown to the user.
- Secondly, a co-simulation can be easily launched without having to generate DLL code or using a third-party software.

- Thirdly, the exploitation of the modular nature of dynamic models and co-simulation slaves shows great prospects for parallel and distributed computation.

The FMI standards come in two forms:

- Model exchange

In this form, compatible programs and modelling environments generate the code (namely, Functional Mock-up Unit or FMU) of a dynamic system model that can be utilized by other modelling and simulation environments. The models can be described by differential, algebraic and discrete equations with time-, state- and step-events. The FMU only contains model equations but not solver information. Other modelling and simulation environment can thus import the FMU and simulate it with their own solver engine.

- Co-simulation

The co-simulation form provides an interface standard for coupling of simulation tools in a co-simulation environment. The data exchange between subsystems is restricted to discrete communication points. In the time between two communication points, the subsystems are solved independently from each other by their individual solver. Therefore, apart from the model equations of a dynamic model, the FMU generated by compatible programs and modelling environments also contains the solver itself or a link towards it.

The differences between the two forms of the FMI standard are illustrated in Figure 2.1 and Figure 2.2.

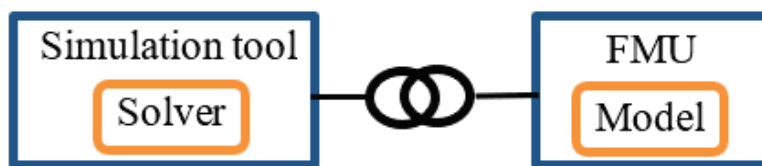


Figure 2.1 FMI for model exchange

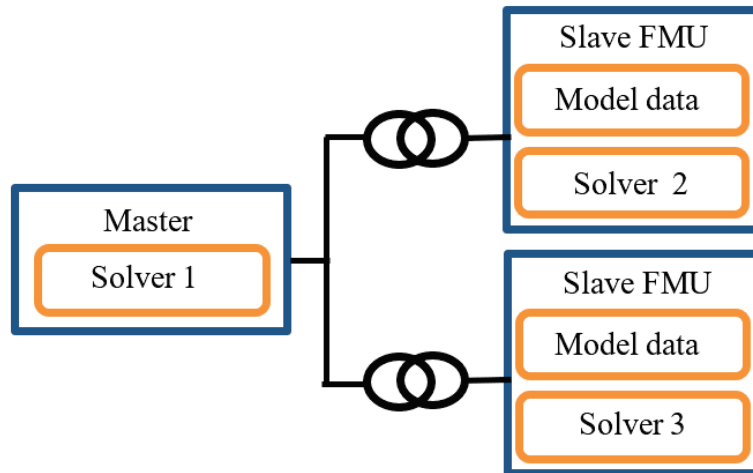


Figure 2.2 FMI for co-simulation

It is noted that in co-simulation, the entire system is decoupled into master and slave subsystems that are solved independently from one another, with the master algorithm controlling the data exchange between subsystems and the synchronization of all simulation slaves.

2.1.2 Functional Mock-up Unit (FMU)

The Functional Mock-up Unit (FMU) is an executable zip file implementing the interface defined by the FMI standard. A model, a co-simulation slave or the coupling part of a tool is enclosed in this file. It contains in general [19]:

- `modelDescription.xml`: an xml file that contains the definition of all exposed variables in the FMU and other static information of the interface (e.g., the number, name, and type of inputs and outputs). The format of the xml file is standardized and documented in the specifications of the FMI standard.
- `resources\`: folder containing all model equations and files specific to the model.
- `binaries\`: folder containing solvers as runnable C-code.

A model description file in the format xml must be present in each FMU. The role of this file is to provide essential information used in co-simulation for the master program. The xml file specifies in particular:

- The version of the FMI standard used

- The name of the model
- The number and type of each input and output
- The number and type of each parameter
- Implementation of the slave

An example of the contents of the model description file xml inside an FMU is presented in Appendix A. The FMU represents the control system of a wind generator using the average value model [1]. In the following sections, the structure of the xml file is defined, and the characteristics of the mandatory nodes and attributes used in co-simulation are specified.

2.1.2.1 Node: fmiModelDescription

The node fmiModelDescription is the root node of the xml file. It contains the general attributes of the FMU, as listed in Table 2.1.

Table 2.1 Attributes of node fmiModelDescription

Attribute	Type	Use	Possible values	Default value
fmiVersion	normalized string	Mandatory		2.0
modelName	string	Mandatory		
guid	normalized string	Mandatory		
description	string	Optional		
author	string	Optional		
version	normalized string	Optional		
copyright	string	Optional		
licence	string	Optional		
generationTool	normalized string	Optional		
generationDateAndTime	dateTime	Optional		
variableNamingConvention	choice	Optional	Flat structured	Flat
numberOfEventIndicators	unsigned int	Optional		

2.1.2.2 Node: fmiModelDescription\CoSimulation

The presence of this node indicates that the FMU will function in the co-simulation mode. The values of the attributes (presented in Table 2.2) of this node depend on the slave characteristics and functionalities.

Table 2.2 Attributes of node fmiModelDescription\CoSimulation

Attribute	Type	Use	Possible values	Default value
sourceFiles	string array	Optional		
modelIdentifier	normalized string	Mandatory		
needsExecutionTool	Boolean	Optional		False
canHandleVariableCommunicationStepSize	Boolean	Optional	True/False	False
canInterpolateInputs	Boolean	Optional	True/False	False
maxOutputDerivativeOrder	unsigned int	Optional		0
canRunAsynchronously	Boolean	Optional	True/False	False
canGetAndSetFMUstate	Boolean	Optional	True/False	False
canSerializeFMUstate	Boolean	Optional	True/False	False
providesDirectionalDerivative	Boolean	Optional	True/False	False

Of all the attributes listed in Table 2.2, it is worthwhile to specify the definition and functionalities of each attribute:

- **sourceFile:** an optional vector containing a list of source files that exist in the folder “Source”, which needs to be compiled to generate the DLL file of the FMU.
- **modelIdentifier:** a prefix used to locate the functions that must be compiled in the source files.
- **needsExecutionTool:** should be set to “true” if an external solver engine is required.
- **canHandleVariableCommunicationStepSize:** indicates whether the FMU can manage variable numerical integration time-steps.
- **canInterpolateInputs:** indicates whether the FMU can interpolate the input variables.
- **maxOutputDerivativeOrder:** indicates the maximum order of derivatives of the output variables that the slave can generate.

- **canRunAsynchronously**: indicate whether the FMU can manage function “fmi2DoStep” in an asynchronous manner.
- **canGetAndSetFMUstate**: indicates whether the slave is able to record and restore the states of FMU, is used mainly for iteration of the slave.
- **canSerializeFMUstate**: indicates whether the slave can serialize the states of FMU.
- **providesDirectionalDerivative**: indicates whether the directional derivatives can be calculated.

2.1.2.3 Node: fmiModelDescription\ModelVariables

The node ModelVariable, comprised of subnodes named “ScalarVariable”, lists all the variables of the FMU. Each subnode “ScalarVariable” corresponds to a variable of the FMU. The attributes of node ModelVariable are shown in Table 2.3.

Table 2.3 Attributes of node fmiModelDescription\ModelVariable

Attribute	Type	Use	Possible values	Default value
name	normalized string	Mandatory		
valueReference	unsigned int	Mandatory		
description	string	Optional		
variability	choice	Optional	Constant/Fixed/Tunable/Discrete/Continuous	Continuous
causality	choice	Optional	Parameter/CalculatedParameter/ Input/Output/Local/Independent	Local
initial	choice	Optional	Exact/Approx/Calculated	
canHandleMultipleSetPerTimeInstant	Boolean	Optional		

Of the attributes listed in Table 2.3:

- “ValueReference” is an identifier. Every variable possesses its own unique identifier.
- “Variability” has 4 possible values:
 - Constant: the value of the variable is fixed and does not change.
 - Fixed: the value of the variable cannot be modified after initialization of FMU.

- Tunable: the value of the variable is constant between two events (model exchange) or between two time-steps (co-simulation).
- Discrete: the value of the variable can only be modified at initialization and during “events”.
- Continuous: no restriction on the value of the variable. Only variables of type “fmiReal” can be continuous.
- “Causality” has 4 possible values:
 - Parameter: independent parameter whose value is chosen by the master during initialization and is fixed during the simulation.
 - CalculatedParameter: independent parameter whose value is fixed during the simulation but can be calculated during initialization.
 - Input: the value can be set by the master.
 - Output: the value can be read the master.
 - Local: local variable calculated from another variable. It is not allowed to use local variables in the master or the other slaves.
 - Independent: independent variable (in general, “time”).
 - None: the value of the variable is of no importance in the simulation.
- “initial” indicates how the variable is initialized. Three values are possible:
 - Exact: the variable is initialized with the value defined in “start”.
 - Approx: the variable is initialized with the value defined in “start”. It is possible to modify it during the initialization stage.
 - Calculated: the variable is calculated from other variables during initialization.

2.1.2.4 Node: fmiModelDescription\Implementation

The node Implementation is comprised of two subnodes, CoSimulation_StandAlone and CoSimulation_Tool. If the subnode CoSimulation_StandAlone is present, all the slave code (model

equations and solver information) is included in the FMU; if instead the subnode `CoSimulation_Tool` is present, a third-party solver will be needed to communicate with the FMU.

The description of all mandatory nodes and their attributes used in co-simulation are listed and explained above. Information regarding nodes mainly used in model exchange and those that are not mandatory can be found in [19].

2.1.3 Implementation of FMI functions for co-simulation

The co-simulation form of the FMI standard 2.0 has been fully implemented in this thesis for control signals (integer, floating-point or Boolean) that can be exchanged by means of the co-simulation interface. It is worth noting that EMTP uses two separate solvers for the solution of the power and control parts of a network (an iterative sparse matrix (Jacobian) solver based on the modified-augmented-nodal analysis for the power part and a Jacobian-based solver for the control part) [2], [6]. For every discrete solution time-point, EMTP first solves the power part then the control part.

As previously mentioned, this new off-line simulation approach also adopts a master-slave co-simulation scheme. The control systems (slaves) are decoupled from the power network and encapsulated into FMUs which contain model and simulator information. The power network (master) then loads these FMUs without prior knowledge of their contents. As the simulation proceeds, the master communicates with the slaves through two layers of DLL files incorporating standardized FMI functions via a co-simulation bus composed of a shared memory zone (buffer) for data exchange, as well as associated synchronization mechanisms created by the master. This concept is illustrated in Figure 2.3 in which only two slave subsystems are shown for simplicity.

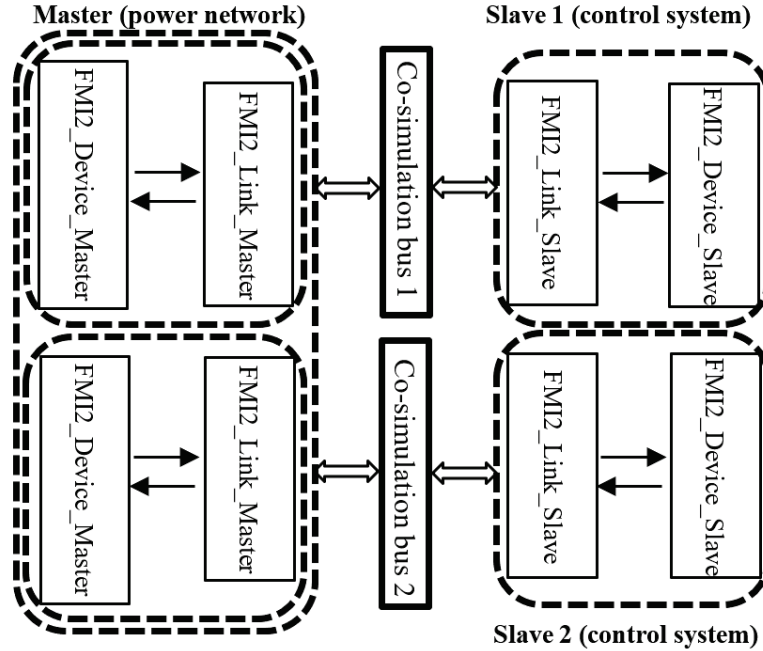


Figure 2.3 Interfacing three subnetworks (Master, Slave 1 and Slave 2) via DLL files and two co-simulation buses

FMI2_Device_Master defines the functions to interact with the EMTP solver, compatible with the FMI standard 2.0 functions defined and implemented in FMI2_Link_Master. FMI2_Device_Slave is in charge of interaction between a slave instance and EMTP, and FMI2_Link_Slave provides auxiliary coordination of master-slave data exchange by communicating with functions defined in FMI2_Link_Master. The contents of the latter two files are not constrained by the FMI standard. It is noted that a co-simulation bus is created for each slave instance in the simulation, and although a sperate set of master DLL files are used for every master-slave interface, they interact with the same solver instance in the master.

In the following sections, the definition, functionality and implementation of principal functions (whether defined in the FMI standard or designed for the interaction with the EMTP solver) in the 4 DLL files, as shown in Figure 2.3, are explained. Due to the high level of complexity in most of the implementation, the purpose and functionality of each function are summarized and made concise, and their implementation is explained in a descriptive manner. Detailed implementation of functions `fmi2DoStep` and `stepFunc` which are used to coordinate co-simulation between master and slaves is given in Appendix B for further references. Information with regards to the FMI function arguments and data types can be found in [19].

2.1.3.1 Implementation of functions in FMI2_Link_Master

- fmi2Instantiate

This function is called by the master program in order to load an FMU at the beginning of the co-simulation stage. It is called once for every slave (FMU) instance and returns an element of type “fmiComponent”. Its principal execution routines are:

- 1) Allocate memory for an FMU object
- 2) Assign certain values to the object
- 3) Create a file with a specified instance name in the same path where the FMU is extracted
- 4) Read values from the xml file and assign those values to the created FMU object
- 5) Read the location of the netlist file of the FMU

- fmi2FreeInstance

This function releases the slave instance, that is to say, it releases all the memory allocated for the FMU.

- fmi2SetDebugLogging

This function allows the master to choose whether the FMU can send log messages to the master as well as the level of details in these messages.

- fmi2SetupExperiment

This function asks the FMU to configure the co-simulation. It includes function calls to setINI and setIN [19].

- fmi2EnterInitializationMode

This function asks the FMU to enter the initialization mode in which variables with attributes “initial” = “exact” or “approx” can be defined. It is noted that the co-simulation bus used for data exchange between master and slave is realized by file mapping. Its main execution routines are:

- 1) Create the co-simulation bus, which is a file mapping object with a certain size
- 2) Create a map view associated with the file mapping object

- 3) Initialize certain elements in the co-simulation bus
- 4) Initialize all three synchronization semaphores by setting them to 0
- 5) Release semaphore SemInitialization to the slave, informing the slave that the co-simulation bus is successfully created and asking the slave to connect to it

- fmi2ExitInitializationMode

This function asks the FMU to exist the initialization mode. It executes the following routines:

- 1) Wait for the slave to release semaphore SemMaster to finish the initialization mode
- 2) Release SemMaster to itself so that the master can start its execution in the case of asynchronous mode

Details on synchronization semaphores and different co-simulation modes will be explained hereinafter.

- fmi2Terminate

This function notifies the FMU that the simulation is terminated by releasing the semaphore SemSlave one last time.

- fmi2Reset

This function demands that the FMU return to its initial state.

- fmi2SetReal, fmi2SetInteger, fmi2SetBoolean, fmi2SetString, fmi2GetReal, fmi2GetInteger, fmi2GetBoolean, fmi2GetString

These functions allow to transfer values from the master to the slave and vice versa. They are called at every simulation time-step of the master and can also be used to initialize variables in the FMU.

- fmi2GetStatus, fmi2GetRealStatus, fmi2GetIntegerStatus, fmi2GetBooleanStatus, fmi2GetStringStatus

These functions are called by the master to require the status of the co-simulation.

- fmi2DoStep

A call to this function starts the calculation of a new time-step in the FMU. It consists principally of the following execution routines:

- 1) Write the current communication time-point and the master numerical integration time-step onto the co-simulation bus
- 2) Read from the co-simulation bus the current slave time-point and its numerical integration time-step
- 3) Verify the co-simulation mode (synchronous or asynchronous)
 - a) If the co-simulation mode is synchronous
 - i. If the slave lags behind the master
 - ii. Release semaphore SemSlave to the slave
 - iii. Wait for semaphore SemMaster from the slave
 - iv. Go back to i.
 - b) If the co-simulation mode is asynchronous
 - i. If the current communication time-point is 0
 - A. Release SemSlave to the slave
 - ii. If the current communication time-point is larger than 0
 - A. Wait for SemMaster from the slave
 - B. Release SemSlave to the slave

2.1.3.2 Implementation of functions in FMI2_Device_Master

Functions defined and implemented in this file are used for interaction between the FMI and EMTP solver. They are not constrained by the FMI standard and are executed by the master program.

- init

This function performs all necessary procedures in the initialization stage of the simulation. Its principal execution routines are:

- 1) Initialize all variables
- 2) Instantiate all FMI functions defined and implemented in the file FMI2_Link_Master
- 3) Unzip the FMU and place its contents in a temporary location
- 4) Read model data from the FMU

- saveDataPointers

This function has the following execution routines:

- 1) Load all FMI functions instantiated in the function init
- 2) Instantiate and allocate memory for the FMU
- 3) Call FMI function fmi2SetupExperiment and initialize certain variables used in co-simulation

- setFMIInput, getFMIOutput

These functions are in charge of data exchange with the slave FMU. They contain function calls to FMI functions fmi2Setxxx and fmi2Getxxx (xxx represents Real, Integer, Boolean or String) defined and implemented in the file FMI2_Link_Master.

- stepFunc

This is the function in the master that controls the co-simulation process as well as synchronization with the slave FMU. Its principal execution routines are as follows:

- 1) If the current master simulation time-point is 0 (initialization stage)
 - a) Call function fmi2EnterInitializationMode to start the initialization procedure
 - b) Call functions setIN, setINIE to assign values to certain variables
 - c) Call functions setFMUInput and getFMUOutput to write and retrieve initial values onto and from the co-simulation bus
- 2) If the current master simulation time-point is larger than 0 (co-simulation stage)
 - a) Call function fmi2ExitInitializationMode

- b) Call function `setFMUInput` to write inputs to the slave FMU onto the co-simulation bus
- c) Call function `fmi2DoStep` to synchronize with the slave FMU
- d) Call function `getFMUOutput` to read outputs from the slave FMU

- end

This function terminates the simulation by calling FMI function `fmi2Terminate`.

2.1.3.3 Implementation of functions in `FMI2_Link_Slave` and `FMI2_Device_Slave`

These two files are responsible for the communication between the slave instances and EMTP. In co-simulation, these files communicate with the FMI functions defined and implemented in the file `FMI2_Link_Master`, but not directly with the co-simulation interface. Therefore, their contents are not constrained by the FMI standard. The definition and implementation of chief functions in these two files are explained in this section.

- `CreateSlaveInstance`

This function is called by the slave program to create a slave instance for the co-simulation. It allocates memory for a new slave FMU object.

- `LinkSlaveInstance`

This function connects the slave instance with the master. Its main execution routines are:

- 1) Wait for the master to release semaphore `SemInitialization` (at the end of function `fmi2EnterInitializationMode`) after the master creates the co-simulation bus through file mapping
- 2) Create a map view of the file mapping object created by the master (co-simulation bus), establishing the connection between the slave FMU itself with the master
- 3) Send the slave FMU co-simulation mode (synchronous or asynchronous) to the master via the co-simulation bus

- 4) Release semaphore SemMaster to the master, informing the latter that the slave is now “linked”

- LinkSlaveIsReadyForStep

This function waits for the master to modify its outputs to the FMU in order to retrieve the current simulation time-point and numerical integration time-step of the master. It then updates the co-simulation bus with the current simulation time-point and numerical integration time-step of the slave. Its execution routines are:

- 1) Wait for the master to release semaphore SemSlave to the slave
- 2) Read master’s current simulation time-point and its numerical integration time-step from the co-simulation bus
- 3) Write the slave’s current simulation time-point and its numerical integration time-step onto the co-simulation bus

- LinkSlaveStepEnded

This function is called after the slave has advanced one step in time. It verifies if the slave lags behind the master and can ask the slave to catch up with the master depending on the specific co-simulation mode. At the end, it releases semaphore SemMaster to the master to authorize the latter to advance in its calculation.

- LinkSlaveTerminate

This function is called at the end of the simulation to terminate the slave instance.

- LinkSlaveReadInput, LinkSlaveSetOutput

These functions allow the slave instance to retrieve its inputs from the co-simulation bus before advancing one step in time or to write its output onto the co-simulation bus after the calculation of one step.

2.1.4 Function calling sequence in the master and slave programs

The master program is responsible for the entire co-simulation process. It is in charge of the communication with different FMUs and the general orchestration of the co-simulation. There

exists therefore an algorithm, namely the master algorithm, to fulfill the tasks of such a role. This algorithm is implemented in the DLL file FMI2_Device_Master. The master algorithm is not defined in the FMI standard. Hence, different implementations of such an algorithm are allowed. One possible solution is illustrated in Figure 2.4. It is given in the form of UML 2.0 state machine. More details about this solution can be found on Page 103 in [19].

The master algorithm controls the co-simulation process by calling the FMI functions defined and implemented in the DLL file FMI2_Link_Master and coordinating them with the EMTP solver. Considering the specifics of the implementation platform (the EMT-type solver in [2]), the master algorithm adopted in this thesis (communication between the two DLL files FMI2_Device_Master and FMI2_Link_Master) in terms of function calling sequence is presented in Figure 2.5. It is noted that the dotted red arrows denote that functions in the file FMI2_Link_Master are called by those in the file FMI2_Device_Master. Based on Figure 2.5, the master algorithm can be summarized as follows:

1. Initialization

- a. Unzip slave FMUs and instantiate master-slave interface instances
- b. Load FMI functions from the DLL file and read model data from the xml file
- c. Ask the slaves to configure the co-simulation
- d. Create co-simulation bus
- e. Initialize synchronization semaphores

2. Co-simulation

- a. Exchange data and synchronize with slaves
- b. Solve for the power network, advance one time-step then go back to step 2a

3. Terminate co-simulation

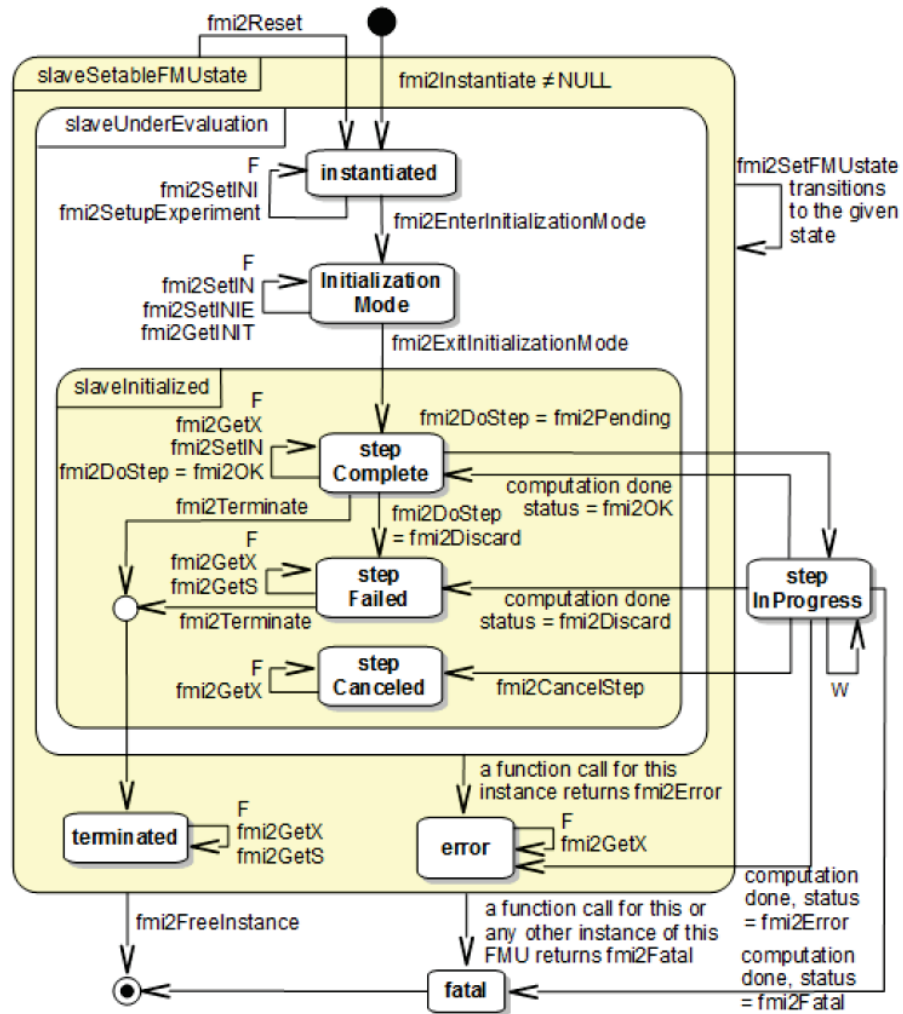


Figure 2.4 One possible implementation of the master program of co-simulation of an UML 2.0 state machine

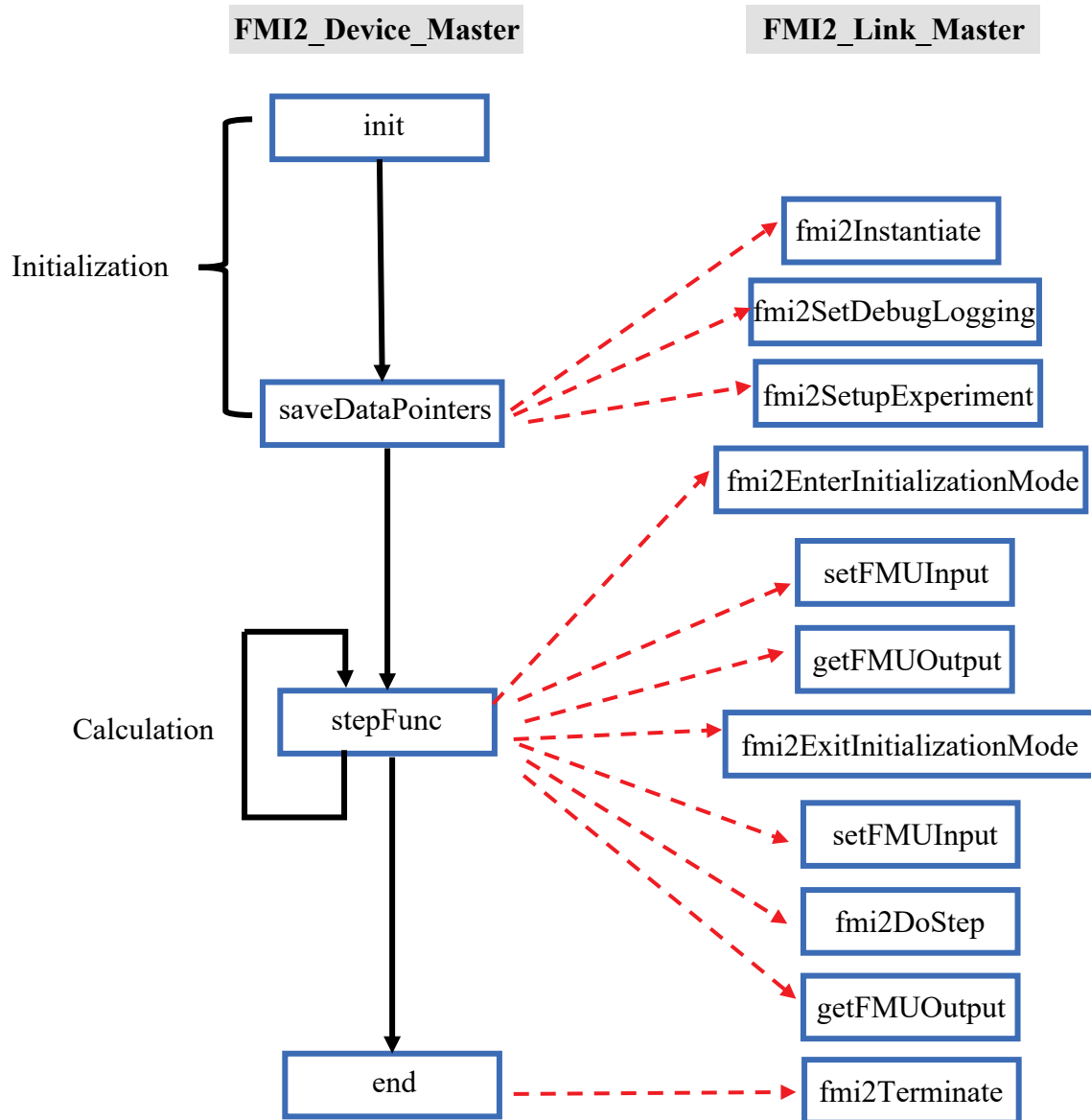


Figure 2.5 Principal function calling sequence in the master algorithm

The slave algorithm answers to the master by performing tasks demanded by the master. The principal function calling sequence in the slave algorithm adopted in this thesis, based on the definition and implementation of slave functions explained in the previous section, is presented in Figure 2.6. Once again, the dotted red arrows indicate that functions in the DLL file FMI2_Link_Slave are called by those in the DLL file FMI2_Device_Slave. It is to mention that the functions LinkSlaveSetOutput and LinkSlaveStepEnded inside the DLL file FMI2_Link_Slave (as enclosed in the dotted green rectangle with rounded corners) are only called by the function LoadObservables for the calculation of the first time-step.

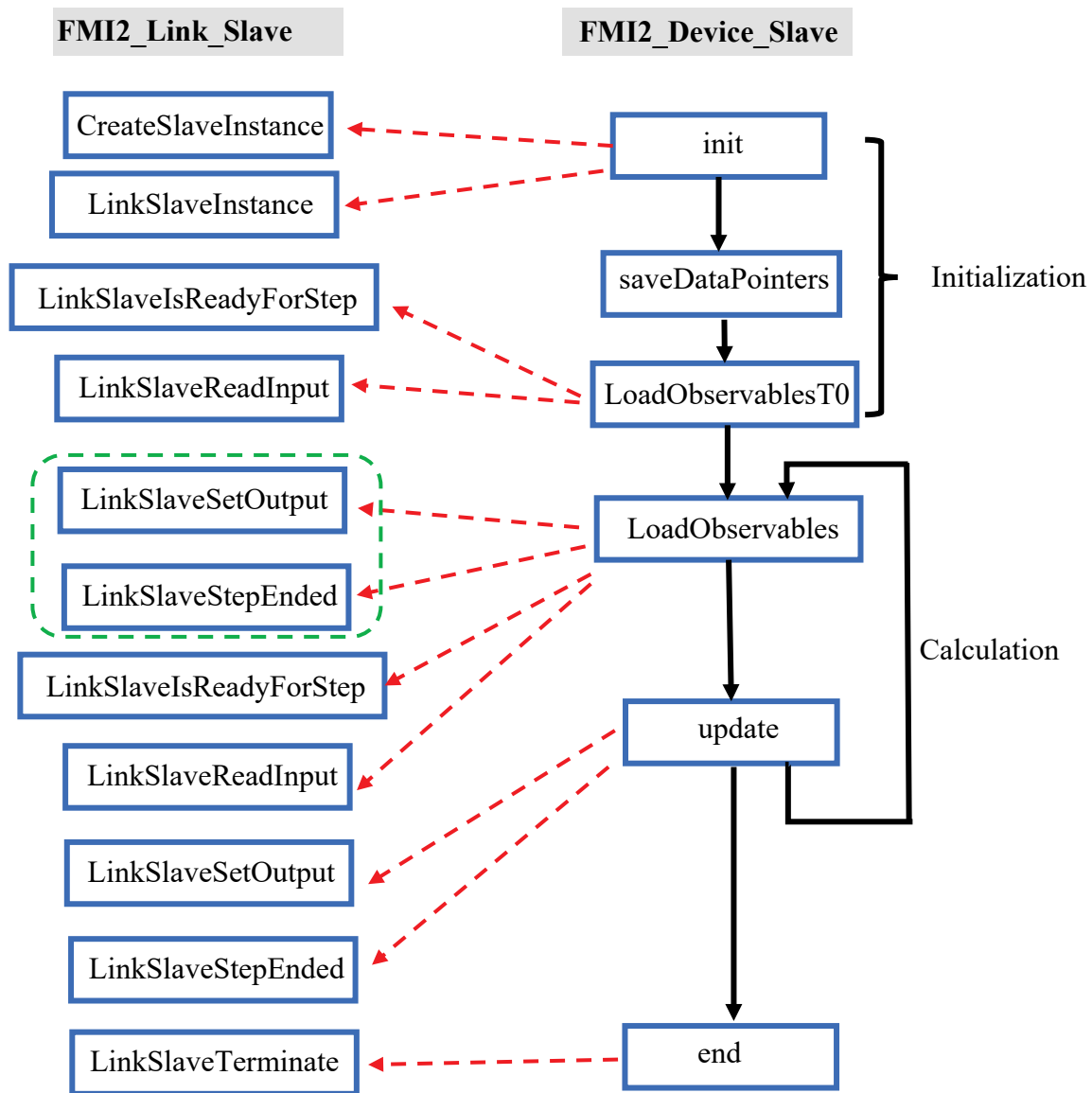


Figure 2.6 Principal function calling sequence in the slave algorithm

As is observed in Figure 2.6, the slave algorithm can be described as follows:

1. Initialization
 - a. Create slave instances and load slave DLL functions
 - b. Connect slave instances to the co-simulation bus after the latter is created
 - c. Solve for the control system at $t = 0$
2. Co-simulation
 - a. Exchange data and synchronize with the master
 - b. Solve for the control system, advance to the next time-point, then go back to step 2a
3. Terminate slave instances

2.2 Co-simulation management

This section tackles the synchronization between master and slaves using low-level synchronization primitive semaphore as well as master-slave data exchange with the help of file mapping. Code details of important functions used in semaphore management are presented, and two co-simulation modes (synchronous and asynchronous) are introduced. The structure and implementation of the co-simulation bus used in master-slave data exchange are explained.

2.2.1 Synchronization between master and slaves

The synchronization between the master and slave subsystems is achieved through the implementation of low-level synchronization primitive semaphores in the DLL files FMI2_Link_Master and FMI2_Link_Slave. A semaphore, serving both as a lock and a condition variable, is a variable used to control read and write access to a shared resource by multiple processes or threads in a concurrent programming environment to prevent race conditions and other concurrency problems [85] (see also [16]). By calling the pre-built function WaitSemaphoreTimeout on a specified semaphore, the current program (either master or slave) temporarily halts its operation until the aforementioned semaphore is signaled, which indicates the action of “waiting for a semaphore”; a call to the pre-built function ReleaseSemaphore on a

specified semaphore signals it and permits any program that is waiting for this semaphore to continue with their operation, representing thus the action of “releasing a semaphore”.

2.2.1.1 Semaphore management

The implementation of the two pre-built functions `WaitSemaphoreTimeOut` and `ReleaseSemaphore` are presented as follows:

- `WaitSemaphoreTimeOut`

```
bool WaitSemaphoreTimeOut(ModelInstance *comp, int SemNumber, int NumSemToWait, int
TimeOut)
{
    int ctr;
    HANDLE Sem1 = NULL;
    Sem1 = allocateSemaphore(comp->instanceName, SemNumber, comp->nbInputPins +
comp->nbOutputPins);
    if (Sem1 != NULL)
    {
        for (ctr = 0; ctr < NumSemToWait; ctr++)
        {
            DWORD ret = WaitForSingleObject(Sem1, TimeOut);
            DWORD err = GetLastError();
            if (ret != WAIT_OBJECT_0)
            {
                return false;
            }
        }
        return true;
    }
    return false;
}
```

- `ReleaseSemaphore`

```
bool ReleaseSemaphore(ModelInstance *comp, int SemNumber, int NumSemToRelease)
{
    HANDLE Sem1;
    Sem1 = allocateSemaphore(comp->instanceName, SemNumber, comp->nbInputPins +
comp->nbOutputPins);
    try
    {
        ReleaseSemaphore(Sem1, (LONG)NumSemToRelease, (LPLONG)NULL);
    }

    catch (Exception^ ex)
    {
        ex;
        return(false);
    }
    return true;
}
```

where:

- comp: of type `ModelInstance*`, refers to a specific slave instance (FMU).
- SemNumber: of type integer, indicates the number of the semaphore, which is attached to the name of the slave instance to uniquely identify the semaphore.
- NumSemWait: of type integer, is the number by which the semaphore count is decreased.
- NumSemToRelease: of type integer, is the number by which the semaphore count is increased.

An auxiliary function named `allocateSemaphore`, whose functionality is self-evident, is also used in the implementation of the aforementioned two pre-built functions:

- `allocateSemaphore`

```
HANDLE allocateSemaphore(fmi2String name_instance, int numSem, int nb_Pins_max)
{
    HANDLE hdl = NULL;
    DWORD ret = 0;
    char name[256];
    map<string, HANDLE>::iterator it;
    sprintf_s(name, "%sSem%d", name_instance, numSem);
    it = sem_hdl.find(name);
    if (it == sem_hdl.end())
    {
        hdl = OpenSemaphoreA(SEMAPHORE_ALL_ACCESS, false, name);
        ret = GetLastError();
        if (hdl == NULL)
        {
            hdl = CreateSemaphoreA(NULL, 0, nb_Pins_max + 1, name);
            ret = GetLastError();
        }
        sem_hdl[name] = hdl;
    }
    else
    {
        hdl = (HANDLE)it->second;
    }
    return hdl;
}
```

where:

- nb_Pins_max: of type integer, defines the maximum count of the semaphore.

Another function used in semaphore management in the algorithm is named `closeSemaphore`, whose implementation is as follows:

- closeSemaphore

```
bool closeSemaphore(fmi2String name_instance, int numSem)
{
    HANDLE hdl;
    hdl = allocateSemaphore(name_instance, numSem, 1);
    if (hdl)
    {
        char name[256];
        map<string, HANDLE>::iterator it;
        it = sem_hdl.find(name);
        if (it != sem_hdl.end())
        {
            sem_hdl.erase(it);
        }
        CloseHandle(hdl);
    }
    return true;
}
```

A total of three semaphores are employed in the initialization and co-simulation stages in both the master and slave algorithms, with the names SemInitialization, SemMaster and SemSlave. The first one is used in the initialization mode, allowing a slave instance to connect to the co-simulation bus created by the master. The second one is released by the slave to the master after the former finishes certain operations. Similarly, the third one is released by the master to the slave, permitting the latter to continue with its operations. For demonstrative purposes, only one slave instance is considered in the following discussions.

2.2.1.2 Master-slave synchronization: initialization mode

After loading each FMU, the master starts the simulation and enters the initialization stage by calling function fmi2EnterInitializationMode, in which the master creates the co-simulation bus, releases SemInitialization to the slave and writes initial conditions onto the co-simulation bus whilst the slave waits for SemInitialization. Once the slave receives SemInitialization, it connects to the co-simulation bus, writes its initial parameters onto it and releases SemMaster. As a result, the master receives SemMaster from the slave and finishes the initialization stage. The synchronization scheme between master and slave during the initialization stage is illustrated in Figure 2.7. It is worth noting:

- 1) Every arrow represents an “exchange” of semaphore between the master and slave.

- 2) The start of the arrow represents the call to function ReleaseSemaphore, in which the program involved “releases” the semaphore by increasing its count.
- 3) The end of the arrow represents the call to function WaitForSemaphoreTimeout, in which the program involved “waits for” the semaphore by decreasing its count in order to further continue its operations.

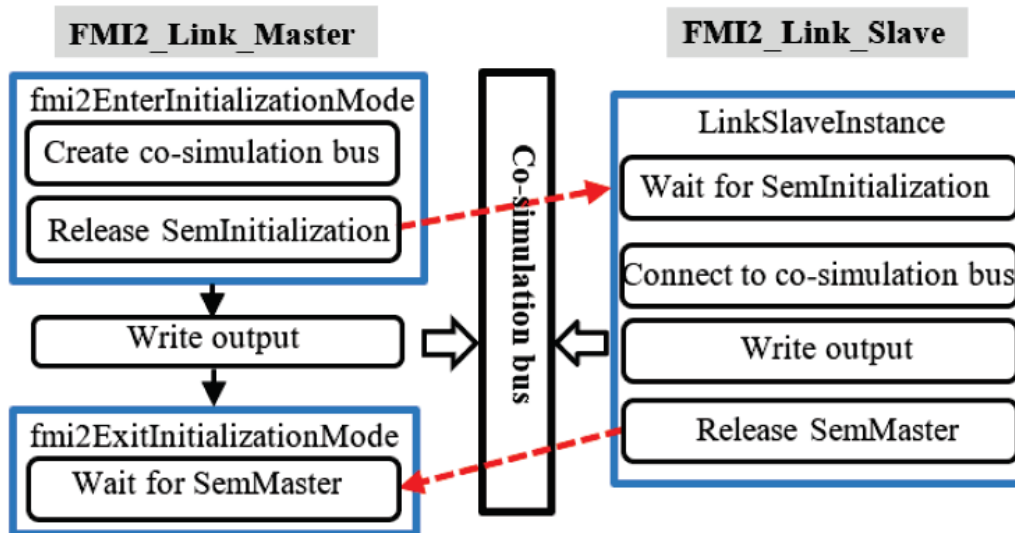


Figure 2.7 Synchronization scheme between master and slave during initialization stage

2.2.1.3 Master-slave synchronization: synchronous mode (with a single numerical integration time-step)

As was stated earlier on in this thesis, two co-simulation modes are designed for the new off-line simulation approach: the asynchronous and synchronous mode. The synchronous mode, despite its sequential nature, allows not only using a single time-step in all master and slave subsystems, but also different numerical integration time-steps in the master and each slave subsystem.

For the synchronous mode with a single numerical integration time-step in both master and slave, as is shown in Figure 2.8, the master releases SemSlave to the slave such that the slave could read inputs from the master, write output to the master, release SemMaster to the master, solve for the current time-point and advance one time-step. Consequently, the master receives SemMaster from the slave, retrieves the outputs from the slave and continues its own execution by advancing one time-step as well. Hence, the execution of master and slave(s) is performed sequentially.

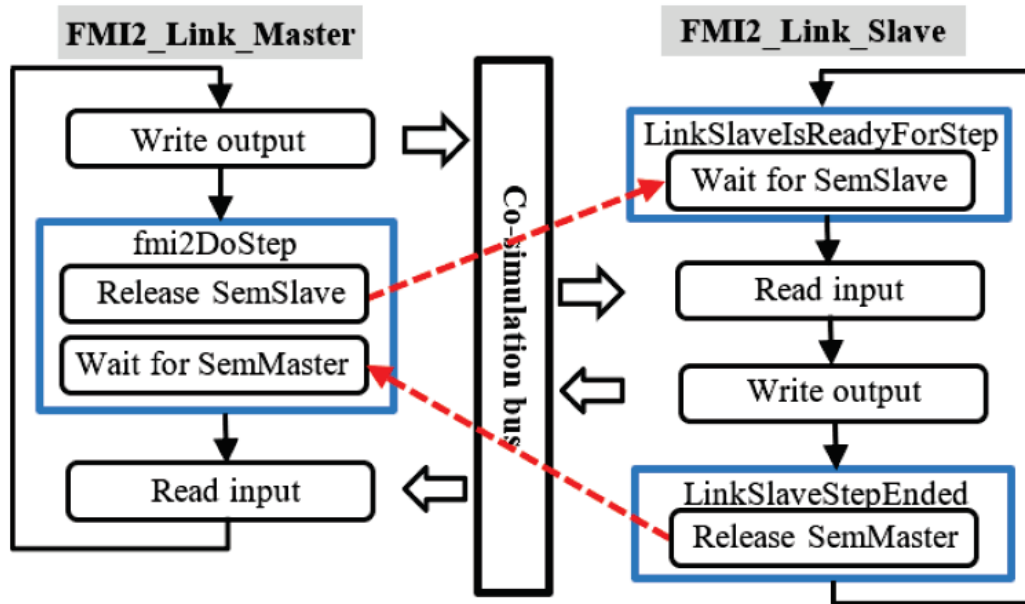


Figure 2.8 Synchronization scheme between master and slave in synchronous mode (with a single numerical integration time-step)

2.2.1.4 Master-slave synchronization: asynchronous mode (with a single numerical integration time-step)

In the asynchronous mode, the solutions of master and slave subsystems are executed in parallel with a single numerical integration time-step on separate logical processors.

Specifically, in the asynchronous mode with a single numerical integration time-step in both master and slave, the master first waits for the slave to release SemMaster after the solution of the previous time-point. It then immediately releases SemSlave to the slave such that the slave can solve for its current time-point and advance one time-step in time. Meanwhile, the master solves for the current time-point likewise, advances one time-step and again waits for the release of SemMaster from the slave, as is illustrated in Figure 2.9. Therefore, in the asynchronous mode, the master and the slave(s) can each execute their computations simultaneously in parallel.

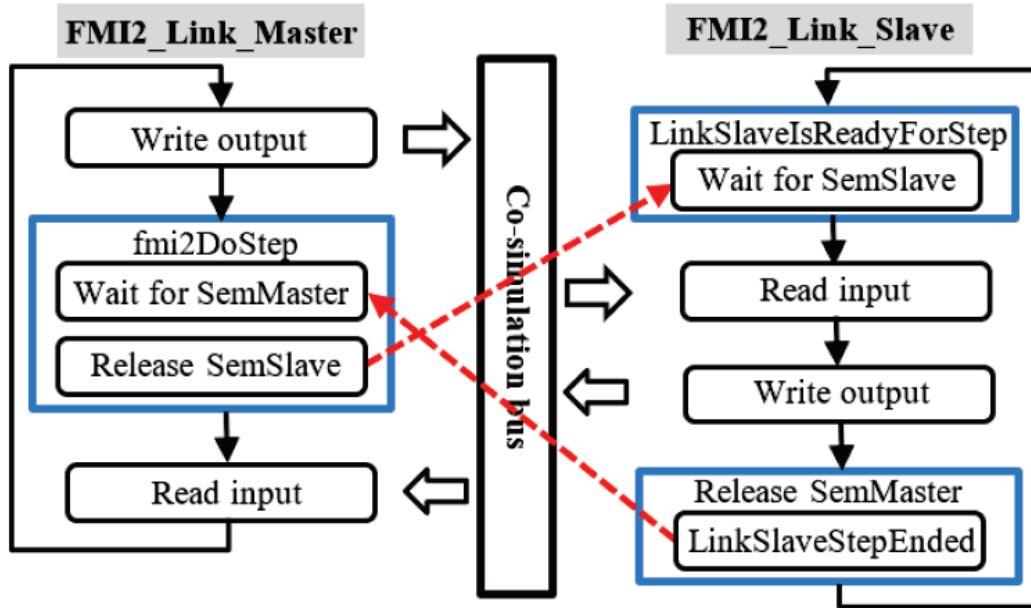


Figure 2.9 Synchronization scheme between master and slave in asynchronous mode (with a single numerical integration time-step)

2.2.1.5 Master-slave synchronization: synchronous mode (with different numerical integration time-steps)

In the synchronous mode with different numerical integration time-steps utilized in master and slave subsystems, the principle of operation is similar to that of the synchronous mode with a single numerical integration time-step. Nonetheless, a few changes have been made in the FMI function `fmi2DoStep`, as is shown in Figure 2.10. Before releasing and waiting for any semaphores, the `fmi2DoStep` function first compares the current time-point of the master and the slave. If the master lags behind the slave, which occurs when the master's numerical integration time-step is smaller than that of the slave, function `fmi2DoStep` terminates without performing any tasks and the master continues executing its calculation in time without releasing or waiting for any semaphores until its current simulation time-point catches up with that of the slave. On the contrary, if the slave lags behind the master with the numerical integration time-step of the slave smaller than that of the master, the master keeps releasing `SemSlave` to the slave whilst staying in the loop in order that the slave continues advancing in time and executing its calculation until it no longer lags behind the master. The master thereby retrieves the outputs from the slave and continues its execution. Once again, the execution of master and slave(s) is performed in sequence.

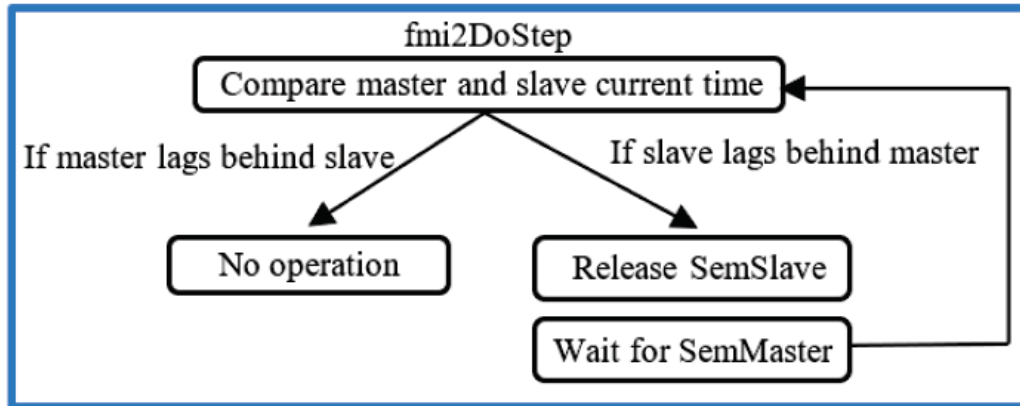


Figure 2.10 Function `fmi2DoStep` in synchronous mode (with different numerical integration time-steps)

The discussion of the synchronization scheme between master and slave in the synchronous mode is divided into two sections (2.2.1.3 and 2.2.1.5) for clarity. In fact, it is the synchronization scheme illustrated in Figure 2.11 that is implemented in the algorithm, combining the scenarios in the two sections.

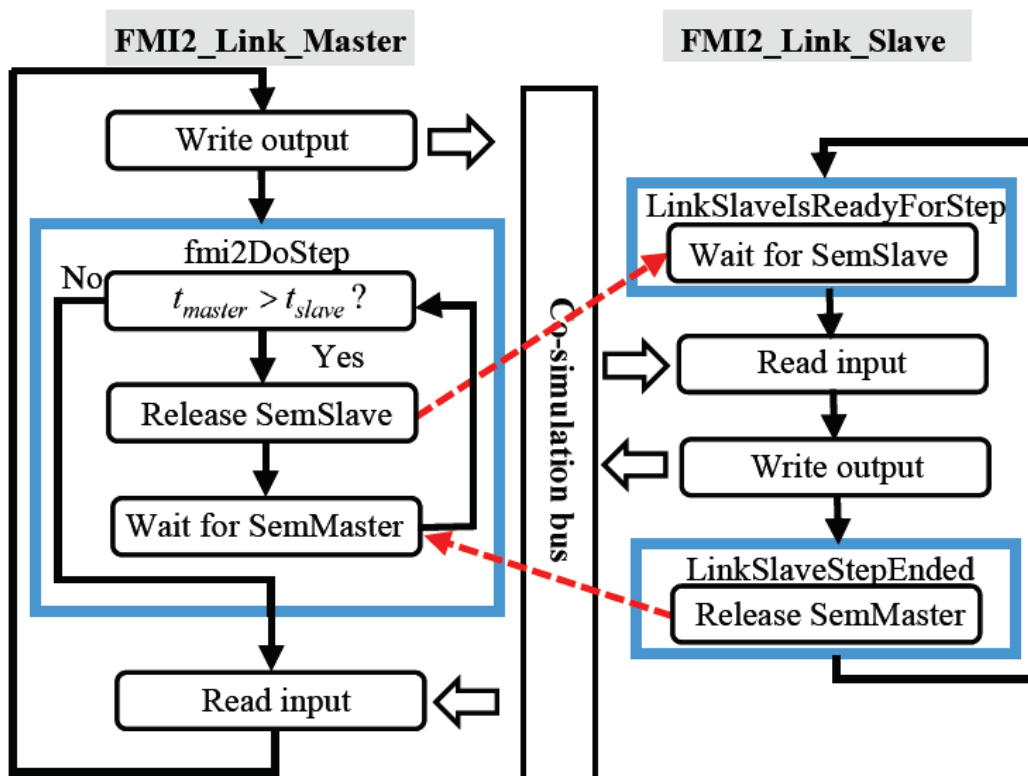


Figure 2.11 Synchronization scheme between master and slave in synchronous mode

2.2.2 Data exchange between master and slaves

The co-simulation bus, which is implemented with file mapping, is used to exchange data between master and slaves in the algorithm. File mapping is the association of a file's contents with a portion of the virtual address space of a process [86]. The system creates a file mapping object to maintain this association. A file view is the portion of virtual address space that a process uses to access the file's contents. File mapping allows the process to work efficiently with a large data file, such as a database, without having to map the whole file into memory. Multiple processes can also use memory-mapped files to share data [87].

In the algorithm, the file mapping object (co-simulation bus) is created in the function `fmi2EnterInitializationMode` in the DLL file `FMI2_Link_Master`. Additionally, function `LinkSlaveInstance` in the DLL file `FMI2_Link_Slave` creates a file view, indicating the slave is "linked" to the co-simulation bus created by the master. The code snippets of implementation of file mapping in functions `fmi2EnterInitializationMode` and `LinkSlaveInstance` are presented as follows:

- Snippet of file mapping in function `fmi2EnterInitializationMode`

```
comp->sizetb = sizeof(double)*(tab_index_InOut + comp->nbOutputPins +
comp->nbInputPins);
comp->hFileMap = CreateFileMapping(INVALID_HANDLE_VALUE, 0, PAGE_READWRITE, 0,
comp->sizetb, TCHAR_InstanceName);
InOut = MapViewOfFile(comp->hFileMap, FILE_MAP_ALL_ACCESS, 0, 0, comp->sizetb);
```

The function `fmi2EnterInitializationMode` first creates a file mapping object using function `CreateFileMapping`, then it maps a view of the file mapping and returns the starting address of the mapped view to a container `InOut`. The container `InOut`, therefore, serves as the co-simulation bus in the algorithm.

- Snippet of file mapping in function `LinkSlaveInstance`

```
instanceFmiSlave->hMapFile = OpenFileMapping(
    FILE_MAP_ALL_ACCESS,
    FALSE,
    szName);

if (instanceFmiSlave->hMapFile==NULL)
{
    SendErrorMsg(instanceFmiSlave, "LinkSlaveInstance", Def_Error_Msg12) ;
    return(false);
}
```

```

buffer1= (LPVOID) MapViewOfFile(instanceFmiSlave->hMapFile,
    FILE_MAP_ALL_ACCESS,
    0,
    0,
    size1);

CopyMemory(bufferNbinOut1, buffer1,size1);

instanceFmiSlave->nbOutputPins=int(bufferNbinOut1[tab_index_Nbout]);
instanceFmiSlave->nbInputPins=int(bufferNbinOut1[tab_index_Nbin]);
instanceFmiSlave->sizeBuf=sizeof(double)*(tab_index_InOut+instanceFmiSlave->nbOut
putPins+instanceFmiSlave->nbInputPins);
InOut = (LPVOID) MapViewOfFile(instanceFmiSlave->hMapFile,
    FILE_MAP_ALL_ACCESS,
    0,
    0,
    instanceFmiSlave->sizeBuf);

```

Subsequently, the slave instance creates a mapped view using the handle returned by the file mapping object created in the function `fmi2EnterInitializationMode` and “links” itself to the co-simulation bus. The structure of the container `InOut` (co-simulation bus) is illustrated in Table 2.4.

Table 2.4 Structure of the container `InOut` (co-simulation bus)

Contents	Index
NbIn	0
NbOut	1
Slave Mode	2
Time Out	3
Time Slave	4
Step Slave	5
Time Master	6
Step Master	7
Error Master	8
Error Slave	9
Y 1	10
...	...
Y NbIn	9+NbIn
X 1	10+NbIn
...	...
X NbOut	9+NbIn+NbOut

where:

- NbIn: the number of inputs of the slave instance.
- NbOut: the number of outputs of the slave instance.

- Slave Mode: the chosen co-simulation mode of the slave instance.
- Time Out: the chosen maximum synchronization time.
- Time Slave: the slave simulation time-point.
- Step Slave: the slave numerical integration time-step.
- Time Master: the master simulation time-point.
- Step Master: the master numerical integration time-step.
- Error master: the value is 1 if the master signals an error; 0 if not.
- Error slave: the value is X if the X^{th} input or output of the slave cannot be updated; 0 if not.
- Y and X: different inputs and outputs of the slave instance.

The total size of the container InOut is thus $\text{NbIn} + \text{NbOut} + 10$.

2.3 Improvements on the original methodology

In the previous section, two co-simulation modes, the parallel asynchronous mode and the sequential synchronous mode, were proposed, in which only the latter could accommodate the use of different numerical integration time-steps in different decoupled subsystems. Based on this methodology and continuing with the master-slave co-simulation scheme adopted in this thesis, in this section, the computation capacity of the parallel asynchronous mode is extended into accommodating the use of different numerical integration time-steps in different subsystems decoupled in memory, greatly improving simulation flexibility and efficiency. Furthermore, a signal correction procedure based on linear extrapolation is introduced to achieve higher accuracy in a multistep simulation environment.

2.3.1 Parallel multistep asynchronous mode

The improved asynchronous mode is divided into 3 scenarios based on the comparison of master and slave numerical integration time-steps. The synchronization scheme between master and slave in the improved parallel multistep asynchronous mode is presented in Figure 2.12. It is noted that the 3 scenarios are indicated by the 3 blue boxes, and only one scenario is executed in practice.

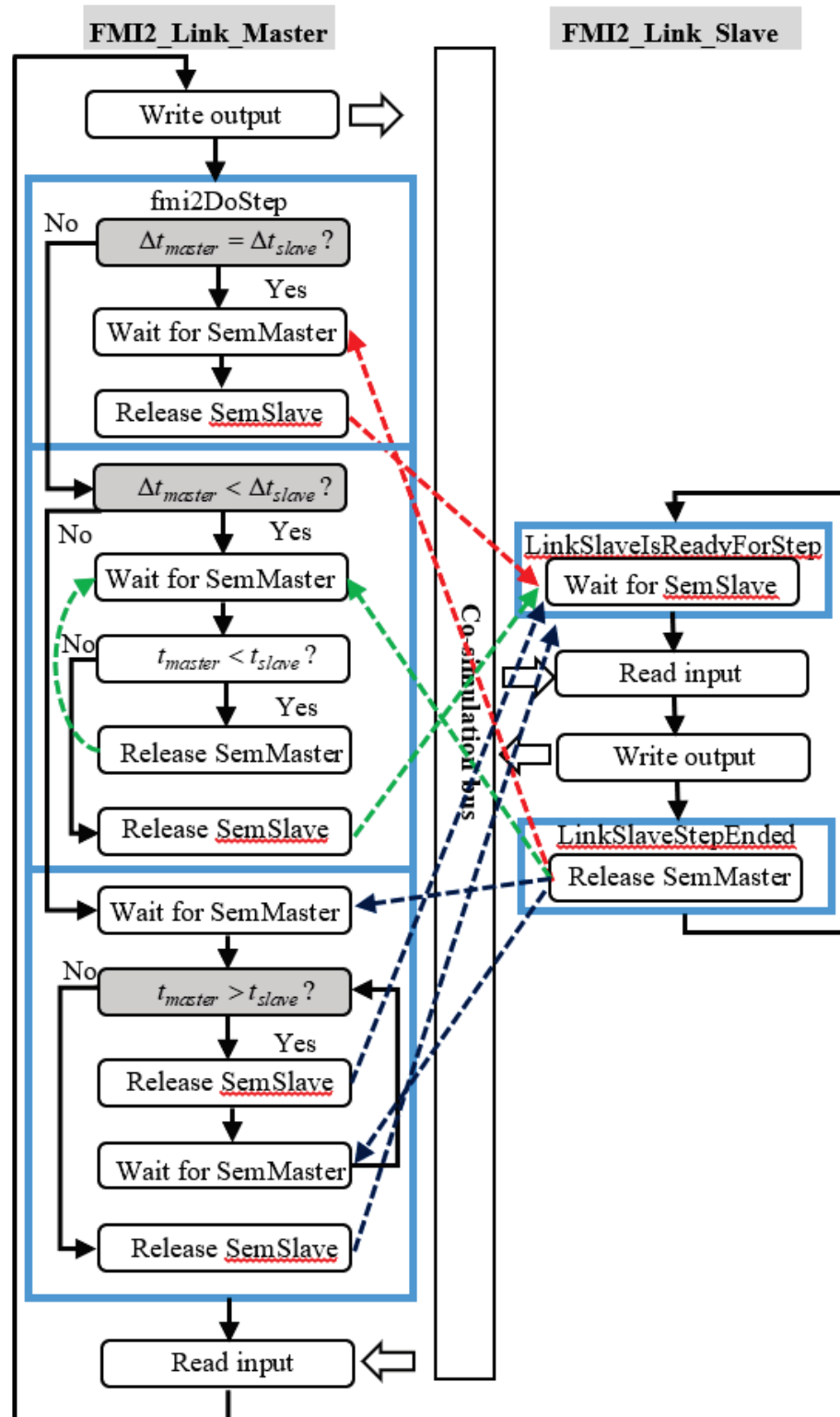


Figure 2.12 Synchronization scheme between master and slave in the improved asynchronous mode (parallel multistep asynchronous mode)

The improved master-slave synchronization scheme for the parallel multistep asynchronous illustrated in Figure 2.12 might seem daunting due to the level of complexity. A scenario-wise explanation is hence needed to understand its principle of operation. The 3 different scenarios are separated based on the comparison of master and slave numerical integration time-steps and are presented respectively in Figure 2.13, Figure 2.14 and Figure 2.15. It is worthwhile to note that all arrows, in spite of their colors, denote the release of semaphore.

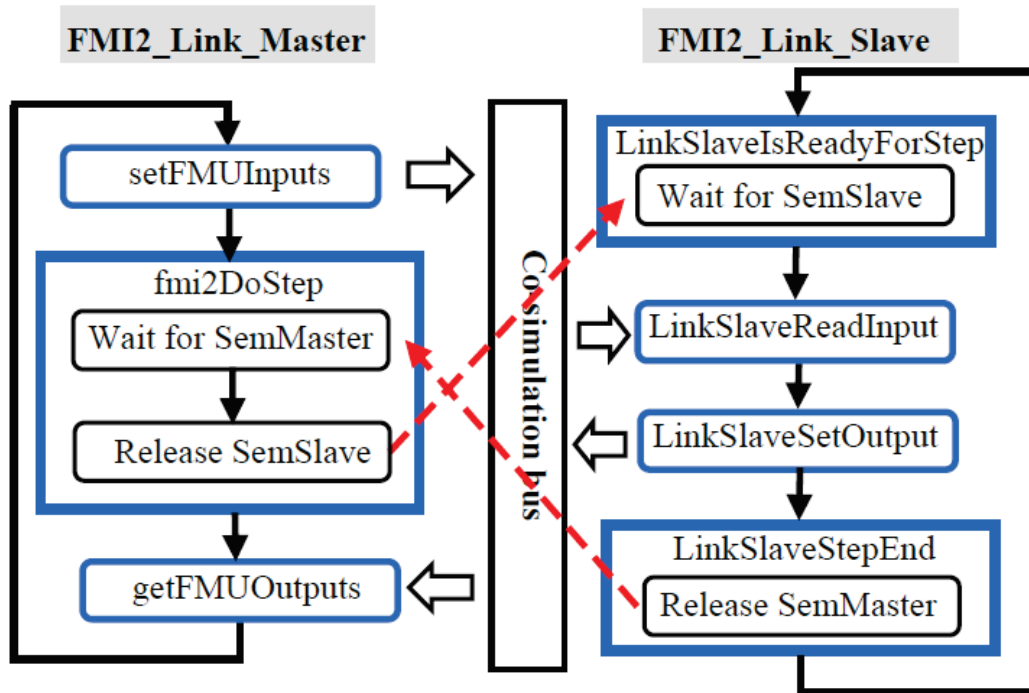


Figure 2.13 First synchronization scenario ($\Delta t_{master} = \Delta t_{slave}$) between master and slave in the parallel multistep asynchronous mode

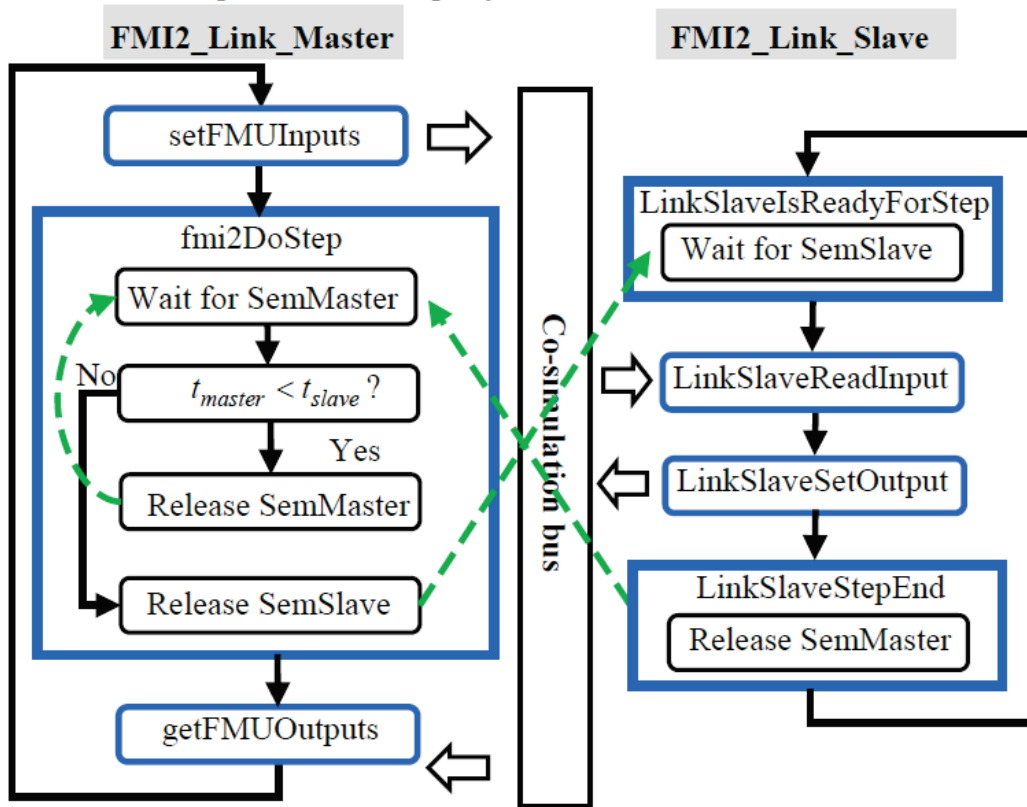


Figure 2.14 Second synchronization scenario ($\Delta t_{master} < \Delta t_{slave}$) between master and slave in the parallel multistep asynchronous mode

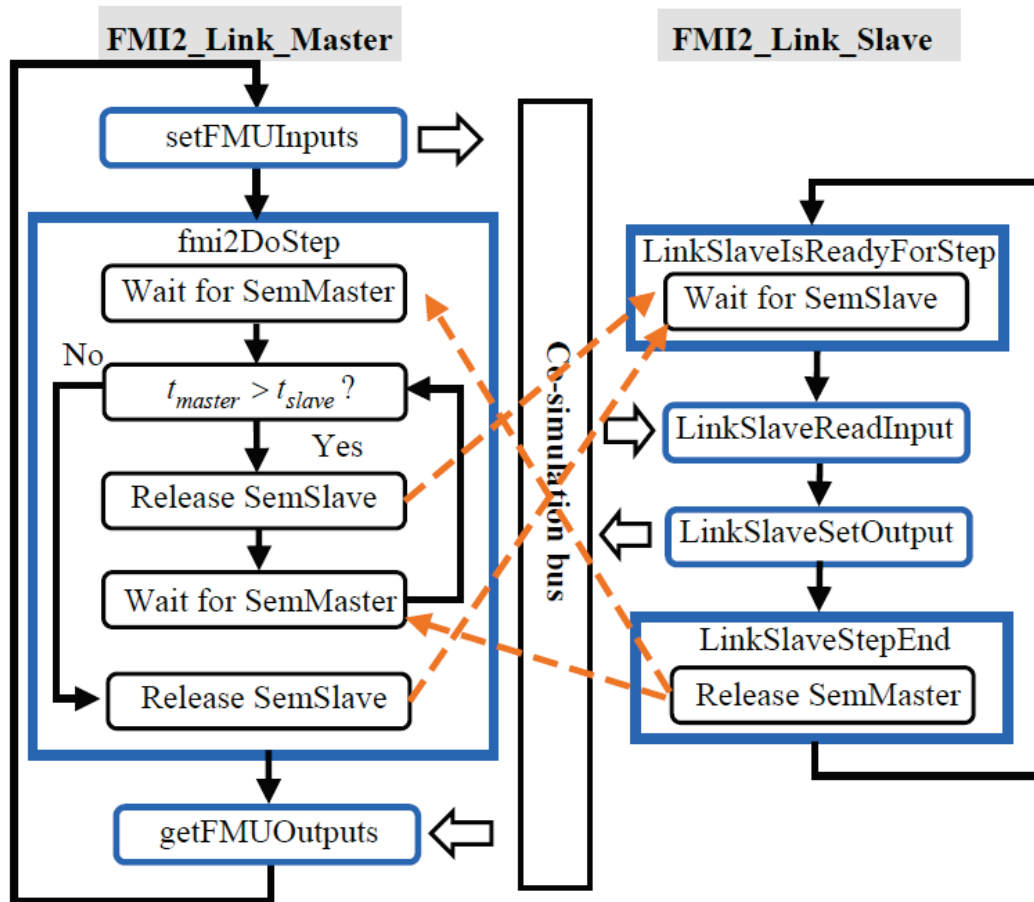


Figure 2.15 Third synchronization scenario ($\Delta t_{master} > \Delta t_{slave}$) between master and slave in the parallel multistep asynchronous mode

The first scenario (Figure 2.13) in which $\Delta t_{master} = \Delta t_{slave}$ is the case implemented in the original methodology discussed in the previous section. Figure 2.13 and Figure 2.9, therefore, describe the same principle of operation.

In the second scenario where $\Delta t_{master} < \Delta t_{slave}$, as is shown in Figure 2.14, after waiting for the slave to release SemMaster, the master compares its current simulation time-point with that of the slave. If the master lags behind the slave due to having a smaller numerical integration time-step, it releases SemMaster to itself, solves for its current time-point and advances one step in time. It keeps releasing SemMaster to itself so that it can advance without the help from the slave till its current simulation time-point catches up with that of the slave. Once the master catches up with the slave, it releases SemSlave to the slave such that both can execute their subsequent calculation simultaneously.

If $\Delta t_{master} > \Delta t_{slave}$ which is the third scenario (Figure 2.15), after receiving SemMaster from the slave, the master, as in the second scenario, compares its current simulation time-point with that of the slave. If the master is ahead of the slave due to having a larger numerical integration time-step, it releases SemSlave to the slave immediately such that the latter could read and write data, solve for its current time-point, advance one step in time and release SemMaster back to the master. The master subsequently receives SemMaster, once again compares its current simulation time-point with that of the slave and keeps releasing SemSlave to the slave till the slave current simulation time-point catches up with that of the master. Once the slave catches up, the master releases SemSlave once more and both continue their subsequent calculations in parallel.

Such an improved synchronization mechanism allows multiple numerical integration time-steps to be employed in the parallel asynchronous mode, considerably enhancing simulation flexibility with the use of single or multiple numerical integration time-steps in both co-simulation modes.

2.3.2 Linear extrapolation-based signal correction

Floating-point type signals from slave subsystems employing a larger time-step can be extrapolated using linear extrapolation in the master for enhanced simulation accuracy in a multistep simulation environment. This procedure is performed in the pre-built function getFMUOutput in the DLL file FMI2_Device_master, whose main functionality is to read data exported from the slave FMU with the help of function calls to FMI function fmi2Get $\times\times\times$ (with $\times\times\times$ representing Real, Integer, Boolean or String) defined and implemented in the file FMI2_Link_Master.

Two new arrays are defined, which are **extrapol_outputs** and **extrapol_time_points** respectively:

```
fmi2Real* extrapol_outputs;
fmi2Real* extrapol_time_points;
```

Array **extrapol_outputs** is used to store the values for extrapolation, and array **extrapol_time_points** stores the corresponding time-points of these values. If array \mathbf{y}_{slave} denotes floating-point type output from the slave whose size is n (number of floating-point type signals) at $t = T$, and array **extrapol_outputs** of size $2n$ stores floating-point type outputs from the slave at the previous two slave time-points ($t = T - \Delta t_{slave}$ and $t = T - 2\Delta t_{slave}$) used for extrapolation:

$$\mathbf{extrapol_outputs} = [\overbrace{\mathbf{y}_{slave,T-\Delta t_{slave}}}^n, \overbrace{\mathbf{y}_{slave,T-2\Delta t_{slave}}}^n] \quad (3.1)$$

$$\mathbf{extrapol_time_points} = [T - \Delta t_{slave}, T - 2\Delta t_{slave}] \quad (3.2)$$

The improved implementation of function getFMUOutput is presented as follows, with parts dealing with other types of FMU outputs omitted for clarity:

- 1) Read $\mathbf{y}_{slave,T}$ (floating-point type outputs from the slave FMU at $t = T$) and store it in a temporary array realOutputValues.
- 2) Set Boolean variable isIdentical to true
- 3) Compare $\mathbf{y}_{slave,T}$ with $\mathbf{y}_{slave,T-\Delta t_{slave}}$ (the first n elements of the array **extrapol_outputs**, which is the slave FMU real output values at $t = T - \Delta_{slave}$), and set isIdentical to false if these two sets of data are not identical
- 4) If isIdentical is always equal to true, indicating the two sets of data compared in step 3) are identical, extrapolate $\mathbf{y}_{slave,T}$ using $\mathbf{y}_{slave,T-\Delta t_{slave}}$ and $\mathbf{y}_{slave,T-2\Delta t_{slave}}$ (the two sets of elements, each is of size n , stored in **extrapol_outputs**) and the corresponding time-points stored in **extrapol_time_points**
- 5) If isIdentical is false, indicating that the two sets of data compared in step 3) are different from one another, update values in arrays **extrapol_outputs** and **extrapol_time_points** only:

$$\mathbf{extrapol_outputs}[n+1:2n] = \mathbf{extrapol_outputs}[1:n]$$

$$\mathbf{extrapol_outputs}[1:n] = \mathbf{y}_{slave}$$

$$\mathbf{extrapol_time_points}[2] = \mathbf{extrapol_time_points}[1]$$

$$\mathbf{extrapol_outputs}[1] = T$$

- 6) Values stored in `realOutputValues` (extrapolated or not) are transferred into array **outputs** for further use in EMTP

It is worth pointing out:

- In step 4) where `isIdentical` is true, the two sets of data compared are identical, which indicates that the slave FMU with a larger numerical integration time-step remains idle, waiting for the master to catch up whilst the master keeps reading in the not-yet-updated slave FMU output values stored in the co-simulation bus.
- In step 5) where `isIdentical` is false, the two sets of data compared are not identical, which suggests that the slave FMU has advanced one step in time and new output values have been calculated and updated in the co-simulation bus.
- Due to insufficient data, the slave FMU output values between $t=0$ and the first master-slave communication point (in this case where $\Delta t_{slave} > \Delta t_{master}$, $t = \Delta t_{slave}$) are not available through extrapolation.

2.4 Conclusion

This chapter presented the detailed implementation of the Functional Mock-up Interface (FMI) standard in EMTP as well as a parallel and multistep approach based on co-simulation between different EMTP solver instances using such an implementation for power system EMT simulations with control systems. At first, a brief introduction of the FMI standard and its advantages were provided. Then the contents of a Functional Mock-up Unit (FMU) used as the co-simulation interface as well as its design constraints in order to establish the compatibility between FMI and EMTP were introduced. Concrete implementation of the FMI functions tailored to co-simulation in EMTP was elaborated, and different master-slave synchronization mechanisms using semaphore for two co-simulation modes, the parallel asynchronous mode and the sequential synchronous mode, were then proposed. The former allows decoupled subsystems to be simulated in parallel on separate logical processors using a single numerical integration time-step, whereas the latter accommodates the use of multiple numerical integration time-steps in different decoupled subsystems in a sequential simulation environment. Finally, two improvements of the original methodology, which are the parallel multistep asynchronous co-simulation mode and the linear

extrapolation-based signal correction procedure on floating-point slave outputs, were presented with the goal of enhancing simulation flexibility, efficiency and accuracy.

CHAPTER 3 TEST CASES AND SIMULATION RESULTS OF THE FMI-BASED APPROACH

This chapter presents simulation results of the FMI-based parallel and multistep approach developed in this thesis. It first validates the design of the co-simulation interface in EMTP using the FMI standard through a simple example, then proceeds to validate both the original and the improved methodologies. The validation focuses on two aspects: accuracy and computation time gains. Each aspect is validated using realistic power system simulation benchmarks with complex control systems (wind generator controls and protection relays). Advantages of the proposed FMI-based approach in terms of accuracy, efficiency and scalability are fully demonstrated.

3.1 Validation of the co-simulation interface design

In this section, the design of the co-simulation interface using the FMI standard on the EMT-type solver [2] is validated through the demonstration of a simple example. Simulation results obtained from both co-simulation modes are compared with those using EMTP without co-simulation. The validation of the co-simulation interface design confirms that the implementation of the FMI standard on the EMT-type solver [2] is successful, and further tests using realistic power system simulation benchmarks can thereby be conducted.

3.1.1 Creation of master and slave models

The complete test circuit using the EMT-type solver [2] for the validation of the co-simulation interface design is shown in Figure 3.1. This test circuit consists of two “real” type inputs which are trigonometric functions sine and cosine and two “real” type outputs which are the sum of the two inputs squared and the product of these two inputs respectively. As was mentioned previously, the FMI standard is implemented in EMTP for control signals (integer, floating-point or Boolean) that can be exchanged by means of the co-simulation interface in this thesis. Therefore, all the circuit elements are represented using control blocks. In the original EMTP simulation without using the FMI-based approach, the simulation interval is 100 ms with a numerical integration time-step of 100 μ s.

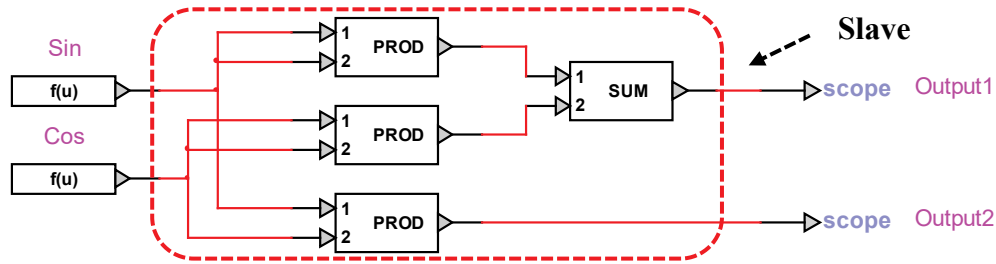


Figure 3.1 Test circuit for the validation of co-simulation interface design

In the FMI-based co-simulation approach, the section inside the red dotted rectangle with rounded corners, as is shown in Figure 3.1, is taken out as the slave subsystem while the rest of the test circuit being the master. Therefore, the slave subsystem using the FMI-based co-simulation approach is given in Figure 3.2.

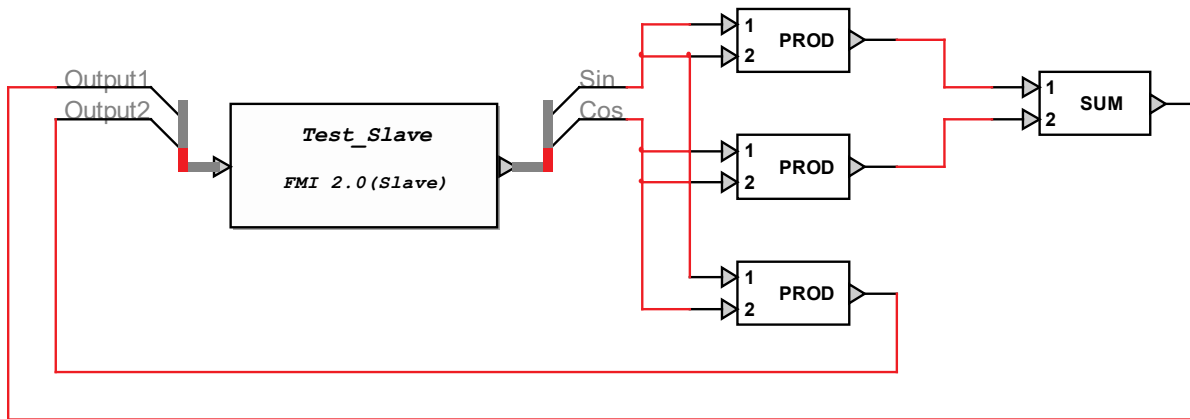


Figure 3.2 Slave subsystem for the validation of co-simulation interface design

The box named “Test_Slave FMI 2.0 (Slave)” indicates the circuit in the current design file is the slave subsystem, and this box serves as the interface with the master. It has 2 inputs (Sin and Cos) and 2 outputs (Output1, Output2). The directions of arrows at the box suggest that the inputs Sin and Cos are the inputs “from” the master, and the outputs Output1 and Output2 are the outputs of the slave “to” the master. The inputs and outputs, co-simulation mode, as well as other parameter settings in order to generate the slave FMU file from the design file shown in Figure 3.2 are presented in Figure 3.3 and Figure 3.4. The slave FMU can thus be generated after effectuating these parameter settings accordingly.

FMI Slave - Generate FMU

FMU Main Data	
Model Name	Test_Slave
Number of FMU Inputs	2
Number of FMU Outputs	2
FMI Version	2.0

FMU Inputs (Device outputs)	
Name	Description
Sin	Description...
Cos	Description...

FMU Outputs (Device inputs)	
Name	Description
Output1	Description...
Output2	Description...

Figure 3.3 Inputs and outputs settings in the slave

FMU Parameters

Number of Parameters	4
----------------------	---

FMU Parameters (see documentation)		
Name	Description	Start Value
emptopt.exe location	Location of the emptopt.exe file	C:\Program Files (x86)\EMTPWorks 3.4\EMTP\emptopt.exe
emptstate.ini location	Location of the emptstate.ini file	C:\Users\Wing Cai\AppData\Roaming\EMTP\IC_Program_Files_x86_EMTPWorks_3_4\emptstate.ini
Co-simulation mode	List of available Co-simulation modes	1
Time Out (ms)	Maximal time for synchronisation of the co-sim	5000

Figure 3.4 Co-simulation mode and other parameter settings in the slave

The master subsystem, as is shown in Figure 3.5, has a box named “Test_Slave FMI Master 2.0” in the place of the circuit elements already enclosed in the slave. This box, similar to that in the slave subsystem, represents the interface with the slave. The master and slave subsystems thus communicate with one another through this interface while not knowing the topology and composition of each other.

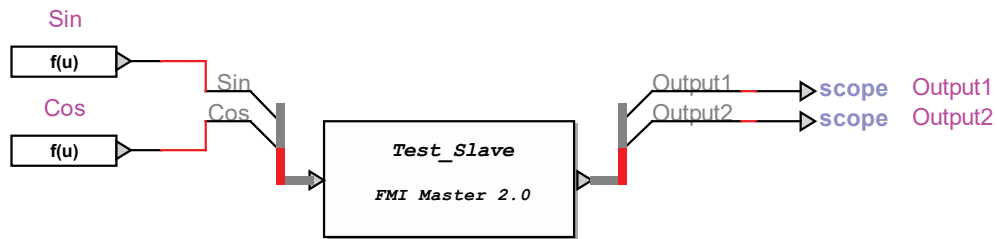


Figure 3.5 Master subsystem for the validation of co-simulation interface design

After loading the generated slave FMU file into the interface represented by the box in Figure 3.5, the master obtains all necessary data and information from the slave in order to execute co-simulation. All loaded data and parameters of the slave in the master are presented in Figure 3.6.

FMU Main Data	
FMU File Name	C:\Users\Ming Cai\Desktop\Test_Slave_sync_100.FMU
Model Name	Test_Slave
Generation Tool	EMTP-RV
Generation Date and Time	2019-3-5 15:2:31
FMI Version	2.0
Number of Inputs	2
Number of Outputs	2
Number of Parameters	4

FMU Simulation Options (Slave)	
Start Time	0 s
Stop Time	0.1 s
Step size	0.1 ms

(a) FMU main data

FMU Inputs

Number of Inputs	2
------------------	---

Name	Description	Start Value
Sin	Description...(Real)	0.0
Cos	Description...(Real)	0.0

(b) FMU inputs

FMU Outputs

Number of Outputs	<input type="text" value="2"/>
-------------------	--------------------------------

Name	Description
Output1	Description...(Real)
Output2	Description...(Real)

(c) FMU outputs

FMU Parameters

Number of Tunable Variables	<input type="text" value="4"/>
Number of Non-Tunable Variables	<input type="text" value="0"/>

Tunable Variables			
Name	Description	Causality / Variability	Start Value
emptopt.exe location	Location of the emptopt.exe file (type = String, start = C:\Program Files (x86)\EMTPWork	parameter / tunable	C:\Program Files (x86)\EMTPWork
emptstate.ini location	Location of the emptstate.ini file (type = String, start = C:\Users\chair\AppData\Roaming	parameter / tunable	C:\Users\Wing Cai\AppData\Roaming
Co-simulation mode	List of available Co-simulation modes (type = Enumeration, start = 1)	parameter / tunable	Mode1 (Synchronous mode)
Time Out (ms)	Maximal time for synchronisation of the co-simulation (type = Integer, min = 2000, start	parameter / tunable	5000

(d) FMU parameters

FMU Advanced Options

FMU Co-Simulation Capabilities (For Information Purpose)	
Option	Value
needsExecutionTool	false
canHandleVariableCommunicationStepSize	false
canInterpolateInputs	false
maxOutputDerivativeOrder	0
canRunAsynchronously	false
canBeInstantiatedOnlyOncePerProcess	false
canNotUseMemoryManagementFunctions	false
canGetAndSetFMUstate	false
canSerializeFMUstate	false
providesDirectionalDerivative	false

(e) FMU advanced options

Figure 3.6 Loaded data and parameters from the slave FMU in the master

3.1.2 Simulation results

The tests performed in this section use 3 simulation scenarios implementing the FMI-based approach in which different co-simulation modes as well as different numerical integration time-steps are utilized in the slave subsystems. This is shown in Table 3.1 where the Original Case is the original single-core test case in EMTP without co-simulation, as is shown in Figure 3.1. It is

noted that although the master and slave concepts do not apply to the Original Case, the numerical integration time-step of $100\ \mu\text{s}$ used in the master and slave in this case indicates that $100\ \mu\text{s}$ is the numerical integration time-step of the entire test case. The simulation interval is 100 ms for all cases. The waveforms of Output 1, Output 2 and closer observation of Output 2 within a short interval in all simulation scenarios are presented in Figure 3.7, Figure 3.8 and Figure 3.9 respectively. For clear observation, only partial waveforms of the entire simulation interval are shown.

Table 3.1 Co-simulation mode scenarios, co-simulation interface design validation

	Co-simulation mode	Master $\Delta t\ (\mu\text{s})$	Slave $\Delta t\ (\mu\text{s})$
Original Case	N/A	100	100
Scenario 1	Synchronous	100	100
Scenario 2	Synchronous	100	200
Scenario 3	Asynchronous	100	100

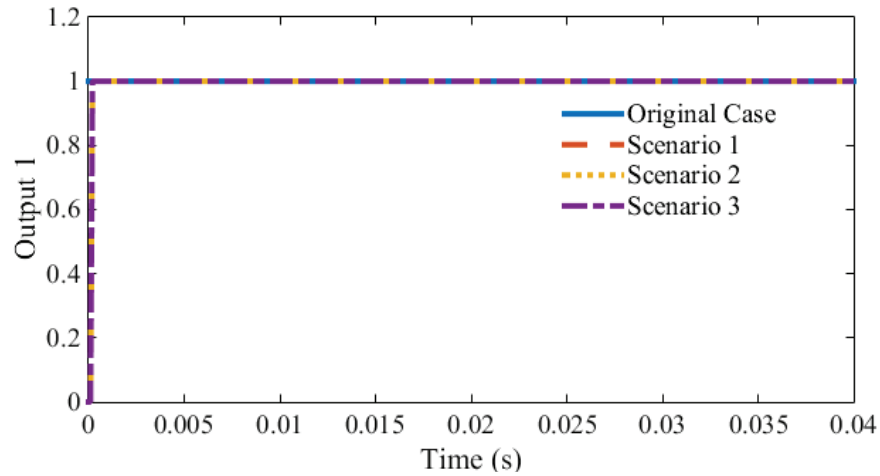


Figure 3.7 Waveforms of Output 1 for all simulation scenarios

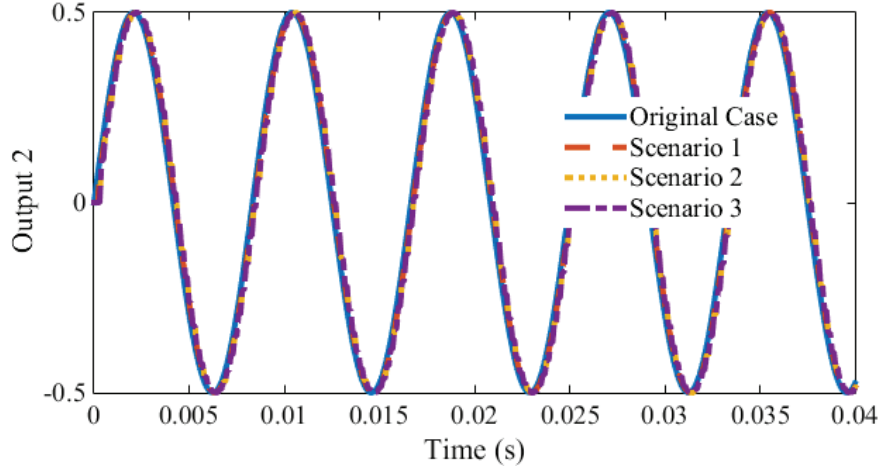


Figure 3.8 Waveforms of Output 2 for all simulation scenarios

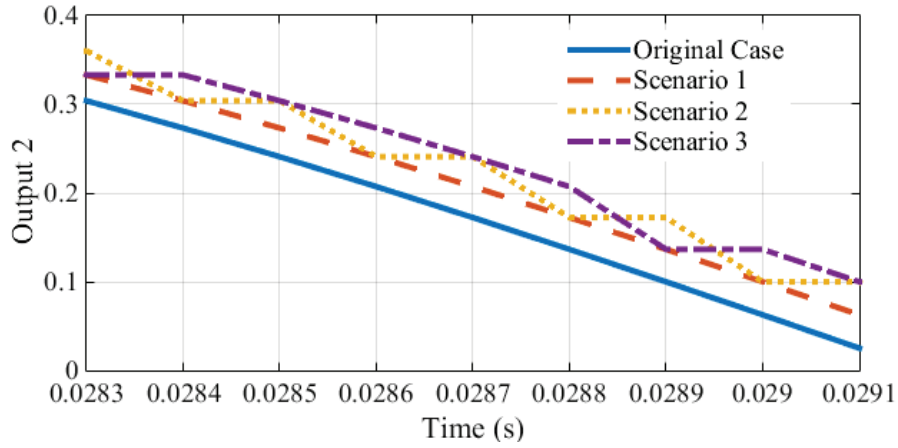


Figure 3.9 Closer observation of waveform of Output 2 for all simulation scenarios

It is observed in Figure 3.7 and Figure 3.8 that the waveforms of Output 1 and Output 2 in all simulation scenarios using the FMI-based co-simulation approach match well to those in the Original Case without co-simulation. Nonetheless, upon closer observation of the waveform of Output 2 within a short interval in all simulation scenarios, as presented in Figure 3.9, a delay of one or two time-steps exists between the slave and the master in all scenarios using the FMI-based co-simulation approach compared to the Original Case. It is worth noting that a one time-step delay is manually added in the function `stepFunc` in the DLL file `FMI2_Device_Master` during the master and slave data exchange for simulation stability and robustness. Therefore, the slave subsystem in Scenario 1 (synchronous mode with a single time-step) exhibits a constant one time-step delay of 100 ms compared to the Original Case. The delays observed in the other scenarios are discussed in detail in Section 3.2.1.3.

By and large, the simple test performed in this section has demonstrated that the co-simulation interface in EMTP using the FMI standard developed in this thesis is functional. Further validation on different aspects of the implementation can thus be conducted.

3.2 Validation of the original methodology

In this section, tests validating the original methodology, which is the two co-simulation modes with master-slave synchronization schemes proposed in Section 2.2, are performed on practical power system simulation benchmarks with complex control systems (wind generator control and protection relays). The tests are performed focusing on two aspects: accuracy and computation efficiency. Detailed error analysis and discussions on the observed computation time gains are also provided.

3.2.1 Accuracy validation

Accuracy of the original methodology is verified by comparing various variables in a simulation, such as offshore wind park outputs as well as line relay tripping instants after a perturbation (fault). The reference waveforms are those found from the simulation using original EMTP single-core benchmarks without co-simulation. In all cases, a multiphase unbalanced load-flow solution is performed for initializing the time-domain solution where a fault condition is simulated.

3.2.1.1 Test case Network-1

The tests performed on Network-1 are to validate the accuracy of co-simulation in offshore wind park control systems by observing the outputs of offshore wind parks through HVDC transmission. 4 simulation scenarios implementing the FMI-based co-simulation approach are used, in which different co-simulation modes as well as different numerical integration time-steps are utilized in the offshore wind generator control systems. This is shown in Table 3.2 where the Original Case is the original single-core benchmark simulated on EMTP without co-simulation.

Table 3.2 Co-simulation mode scenarios, accuracy validation of the original methodology, Network-1

	Co-simulation mode	Main network Δt (μs)	WP control Δt (μs)
Original Case	N/A	50	50
Scenario 1	Asynchronous	50	50
Scenario 2	Synchronous	50	50
Scenario 3	Synchronous	50	100
Scenario 4	Synchronous	50	200

The Network-1 benchmark is shown in Figure 3.10. It is the EMT version of the original IEEE-39 benchmark with onshore and offshore wind generation as well as HVDC transmission [88]. This benchmark represents a portion of the 345-kV New England transmission grid [89], [90] consisting of 4 voltage levels, 39 buses, 8 synchronous generators (the other 2 at buses B25 and B2 are replaced by onshore and offshore wind parks), 34 transmission lines, 12 transformers, and 19 loads. It modifies the original data [89], [90] by representing transmission lines using the CP model and incorporating both static and dynamic load types, together with supplementary data in tower geometry, conductor types, transformers, and machine controls. Further device modelling details can be found in [1], [88] and [91]. This benchmark has a total of 1381 network nodes, with the size of the main network equations (modified-augmented-nodal-analysis [2]) being 4826×4826 with 35718 non-zeros.

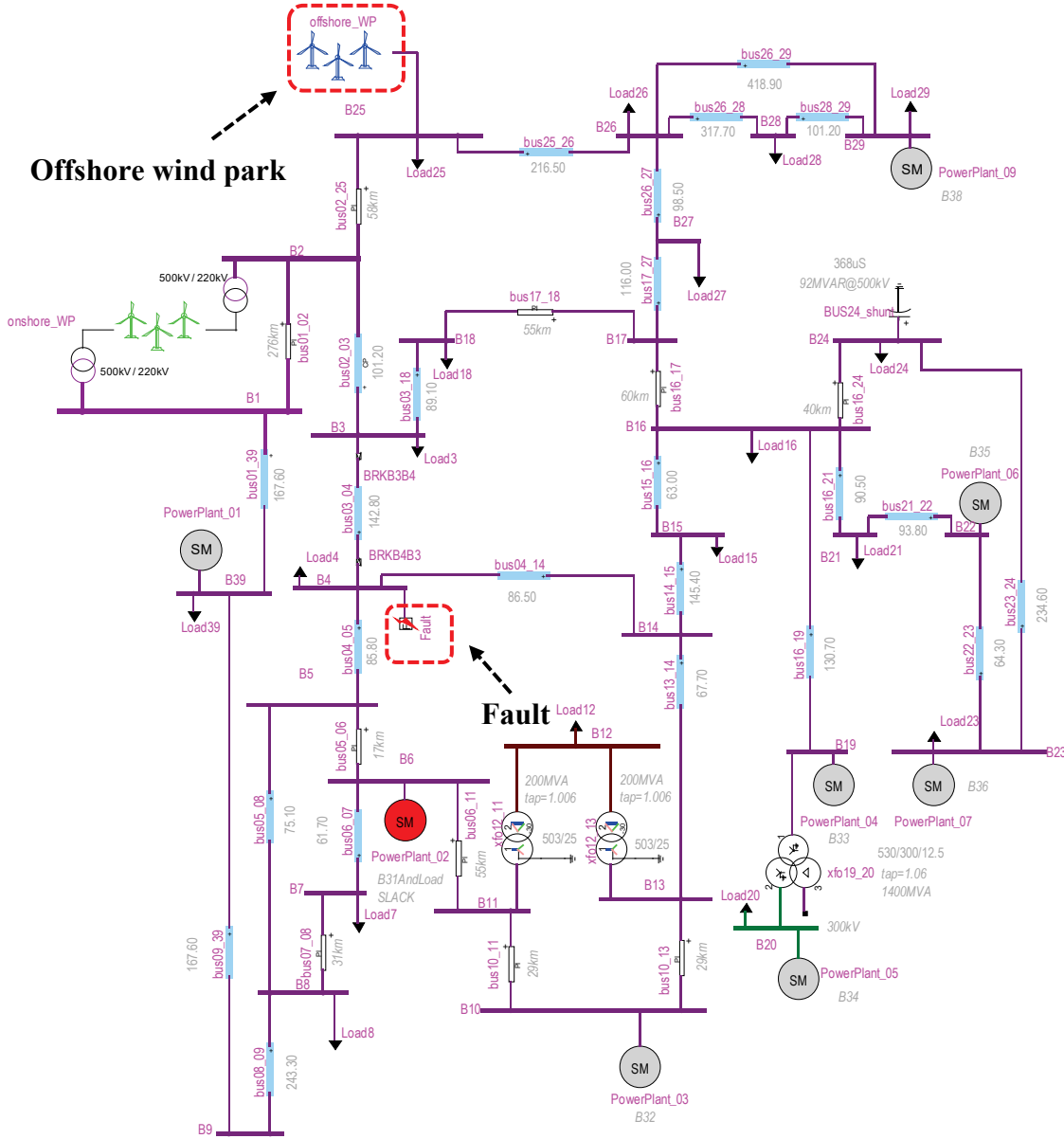


Figure 3.10 Network-1 benchmark

It is noted that the offshore wind park consists of 4 separate aggregated wind parks, shown in Figure 3.11, using the average value model [1] for wind generators. The design of one offshore wind park is presented in Figure 3.12 as an example. In this test, the 4 offshore wind park control systems (see Figure 3.13) are implemented using the FMI-based co-simulation approach in the 4 scenarios shown in Table 3.2. The control systems, whose detailed are presented in Figure 3.14, are realistic and include all necessary blocks (in total 1656 blocks for each aggregated wind park control system) for various control modes and protection systems [2].

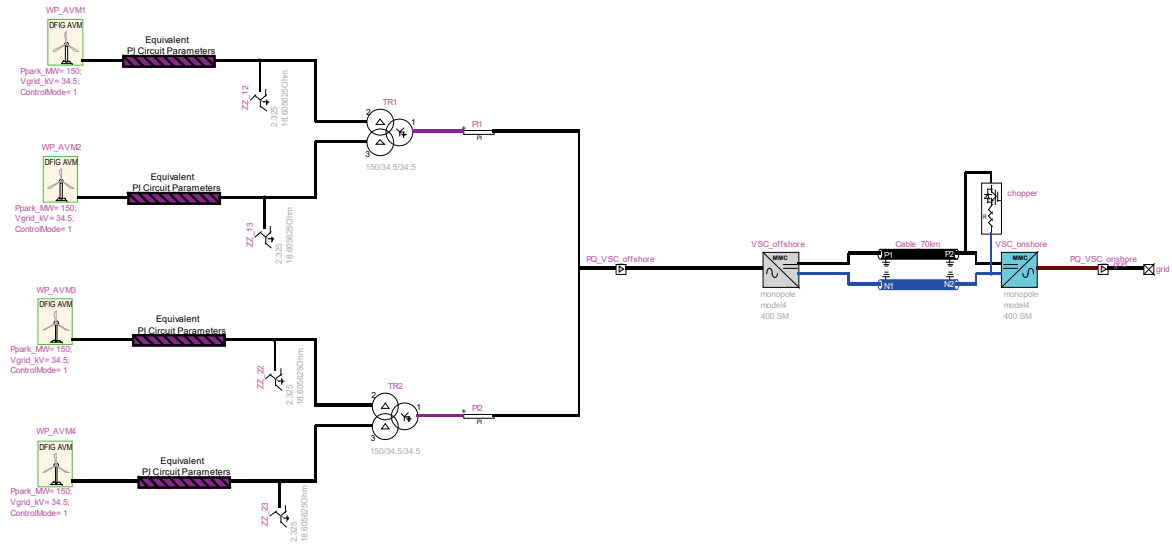


Figure 3.11 Offshore wind parks and HVDC connection in Network-1

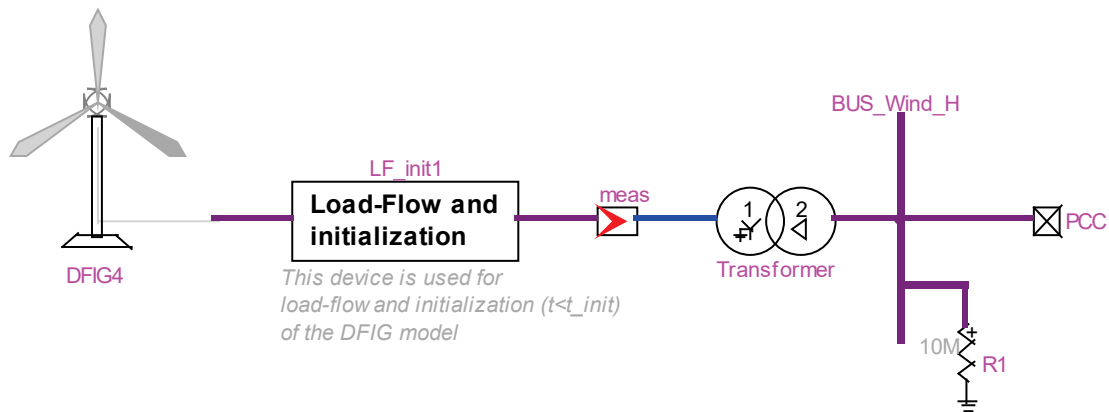


Figure 3.12 Design of one offshore wind park in Network-1

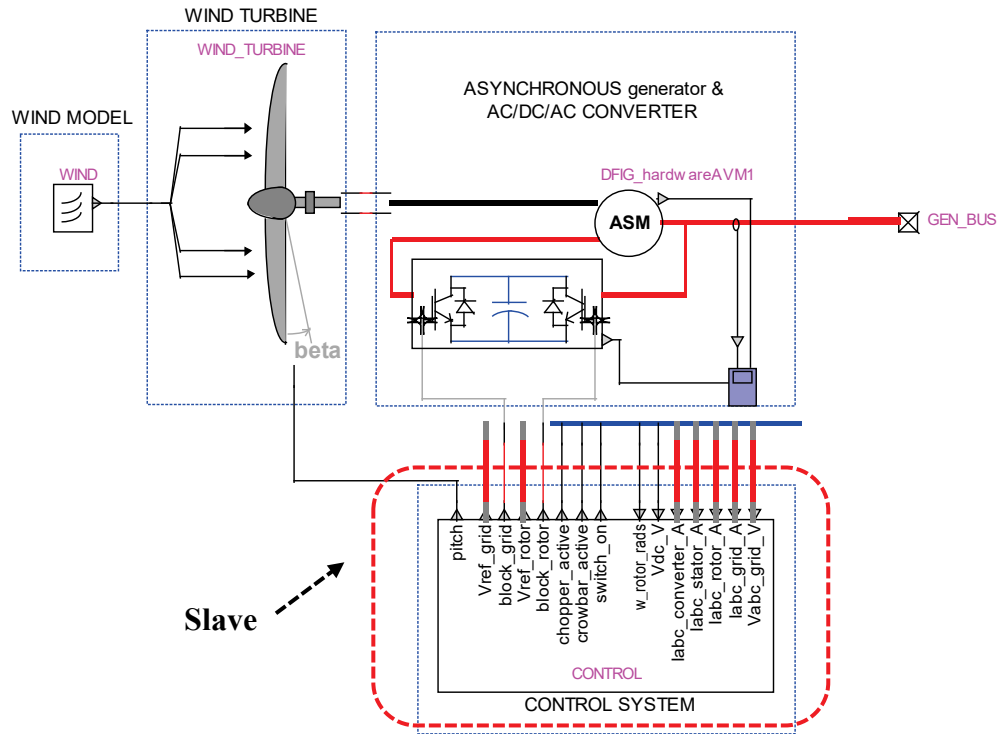


Figure 3.13 DFIG model in the offshore wind parks in Network-1

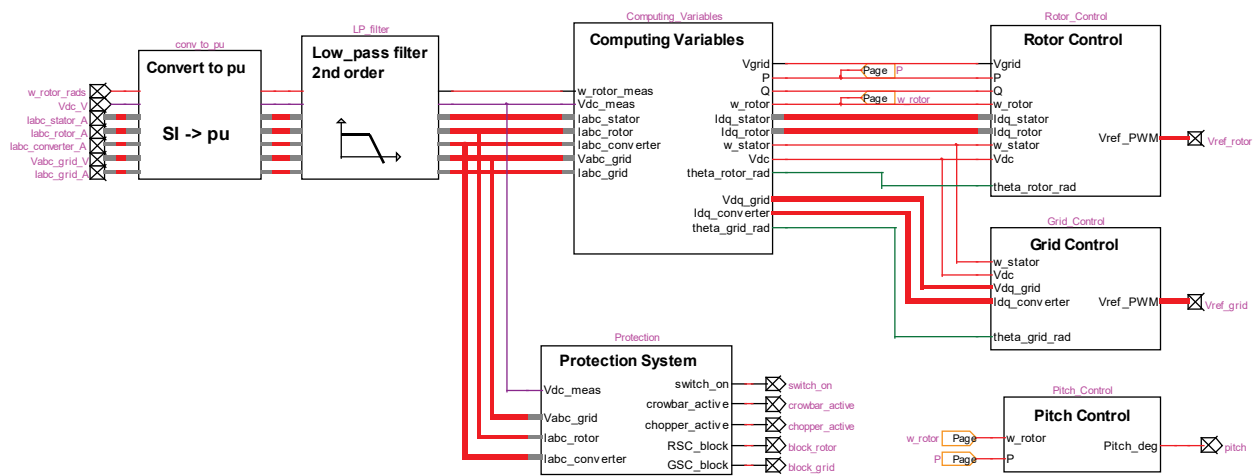
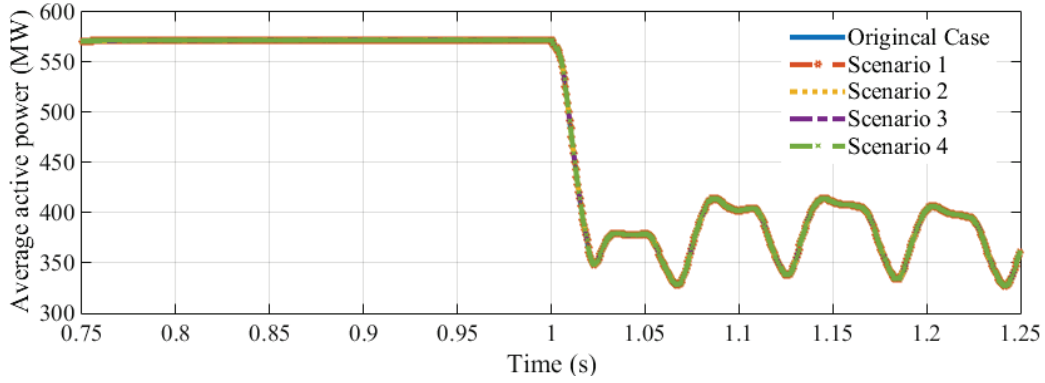
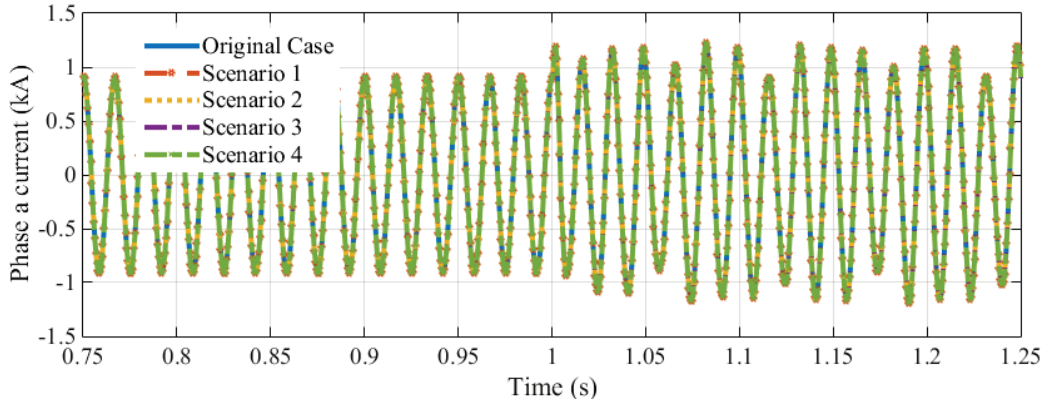


Figure 3.14 Control system design for the DFIG model in the offshore wind parks

The simulation interval is 3 s for all scenarios and the Original Case. The three-phase permanent fault location is indicated in Figure 3.10 and occurs at $t = 1\text{s}$. The waveforms of average active power and phase-a current flowing out of the offshore AC cable (of HVDC converter) into the grid are presented in Figure 3.15. For better observation, only the interval between $t = 0.75\text{s}$ and $t = 1.25\text{s}$ is presented for both waveforms.



(a) Average active power



(b) Phase-a current

Figure 3.15 Waveforms of average active power and phase-a current

Upon closer inspection, it is observed that the maximum (all solution time-points) per-unit error in all simulation scenarios in comparison to the Original Case is less than 0.01% , thus demonstrating that the implementation of the FMI-based new simulation approach on the offshore wind park controls remain accurate. An error analysis is provided in a later section.

3.2.1.2 Test case Network-2

This test case validates the accuracy of the FMI-based co-simulation approach in relay applications by examining line relay tripping instants after a fault. It studies relay tripping events at adjacent lines when a phase-to-ground fault occurs at a line whose own relays are defective hence could not trip for fault clearance. The Network-2 benchmark is used in this test case (see Figure 3.18). It is a realistic power system simulation benchmark based on the data of [92], [93]. This 459-bus benchmark consists of 7 voltage levels (transmission, sub-transmission, distribution and generation), 637 transmission lines represented by both CP models and PI sections, and 10 type-III onshore wind generators connected at sub-transmission level. Parameter and modelling details can be found in [88]. The total number of network nodes in this benchmark is 3407, with the size of main network equations being 12120×12120 (125120 non-zeros). Two line relays are added to every CP line in the benchmark, represented by the rectangular boxes (Figure 3.18) whose contents are shown in Figure 3.16. Settings of Relay 1 related to this particular study are presented in Figure 3.17 for demonstration purposes. It is noted that each line relay contains 4893 control devices [94], and the entire relay is decoupled from the electrical network as the slave subsystem in co-simulation.

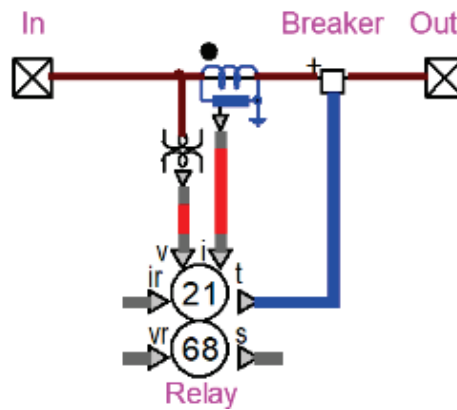


Figure 3.16 Line relay at each end of a CP line

Frequency Hz

Phase CT Connection

Ratio I_{Rated} A RMS secondary

Ground CT

Ratio

Phase VT Connection

Ratio V_{Rated} V RMSLG secondary

Ground VT

Ratio

(a) Electrical characteristics settings

Manufacturer

Line impedances

Positive-sequence magnitude	<input type="text" value="14.71"/>	<input type="text" value="Ω primary"/>	angle	<input type="text" value="86"/>	<input type="text" value="deg"/>
Zero-sequence magnitude	<input type="text" value="56.4"/>	<input type="text" value="Ω primary"/>	angle	<input type="text" value="75"/>	<input type="text" value="deg"/>

Voltage memory polarization type

☐ Use Positive-sequence

(b) Line characteristics settings

Manufacturer Generic ▼

☒ Enable ground distance relay

☒ Phase-to-ground zone 1

Direction Forward ▼

Shape Mho/Lens ▼

Reach 80 % ▼

RCA 85 deg

Comparator limit angle 90 deg

Delay 0 s

Z0/Z1 mag 3.8

Z0/Z1 angle -10 deg

Z0M/Z1 mag 0

Z0M/Z1 angle 0 deg

i_{0min} 0.05 pu

☐ Enable the reactance characteristic supervision

☐ Enable the negative-sequence polarized directional characteristic supervision

☐ Enable the neutral polarized directional characteristic supervision

(c) Phase-to-ground distance settings, Zone 1

Figure 3.17 Relay 1 settings (partial)

Relay 1 Relay 2 Relay 3

A permanent phase-a-to-ground fault occurs on the faulty line (Figure 3.18) at 36.6 km from bus IZMIR. Based on the studied fault type, location, and relay settings, only the relays numbered from 1 to 5, as indicated in Figure 3.18, could trip after relay failure at the faulty line. Table 3.3 presents the tripping zones in which the aforementioned relays could trip as well as the principal relay settings for each zone. Since Relay 2 and 3, 4 and 5 are placed at opposite ends of the same lines, only three relays could trip in this test case, which are Relay 1 (Zone 4), Relay 2 (Zone 3) and Relay 4 (Zone 3).

Table 3.3 Relay 1 to 5 tripping zones

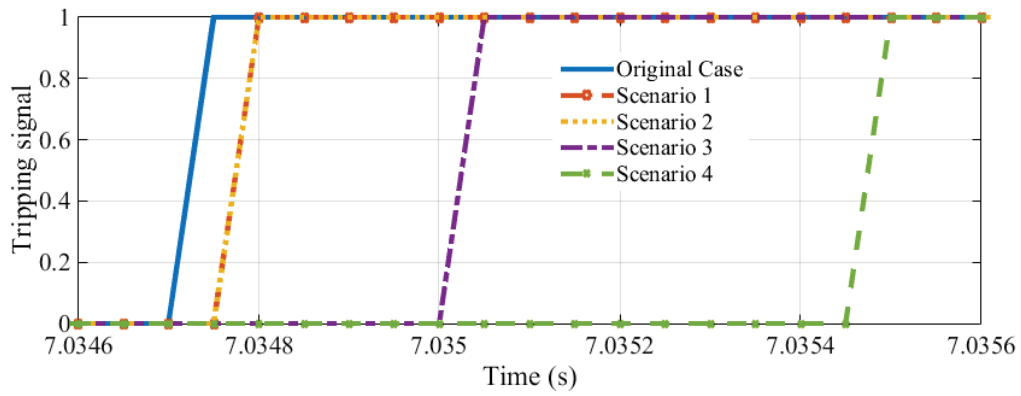
	Zone 1	Zone 2	Zone 3	Zone 4
Direction	Forward	Forward	Reverse	Non-directional
Reach (%)	80	120	80	160
Delay (s)	0	0.5	2	5
Relay 1	✗	✗	✗	✓
Relays 2, 4	✗	✗	✓	✓
Relays 3, 5	✗	✗	✗	✓

This test also consists of 4 co-simulation scenarios, as is shown in Table 3.4, with different numerical integration time-steps employed in the line relays. The simulation interval is 10 s for all scenarios as in the Original Case, and the phase-a-to-ground fault occurs at $t = 2s$.

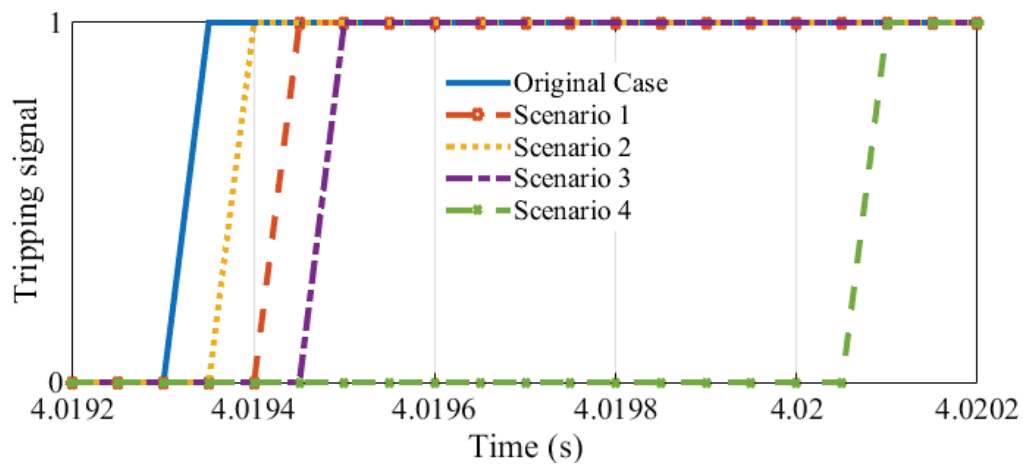
Table 3.4 Co-simulation mode scenarios, accuracy validation of the original methodology, Network-2

	Co-simulation mode	Main network Δt (μs)	Relay Δt (μs)
Original Case	N/A	50	50
Scenario 1	Asynchronous	50	50
Scenario 2	Synchronous	50	50
Scenario 3	Synchronous	50	200
Scenario 4	Synchronous	50	400

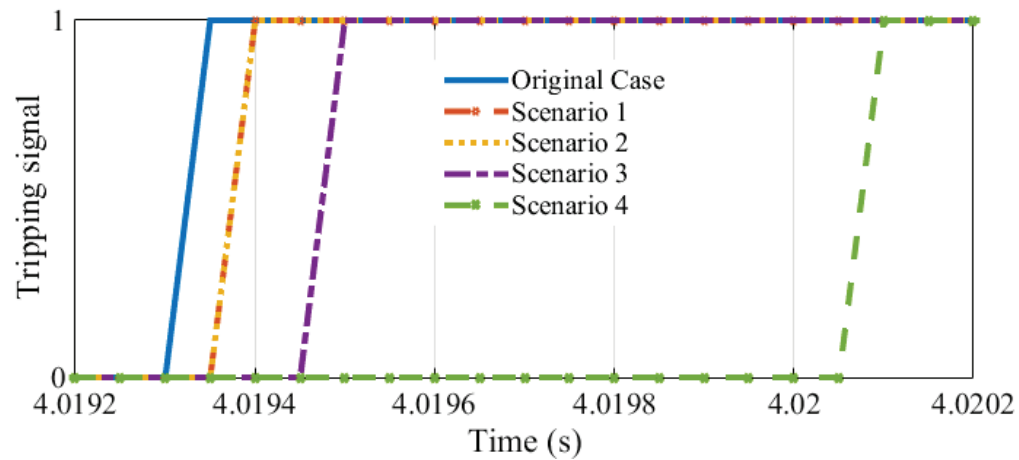
The tripping instants of Relays 1, 2 and 4 are shown in Figure 3.19. Additionally, the phase a locus of Relay 1 in this test case is shown in Figure 3.20, in which it is observed that the locus entering Zone 4 triggers the tripping signal.



(a) Relay 1 tripping in Zone 4



(b) Relay 2 tripping in Zone 3



(c) Relay 4 tripping in Zone 3

Figure 3.19 Tripping instants of Relays 1, 2 and 4

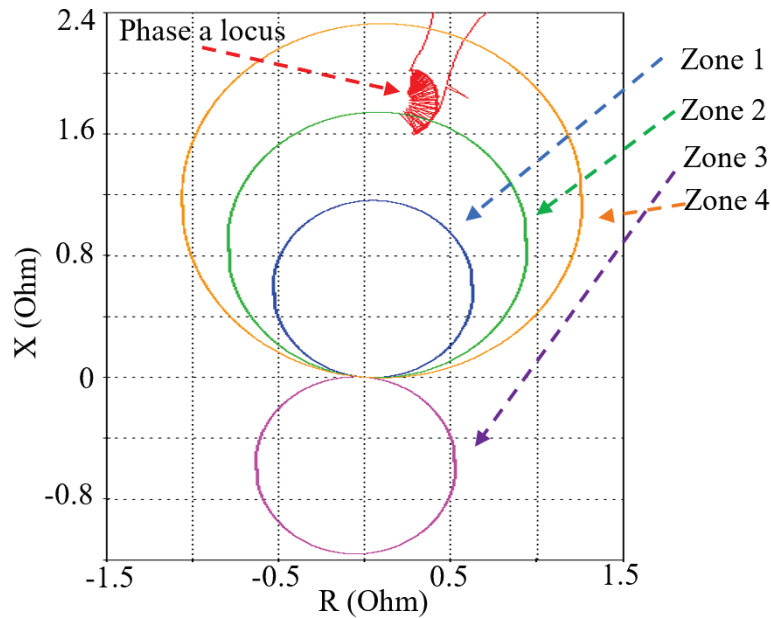


Figure 3.20 Phase a locus of Relay 1

The largest relay tripping instant delay observed in this test case is $750 \mu\text{s}$ (Scenario 4) compared to the Original Case, which is merely $1/22$ cycle approximately. Catering to our need for simulation speedup and considering that the total clearing time (relay plus circuit breaker) typically ranges from 2 to 8 cycles on a 60 Hz base [95], this error is acceptably small. A detailed error analysis is given in the next section.

It is necessary to mention that the purpose of this test case is to validate the accuracy of the original methodology in relay applications. It focuses on the comparison between results obtained from the original approach on EMTP without co-simulation and those from the FMI-based co-simulation approach. The relay settings in practical protection studies might differ from those used in this test case.

3.2.1.3 Error analysis

It is noted that due to the absence of iterations between master and slaves, a one time-step delay is manually added in the function `stepFunc` inside DLL file `FMI2_Device_Master` during master-slave data exchange for simulation stability and robustness. This is the only error source in Scenario 2 in both test cases, and also contributes partially to errors observed in all other scenarios in both test cases. All scenarios mentioned in the following discussion refer to those in Table 3.2 and Table 3.4.

1) Test case Network-1

The aforementioned less than 0.01% per unit error can be explained by examining the co-simulation process of Scenarios 1 (asynchronous mode with a single numerical integration time-step) and 3 (synchronous mode with different numerical integration time-steps) since the errors from Scenario 4 can be explained accordingly. The simplified co-simulation processes of Scenarios 1 and 3 are presented in Figure 3.21 and Figure 3.22 respectively, in which the hollow arrows indicate the time evolution of co-simulation, the dotted black arrows denote the release of semaphores, and P_T as well as C_T represent the outputs from master and slave at $t = T$. The extra one time-step delay added in the communication process is not reflected in the figures.

In Figure 3.21, as the co-simulation proceeds, after the mutual release of semaphores, the master and slave receive data from each other and effectuate their computations simultaneously. Due to the nature of execution parallelism, the master does not verify whether the slave has finished writing onto the co-simulation bus before reading from it. Hence, instead of reading the slave output for the current time-point (pointed by the green dotted arrows) as it might not be immediately available, the master might read the one from the previous time-point (pointed by the red dotted arrow), which would lead to an additional one time-step delay.

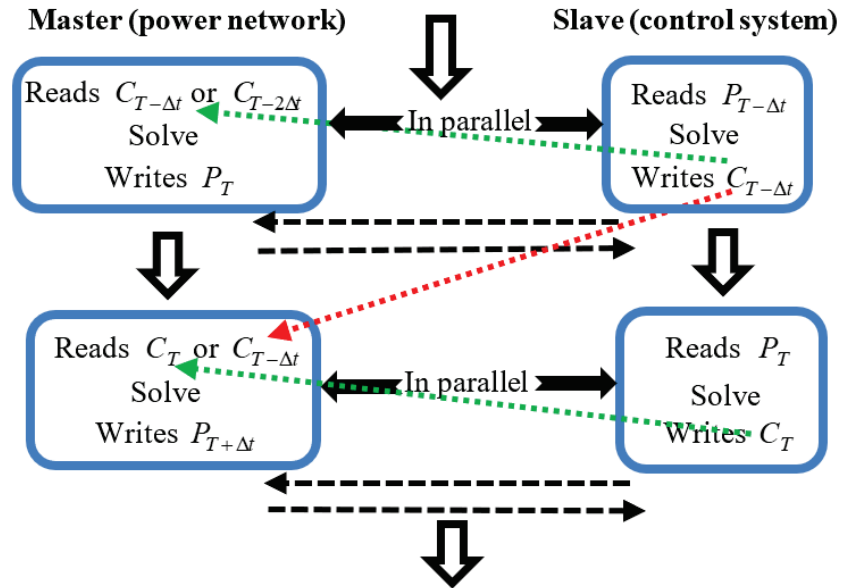


Figure 3.21 Simplified co-simulation process of Scenario 1

Figure 3.22 presents the synchronous mode with different time-steps in which the slave time-step is twice that of the master. After advancing by one step and releasing semaphore to the master, the

slave, having a larger time-step, remains idle, waiting for the master to catch up. Meanwhile, the master reads input from the slave (green dotted arrow), solves for the current time-point and advances by one step. Nonetheless, in its subsequent computations, the master keeps reading the same data from the slave (red dotted arrow) as the latter remains idle (slave output stored in the co-simulation bus has not yet been updated). An additional delay of one time-step is thus introduced between the master and slave. Apparently, if the slave time-step is much larger than that of the master (as in Scenario 4), more delays will be introduced.

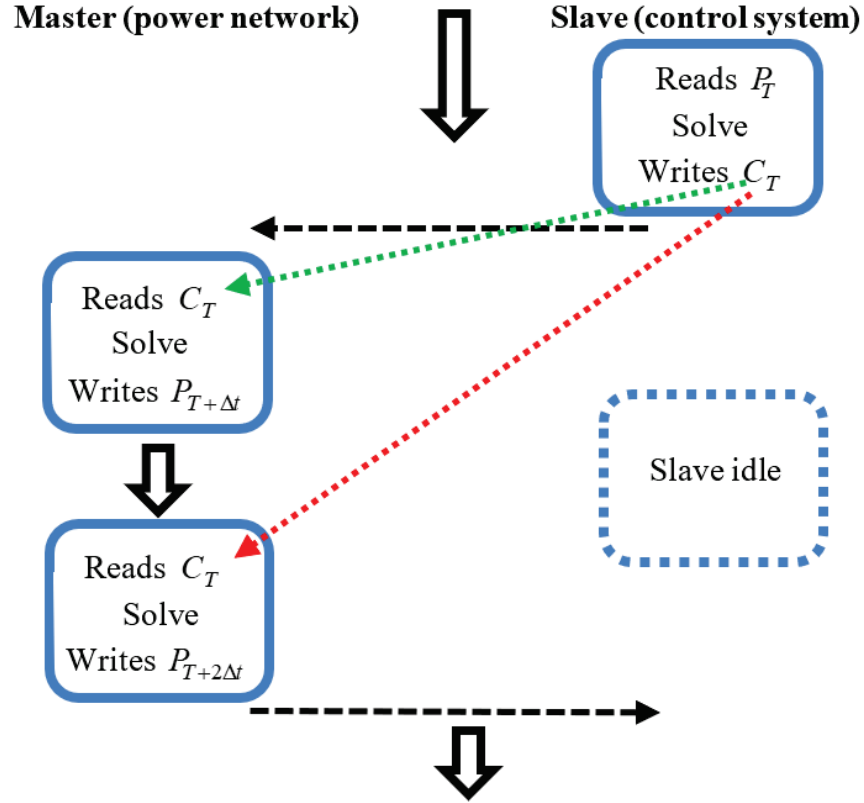


Figure 3.22 Simplified co-simulation process of Scenario 3

It is, therefore, understood that the additional delays in the asynchronous mode with the same time-step (Scenario 1) and the synchronous mode with different time-steps (Scenarios 3 and 4) together with the extra one time-step delay manually added in the master-slave data exchange process, is the source of the observed small error.

2) Test case Network-2

In this test case, the delays discussed in the previous section contribute completely to errors observed in Scenario 1 and partially to those observed in Scenarios 3 and 4.

Another source of errors in Scenarios 3 and 4 (synchronous mode with different numerical integration time-steps) is due to the interpolation process in relay sampling. The relay models used in the work of this thesis require that the interval between sampling instants should be larger than twice the relay numerical integration time-step [95]. For this reason, in Scenario 4 where the relay time-step is $400\ \mu\text{s}$, a sampling rate of 20 samples/cycle is chosen, which renders an interval of $833.33\ \mu\text{s}$ between sampling instants. Such an interval indicates that the relay sampling instants are not on time mesh since the relay time-step is $400\ \mu\text{s}$ and that of the main network is $50\ \mu\text{s}$. An interpolation is thus performed at each sampling instant using data from the two adjacent relay simulation time-points between which the sampling instant occurs [95]. As the interpolation for relay sampling will not be performed until the data from a later simulation time-point becomes available, a delay is, therefore, introduced, which also constitutes the observed errors in relay tripping signals in Scenario 4. Errors in Scenario 3 can be explained in a similar fashion.

The largest errors are observed in Scenario 4 where a numerical integration time-step of $400\ \mu\text{s}$ is adopted in the relays. This is only for demonstrating application capacity of the proposed approach. In practice, a smaller relay time-step can be used for higher simulation accuracy while obtaining similar speedup, as will be seen in the following section.

3.2.2 Computation time gains

The validation of the original methodology in the aspect of computation time gains is based on the comparison of the solution time of the benchmarks simulated on EMTP without co-simulation and with the FMI-based co-simulation approach. Its objective is to demonstrate the numerical performance advantages of the FMI-based new co-simulation approach on practical power system networks of different levels of complexity, Network-1 and -2, as are shown in Figure 3.10 and Figure 3.18 respectively. Two PCs with 2 and 16 cores (4 and 32 logical processors respectively), whose hardware and software configurations are given in Table 3.5, are used in the tests of computation time gains. Again, in all cases, a multiphase unbalanced load-flow solution is performed for initializing the time-domain solution where a fault condition is simulated. A simulation interval of 2 s is used in all tests.

Table 3.5 Hardware and software configurations of the PCs

	PC1	PC2
Processor	Intel (R) Core (TM) i5-5200U	Intel (R) Xeon (R) E5-2667 v3
Number of cores	2	16
Number of logical processors	4	32
Installed memory (RAM)	12.0 GB	32.0 GB
Base speed	2.8 GHz	3.2 GHz
Operating system	Microsoft Windows 10 Pro	Microsoft Windows 10 Pro

3.2.2.1 Test case Network-1

The first batch of tests are performed on the Network-1 benchmark shown in Figure 3.10. Different numbers of line relays, presented in Figure 3.16, are also added into this benchmark (the number of relays used in each test is found in Table 3.7 and Table 3.8). As in the previous sections, the tests performed in this section include the Original Case without any co-simulation and 3 co-simulation scenarios, as is shown in Table 3.6. All 4 offshore aggregated wind park control systems and line relays are implemented using the FMI-based co-simulation approach.

Table 3.6 Co-simulation mode scenarios, computation time gains

	Co-simulation mode	Main network Δt (μs)	WP control Δt (μs)	Relay Δt (μs)
Original Case	N/A	50	50	50
Scenario 1	Asynchronous	50	50	50
Scenario 2	Synchronous	50	200	200
Scenario 3	Synchronous	50	200	400

The total solution times on PCs with 2 and 16 cores (4 and 32 logical processors respectively) for Network-1 for all simulation scenarios compared to the Original Case are presented in Table 3.7 and Table 3.8. The slave subsystems in all simulation scenarios in this test case consist of both offshore wind park control systems and line relays since the latter two are implemented using co-simulation. Based on Table 3.7 and Table 3.8, the speedup of total solution time in all simulation scenarios in comparison to the Original Case with respect to the number of slaves on PCs with 2 and 16 cores can thus be calculated, as is illustrated in Figure 3.23 and Figure 3.24.

Table 3.7 Comparison of solution times (s) on a PC with 2 cores (4 logical processors) for different scenarios, Network-1

Number of WP controls	4						
Number of relays	11	16	22	28	34	40	46
Total solution time (s)							
Original Case	1053	1462	2026	2769	3360	4105	4978
Scenario 1	440	518	620	718	824	901	1044
Scenario 2	361	427	505	566	648	711	807
Scenario 3	326	361	427	469	509	560	621

Table 3.8 Comparison of solution times (s) on a PC with 16 cores (32 logical processors) for different scenarios, Network-1

Number of WP controls	4						
Number of relays	11	16	22	28	34	40	46
Total solution time (s)							
Original Case	824	1292	1963	2415	2989	3901	4644
Scenario 1	220	248	299	345	375	424	476
Scenario 2	256	296	346	400	461	519	576
Scenario 3	242	275	313	355	400	444	480

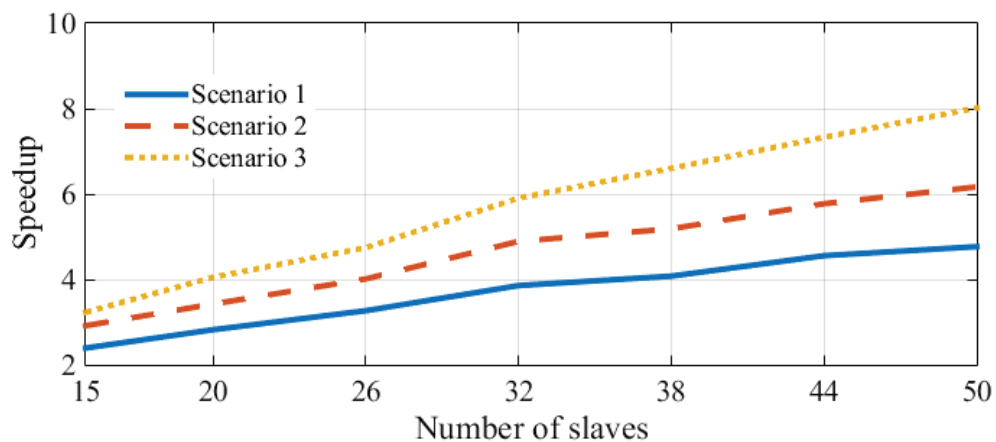


Figure 3.23 Total solution time speedup with respect to the number of slaves in Network-1 on a PC with 2 cores (4 logical processors)

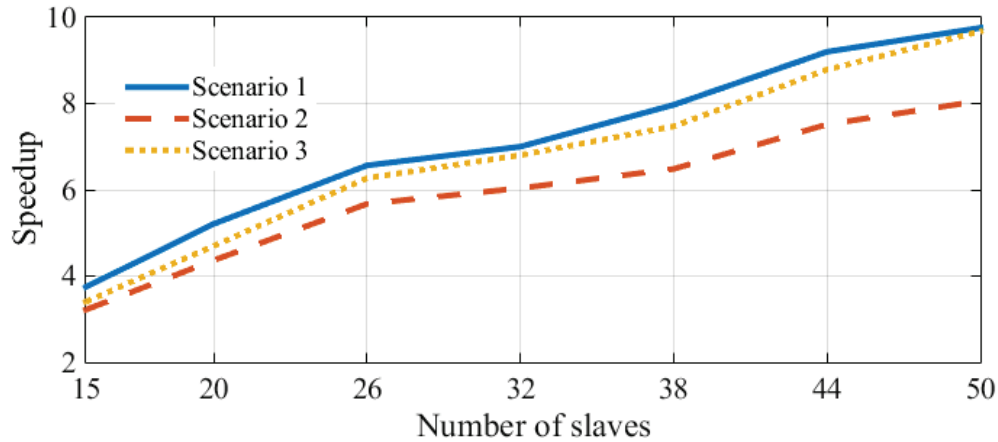


Figure 3.24 Total solution time speedup with respect to the number of slaves in Network-1 on a PC with 16 cores (32 logical processors)

3.2.2.2 Test case Network-2

Network-2 benchmark (see Figure 3.18) is used in the second batch of tests. However, instead of placing two line relays at every CP line as is shown in Figure 3.18, different numbers of line relays are once again used in each test (see Table 3.9 and Table 3.10).

This study also includes the Original Case and 3 co-simulation scenarios detailed in Table 3.6 except that offshore wind parks are not present in this test case. Therefore, the slave subsystems in this test case are only comprised of line relays in all simulation scenarios.

The total solution times on PCs with 2 and 16 cores (4 and 32 logical processors respectively) for Network-2 for different scenarios compared to the Original Case are presented in Table 3.9 and Table 3.10, and the solution time speedup in comparison to the Original Case with respect to the number of slaves (line relays) on PCs with 2 and 16 cores is shown in Figure 3.25 and Figure 3.26.

Table 3.9 Comparison of solution times (s) on a PC with 2 cores (4 logical processors) for different scenarios, Network-2

Number of relays	15	20	40	60	80	100	116
Total solution time (s)							
Original case	2981	3605	6793	11095	16553	23278	29926
Scenario 1	1631	1728	2307	2514	2824	3302	3606
Scenario 2	1581	1656	1943	2269	2543	2859	3142
Scenario 3	1535	1601	1839	2135	2402	2626	2888

Table 3.10 Comparison of solution times (s) on a PC with 16 cores (32 logical processors) for different scenarios, Network-2

Number of relays	15	20	40	60	80	100	116
Total solution time (s)							
Original case	2419	2840	5494	9288	14426	19249	24351
Scenario 1	1237	1287	1471	1778	1939	2184	2563
Scenario 2	1259	1328	1492	1795	2091	2288	2601
Scenario 3	1246	1294	1418	1654	1862	2061	2263

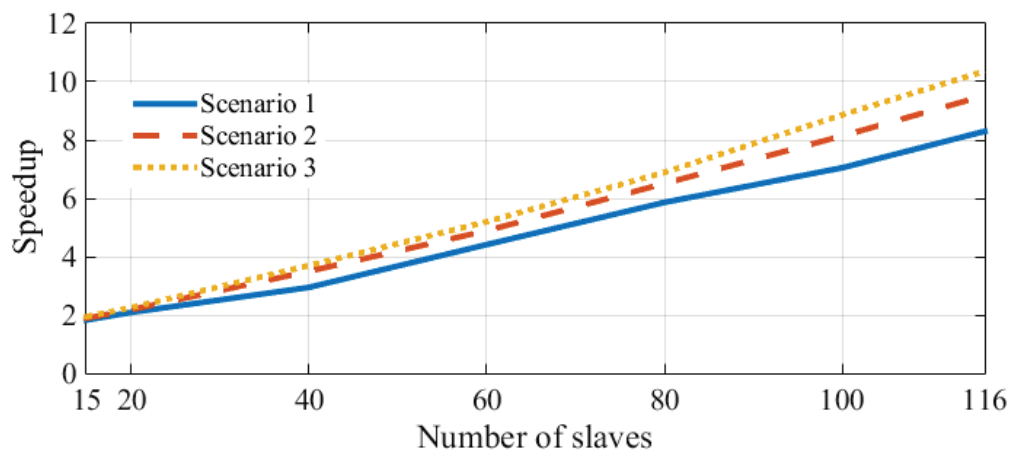


Figure 3.25 Total solution time speedup with respect to the number of slaves in Network-2 on a PC with 2 cores (4 logical processors)

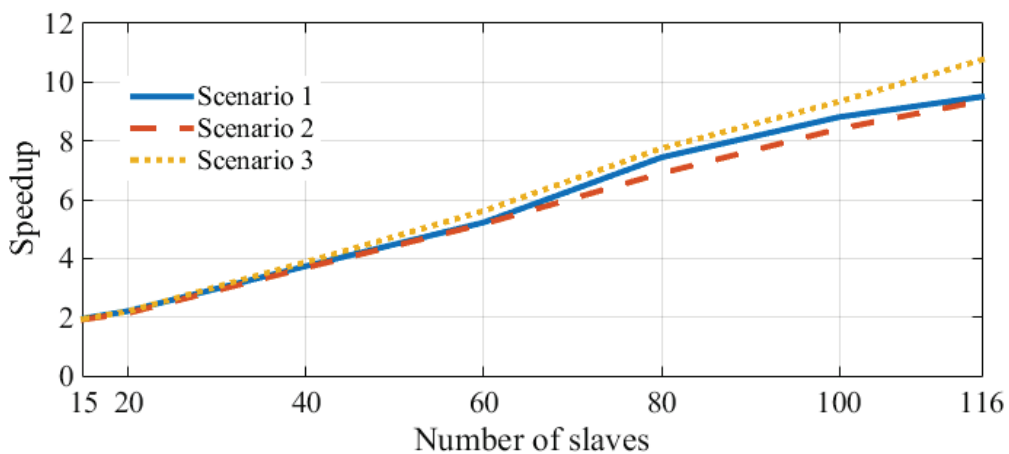


Figure 3.26 Total solution time speedup with respect to the number of slaves in Network-2 on a PC with 16 cores (32 logical processors)

Overall, it can be observed from Figure 3.23 to Figure 3.26 that considerable solution time speedup is obtained in power system transient simulations using the new FMI-based co-simulation approach. Furthermore, several conclusions on the comparison of the two co-simulation modes can also be drawn as follows.

Firstly, since the computation of slave subsystems is distributed among separate logical processors in both synchronous and asynchronous modes, as the number of decoupled slave subsystems increases in a network, the solution time speedup in both modes increases accordingly, with the computing capability of the multi-core CPU sufficiently exploited (see Figure 3.23 to Figure 3.26).

Secondly, in a computation environment where parallelism cannot be adequately fulfilled due to limited available CPU cores, although executed sequentially, the synchronous mode could result in a higher speedup compared to the asynchronous mode performed in parallel thanks to its advantage of being able to adopt different time-steps in the control systems (see Figure 3.23 and Figure 3.25).

Thirdly, the asynchronous mode could be more efficient than the synchronous mode with different time-steps in achieving a higher speedup if the simulation can be parallelized among sufficient logical processors (see Figure 3.24). However, its performance could still be surpassed by the synchronous mode with different numerical integration time-steps in certain cases because of communication overhead brought by data exchange between master and slaves at every time-point in the parallel asynchronous mode (see the comparison between Scenarios 1 and 3 in Figure 3.26).

An in-depth analysis of the observed computation time gains in both co-simulation modes is provided in the following section.

3.2.2.3 Discussion

This section elaborates on the computation time gains achieved using the FMI-based co-simulation approach. The performance of both co-simulation modes (parallel asynchronous and sequential synchronous) are analyzed for both test benchmarks (see Figure 3.23 and Figure 3.26) on two PCs with 2 and 16 cores (4 and 32 logical processors respectively). The scenarios in the following discussion refer to those in Table 3.6.

1) Parallel asynchronous mode (Scenario 1)

First of all, it is worth noting that in the original approach adopted in EMTP, the equations of all control systems are built into a single matrix and solved for every time-step. However, in the co-simulation approach, every single control system is decoupled from the main electrical network and is granted their own memory space. Their computation is performed individually on separate logical processors using different solver instances. Furthermore, the computation time gains are calculated by dividing the solution time using the original approach (EMTP) by the solution time using the co-simulation approach.

For the parallel asynchronous mode with a single numerical integration time-step used in the master and slaves, several observations can be made in Figure 3.23 to Figure 3.26, which are presented as follows together with their associated analyses.

- Observation 1: High gains on a 2-core PC

A computation time gain of approximately 7 is observed on Network-2 with 100 slaves (line relays) on a 2-core PC (with 4 logical processors), as can be seen in Figure 3.25. This can be explained by comparing two different solution schemes adopted in EMTP and the FMI-based co-simulation approach, as shown in Figure 3.27 and Figure 3.28 respectively. It is noted $n \times n$ represents the size of the equation system of one line relay, and $100n \times 100n$ denotes the size of the equation system of 100 line relays, as is the case in EMTP.

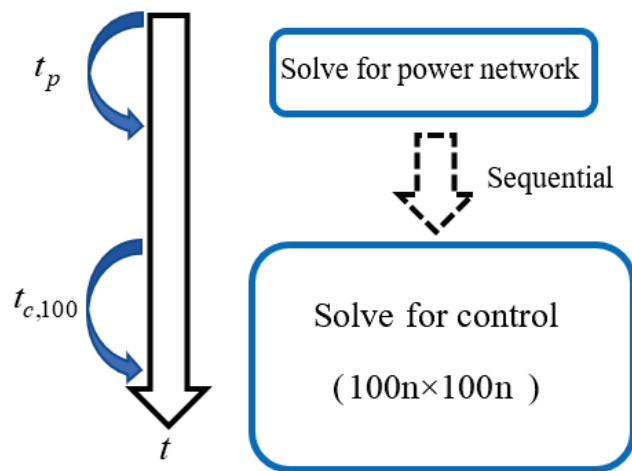


Figure 3.27 Solution scheme in the original approach (EMTP) for the computation of one time-step.

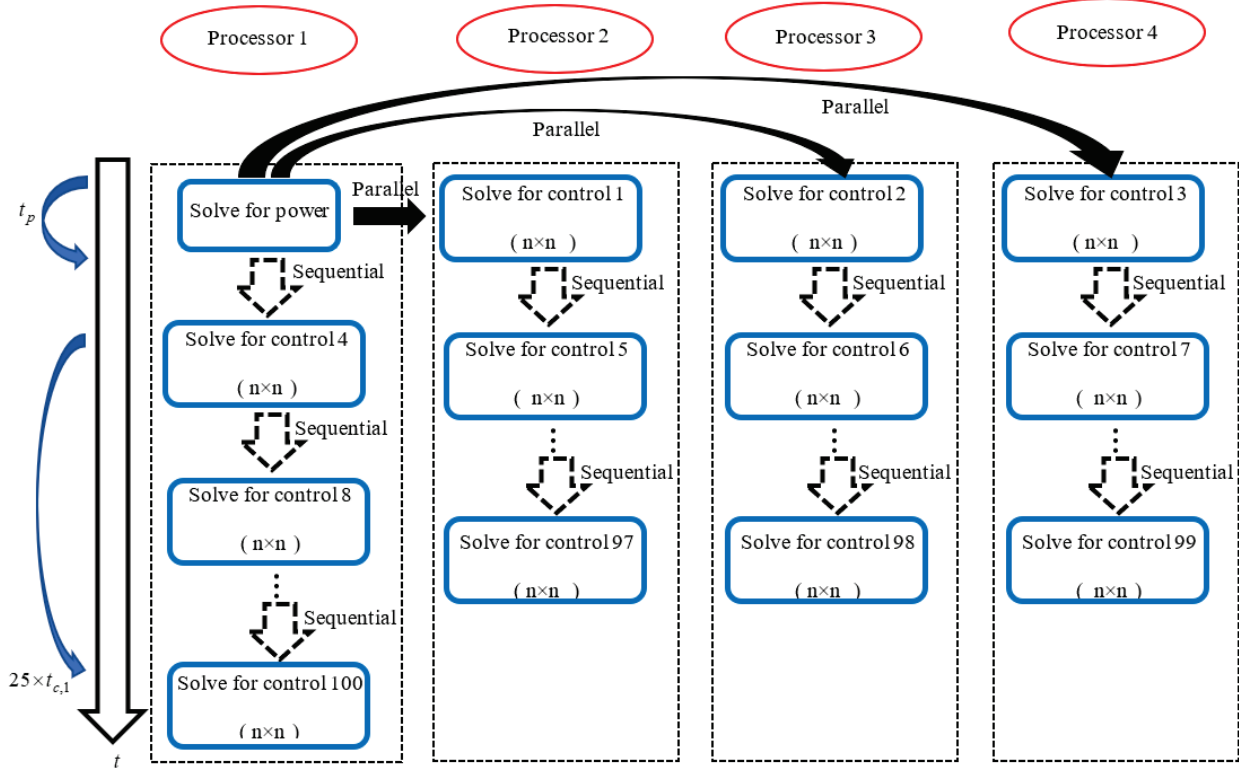


Figure 3.28 Solution scheme using the FMI-based co-simulation approach (parallel asynchronous mode) for the computation of one time-step.

In the original EMTP, as is shown in Figure 3.27, for the computation of one time-step, EMTP first solves the power network, then solves the large equation system consisting of 100 line relays. Ignoring time consumed on the other EMTP inner mechanisms, the total solution time can therefore be approximated as,

$$t_{tol,EMTP} = t_p + t_{c,100} \quad (3.1)$$

where t_p denotes the solution time for the power network, and $t_{c,100}$ represents the solution time for the control systems consisting of 100 line relays.

In the FMI-based co-simulation approach as shown in Figure 3.28, due to insufficient number of processors, only the solution of the power network, control 1, 2 and 3 of size $n \times n$ can be executed in parallel. The computation of the other 97 control systems are distributed sequentially among the 4 processors. The approximate total solution time, therefore, depends on the most load-heavy processor, which is Processor 1 in Figure 3.28,

$$t_{tol,Co-simulation} = t_p + 25 \times t_{c,1} \quad (3.2)$$

where $t_{c,1}$ denotes the solution time for the control system consisting of one line relay. It is worth mentioning that the computation load distribution among different processors is operating-system-dependent in reality, and only one possibility is presented in Figure 3.28 for explanatory purposes.

Hence, the difference in solution times using both approaches lies in that: in EMTP, the equation system of 100 relays is solved once, whereas in the FMI-based co-simulation approach, the equation system of one individual relay is solved 25 times. Considering the arithmetic complexity of sparse matrix solver for control systems used in the EMT-type solver [2], it is easy to deduce that the former is considerably more time-consuming than the latter.

Based on the discussion above, it is understood why a gain larger than the number of processors can be observed using the FMI-based co-simulation approach.

- Observation 2: Higher gains on a PC with more cores

From the test results on Network-1 (see Figure 3.23 and Figure 3.24) on PCs with different number of processors, it is observed that for a specific number of slaves in the system, the gains on the 16-core PC is higher than those on the 2-core PC. The reason is fairly straightforward: the more cores (logical processors) there are in a PC, the more the computation load of individual control systems is distributed. Each individual processor on a PC with more processors executes the solution of fewer control systems. Hence, the total solution time of a benchmark with a specific number of slaves is smaller on a PC with more processors.

- Observation 3: Higher gains on a benchmark with more slaves

It is observed on Network-2 on a 2-core PC (see Figure 3.25) that when the number of slaves increases from 100 to 116, the gains jump from approximately 7.05 to 8.3. In both cases, the number of decoupled slaves largely exceeds the number of available processors, so the computation of most slaves is executed sequentially. The reason for such an observation is due to the fact that in the original approach using EMTP, the extra time consumed to solve the equation system of 116 relays compared to solving that of 100 relays substantially exceeds the extra time for one processor to sequentially solve the equation system of one individual relay 4 times (4 processors share the computation load of the extra 16 relays in the FMI-based co-simulation approach) for reasons

elaborated in the explanation of Observation 1. This can also be seen in Table 3.9, of which only a snippet is presented below in Table 3.11.

Table 3.11 Snippet of comparison of solution times (s) on a PC with 2 cores (4 logical processors) for different scenarios, Network-2

Number of relays	100	116
EMTP (s)	23278	29926
FMI-based co-simulation (s)	3302	3606

As the number of slaves goes from 100 to 116, the total solution time using co-simulation increases from 3302 s to 3606 s. However, using EMTP, the total solution time has seen a much higher jump from 23278 s to 29926 s. Despite the slight increase in total solution time using co-simulation, the computation time gains in the case of 116 relays are still higher.

- Observation 4: Higher gains on Network-1 than Network-2 for similar numbers of slaves

It is observed in Figure 3.24 that on the 16-core PC, Network-1 has a gain of 9.2 with 44 slaves (4 wind generators and 40 slaves) whereas the gain for Network-2 with 40 relays is barely 4 (see Figure 3.26). This is due to the size difference between the two networks. For Network-1, the power network equations are 4826×4826 with 35718 non-zeros, and the size of the power network equations of Network-2 is 12120×12120 (125120 non-zeros). A major chunk of total solution time is consequently spent on solving the power network in the case of Network-2. It can also be observed for Network-2 with a specific number of slaves on the two different PCs, the gains on the 16-core PC are only slightly higher than those on the 2-core PC, indicating that a more distributed computation load of individual control systems for Network-2 could not efficiently counter the heavy solution burden brought by the large power network.

2) Sequential synchronous mode (Scenarios 2 and 3)

In this mode (Scenarios 2 and 3), a larger time-step is used in the control systems (slaves) whereas the main electrical network (master) adopts a relatively smaller time-step. All the slaves are individually solved once and remain idle while the master solves several times to catch up with the slave time-point. In comparison to the original approach in EMTP where all control systems are solved for every single time-step of the main electrical network, the computation time gain using this mode is self-evident. A few observations can be still made, which are presented as follows together with their associated analyses.

- Observation 1: Higher gains in scenarios with larger slave time-step

This is observed on all four figures from Figure 3.23 to Figure 3.26 (comparison between Scenarios 2 and 3). The larger the slave time-step is, the higher the gains are. This is due to the fact that the larger the slave time-step is, the fewer the solutions of individual slaves are executed in the system (more time being idle).

- Observation 2: Higher gains on a PC with more cores

This can be observed in all four figures from Figure 3.23 to Figure 3.26 for any specific number of slaves. It is worth pointing out that in the synchronous mode, the computation of master and slaves are performed sequentially. Nonetheless, there is no constraint with regards to the solution sequence between slaves placed on separate processors as the slaves only communicate with the master (not with each other). Therefore, this observation can be explained in the same fashion as in Observation 2 of the parallel asynchronous mode.

Observations 3 and 4 for the parallel asynchronous mode can be made for the sequential multistep synchronous mode as well. They can also be explained similarly here as in the parallel asynchronous mode. It is also observed that the performance comparison of the parallel asynchronous mode and the sequential multistep synchronous mode varies depending on the number of slaves and the number of cores a PC has. In general, for a test benchmark such as Network-1 with not many slaves (maximum 46 as in this test case), the parallel asynchronous mode is more efficient than the sequential multistep synchronous mode thanks to solution parallelism (on a 16-core PC with 32 logical processors); whereas for Network-2 on a 16-core PC, the performance of the parallel asynchronous mode deteriorates when there are a sufficient number of slaves in the system (from 100 to 116). This is because the number of individual decoupled slaves has far exceeded the parallelization capacity of the PC, sequentially solving many control systems at every single master time-step (parallel asynchronous mode) becomes, therefore, more time-consuming than solving all control systems at a much larger slave time-step.

3.3 Validation of the improved methodology

In this section, tests validating the improved methodology, which is the parallel multistep asynchronous mode and the linear extrapolation-based signal correction procedure proposed in Section 2.3, are performed on power system simulation benchmarks with control systems of

different levels of complexity. Once again, the tests are performed focusing on two aspects: accuracy and computation efficiency.

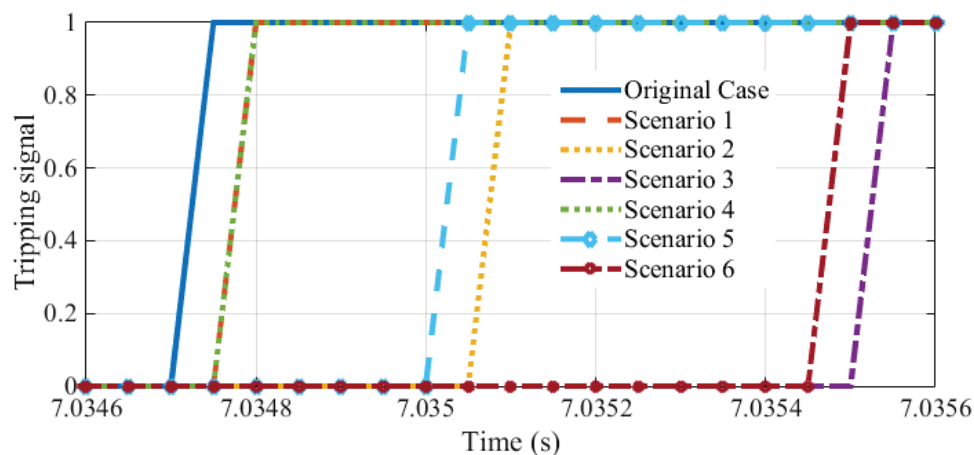
3.3.1 Accuracy validation—parallel multistep asynchronous mode

In this section, the accuracy of the parallel multistep asynchronous mode is verified by comparing line relay tripping instants after a fault. The tests performed in this section use 6 co-simulation scenarios implementing the FMI-based co-simulation approach with the parallel multistep asynchronous mode in which different co-simulation modes as well as different numerical integration time-steps are utilized in the control systems (line relays). This is shown in Table 3.12, where the Original Case is the original single-core benchmark without co-simulation, and Scenarios 2 and 3 are the scenarios implementing the parallel multistep asynchronous mode.

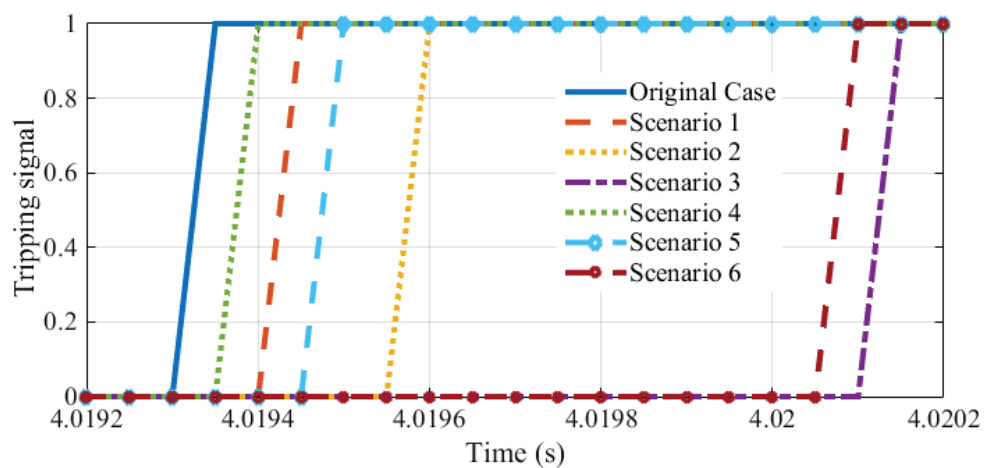
Table 3.12 Co-simulation mode scenarios, accuracy validation of the parallel multistep asynchronous mode

	Co-simulation Mode	Main network Δt (μs)	Relay Δt (μs)
Original Case	N/A	50	50
Scenario 1	Asynchronous	50	50
Scenario 2	Asynchronous	50	200
Scenario 3	Asynchronous	50	400
Scenario 4	Synchronous	50	50
Scenario 5	Synchronous	50	200
Scenario 6	Synchronous	50	400

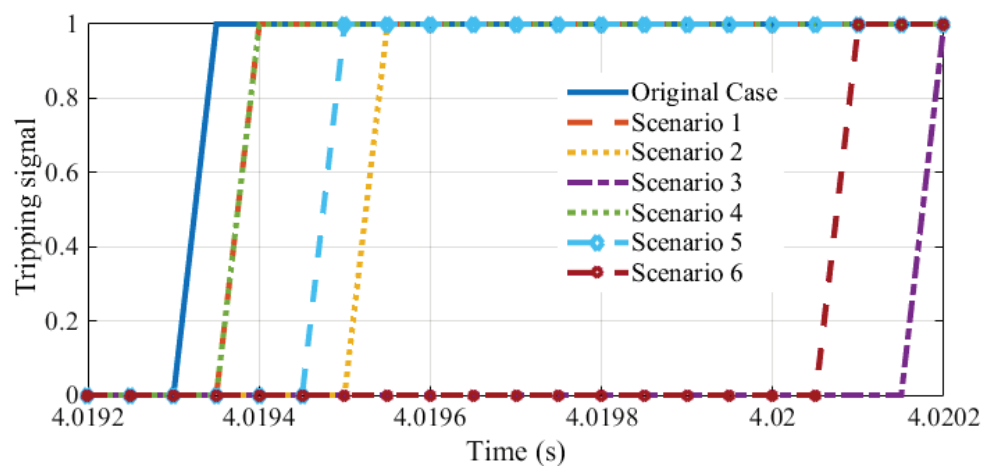
These tests validate the accuracy of co-simulation, particularly the parallel multistep asynchronous mode, in relay applications by examining line relay tripping instants at adjacent lines after phase-a-to-ground fault occurs at a line whose own relays are defective. It is the exact same test case in Section 3.2.1.2 using the Network-2 benchmark (Figure 3.18), with the addition of the parallel multistep asynchronous mode. The simulation interval, relay and fault settings remain unchanged (see Figure 3.16, Figure 3.17 and Table 3.3).



(a) Relay 1 tripping in Zone 4



(b) Relay 2 tripping in Zone 3



(c) Relay 4 tripping in Zone 3

Figure 3.29 Tripping instants of Relays 1, 2 and 4

The tripping instants of Relays 1, 2, and 4 are shown in Figure 3.29. The largest relay tripping instant delay observed in this test case is 850 μs (Relay 4 in Scenario 3 using the parallel multistep asynchronous mode) compared to the Original Case, which is approximately 1/19.6 cycle. Although this error is slightly larger than those in the scenarios proposed in the original methodology, it is still acceptably small catering to our need for simulation speedup and considering that the total clearing time (relay plus circuit breaker) typically ranges from 2 to 8 cycles on a 60 Hz base [95], demonstrating that the use of different numerical integration time-steps in different subsystems in the parallel multistep asynchronous mode does not affect accuracy.

3.3.2 Accuracy validation—linear extrapolation-based signal correction

In this section, the implementation of the linear extrapolation-based signal correction procedure on floating-point type signals is validated in terms of accuracy improvement. Two new benchmarks of different levels of complexity are used in the tests performed in this section. Again, all tests involve several simulation scenarios implementing the FMI-based co-simulation approach with different co-simulation modes as well as numerical integration time-steps. The test results with the implementation of the linear extrapolation-based signal correction procedure are presented in comparison with those without such a procedure. All test results are compared with the original single-core benchmark on EMTP without co-simulation for reference.

3.3.2.1 Test case Network-3

This test is performed on the benchmark shown in Figure 3.30. It is a simple 5-bus 12.5 kV system with one three-phase voltage source modeled using control devices. Other network components and device parameters can be seen in Figure 3.30. It uses 2 scenarios with different numerical integration time-steps used in the three-phase voltage source, as is shown in Table 3.13 in which the Original Case is the original single-core benchmark without co-simulation.

Table 3.13 Co-simulation mode scenarios, accuracy validation of the linear extrapolation-based signal correction procedure, Network-3

	Co-simulation mode	Main network Δt (μs)	Source Δt (μs)
Original Case	N/A	10	10
Scenario 1	Synchronous	10	50
Scenario 2	Synchronous	10	100

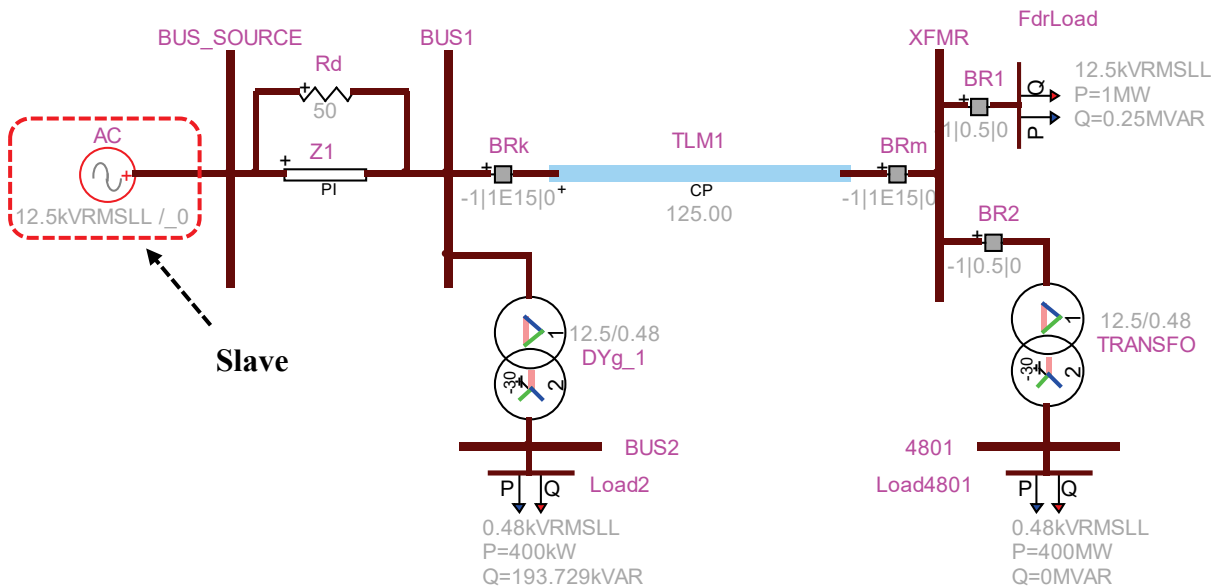
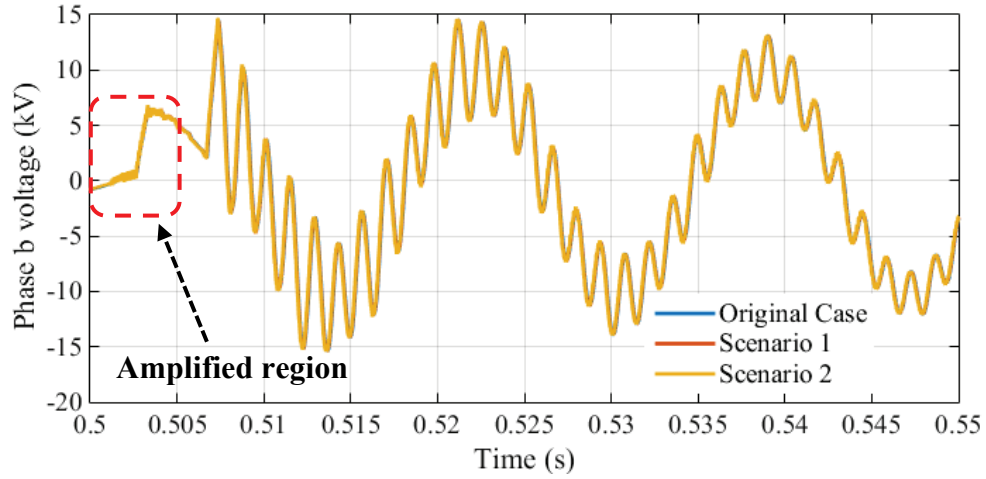


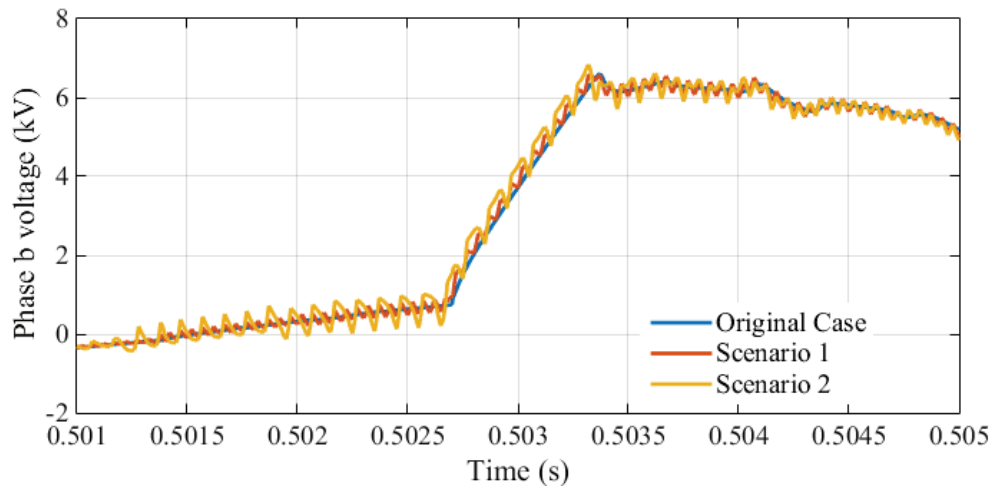
Figure 3.30 Network-3 benchmark

The simulation interval is 1 s for all simulation scenarios. At $t = 0.5s$, breakers BR1 and BR2 open, shedding loads FdrLoad and Load4801 as well as transformers TRANSFO from the system. In both scenarios, the three-phase voltage source is decoupled from the electrical network to form the slave subsystem.

Figure 3.31 and Figure 3.32 present the phase-b voltage at bus XFMR in the Original Case and both co-simulation scenarios without and with signal extrapolation respectively. For clear observation, only the interval between $t = 0.5s$ and $t = 0.55s$ after circuit breaker operation is shown in Figure 3.31 (a) and Figure 3.32 (a).

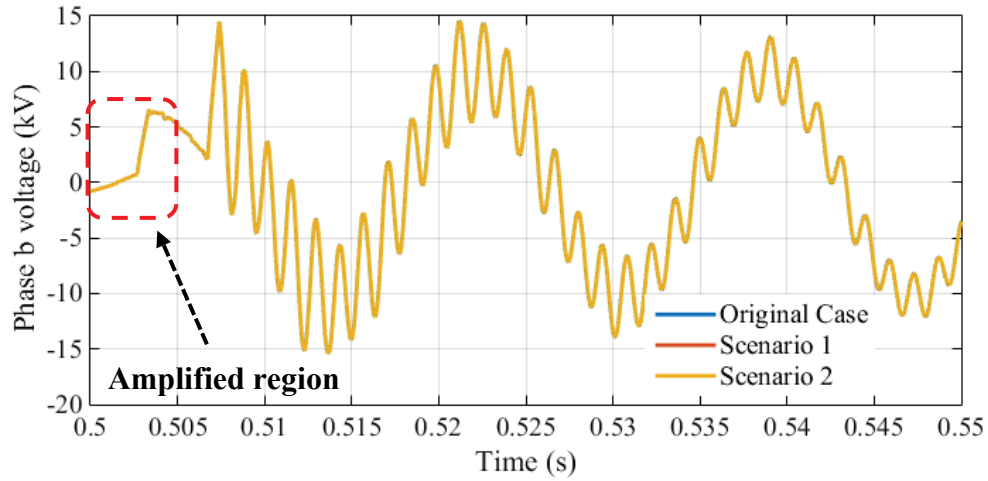


(a) Overall waveforms

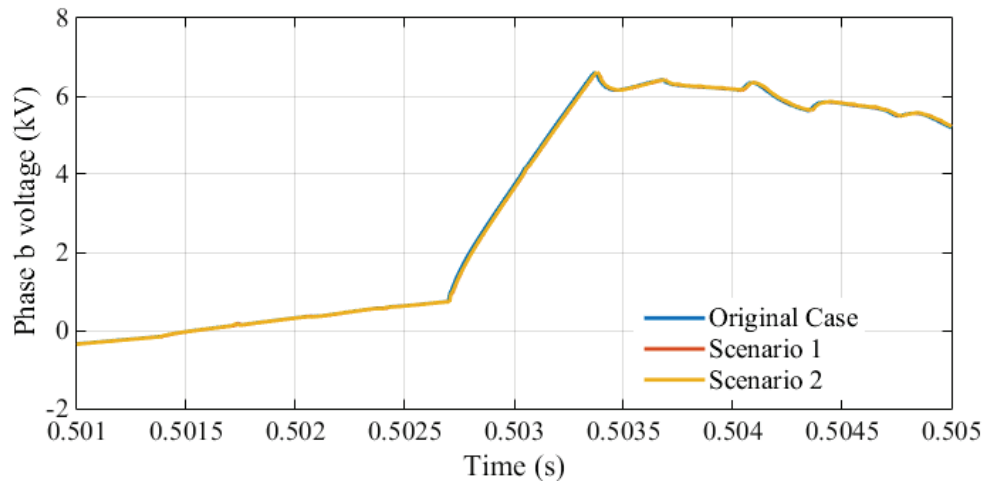


(b) Waveforms in the amplified region

Figure 3.31 Phase-b voltage at bus XFMR between $t = 0.5$ s and $t = 0.55$ s without signal correction with closer observation of waveforms in the amplified region



(a) Overall waveforms



(b) Waveforms in the amplified region

Figure 3.32 Phase-b voltage at bus XFMR between $t = 0.5s$ and $t = 0.55s$ with signal correction with closer observation of waveforms in the amplified region

Without linear extrapolation-based signal correction, although the voltage waveforms transients in the two co-simulation scenarios generally agree with that obtained in the Original Case, as can be seen in Figure 3.31 (a), a closer observation in the amplified region clearly indicates that the waveforms in the two co-simulation scenarios exhibits a “sawtooth” characteristic. This is due to the fact that the master keeps reading the same output from the slave as the former advances whilst the slave (three-phase voltage source in this test case), having a larger numerical integration time-step, remains idle (please refer to Section 3.2.1.3).

However, with the implementation of signal extrapolation on the slave outputs, this “sawtooth” characteristic can be largely smoothed out, allowing the waveforms from the two co-simulation scenarios to better agree with that obtained in the Original Case, as can be observed in Figure 3.32. It is noted that the extra one time-step delay manually added during master and slave data exchange for simulation stability and robustness still exists.

3.3.2.2 Test case Network-4

This test is performed using the Network-4 benchmark presented in Figure 3.33, which is a network consisting of 3 wind park feeders connected to a collector grid and an equivalent network represented by a Thevenin equivalent. The design of the wind park feeder F1 is presented in Figure 3.34 as an example. It contains 18 wind parks that can be included or excluded in the simulation. Wind park feeders F2 and F3 contain, respectively, 12 and 15 wind parks. The DFIG together with its control system for all wind generators in the 3 wind park feeders are shown in Figure 3.35. The detailed model [1] whose control system contains 2235 blocks, is used for all wind generators, with its design shown in Figure 3.36.

The test performed in this section uses 2 co-simulation scenarios with different co-simulation modes and different numerical integration time-steps. The Original Case provides the reference waveforms from the single-core benchmark without co-simulation. A multiphase unbalanced load-flow solution is performed to initialize the time-domain solution where a fault condition is simulated. A numerical integration time-step of 5 μ s is used in the electrical network for all shown examples. In this test, the results with four wind generators (two in F1, one in F2 and F3 each, with the other wind generators excluded from the simulation) for Scenarios 1 and 2 are presented, in which both scenarios (asynchronous and synchronous) with and without signal correction are

compared with the Original Case. The simulation interval is 10 s for all cases and a three-phase fault with a duration of 0.15 s occurs at $t = 5s$.

Table 3.14 Co-simulation mode scenarios, accuracy validation of the linear extrapolation-based signal correction procedure, Network-4

	Co-simulation mode	Main network Δt (μs)	WP control Δt (μs)
Original Case	N/A	5	5
Scenario 1	Asynchronous	5	100
Scenario 2	Synchronous	5	100

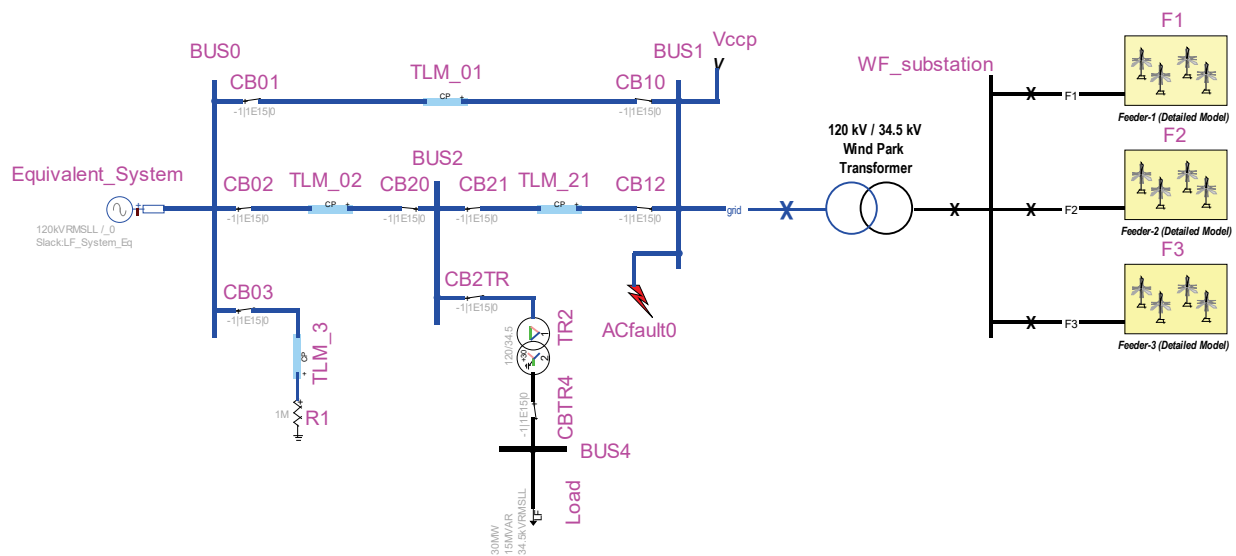


Figure 3.33 Network-4 benchmark

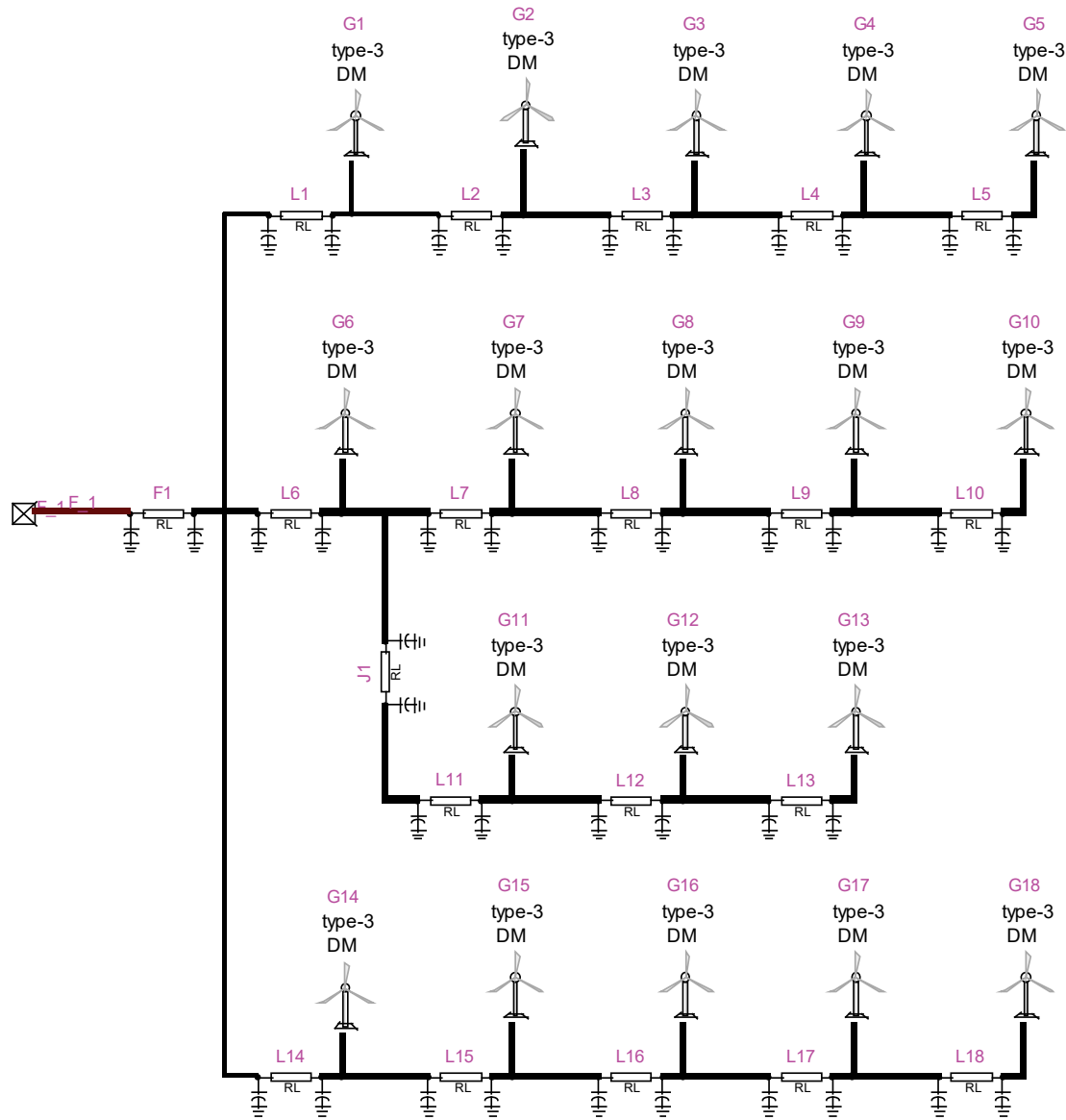


Figure 3.34 Design of the F1 wind park feeder

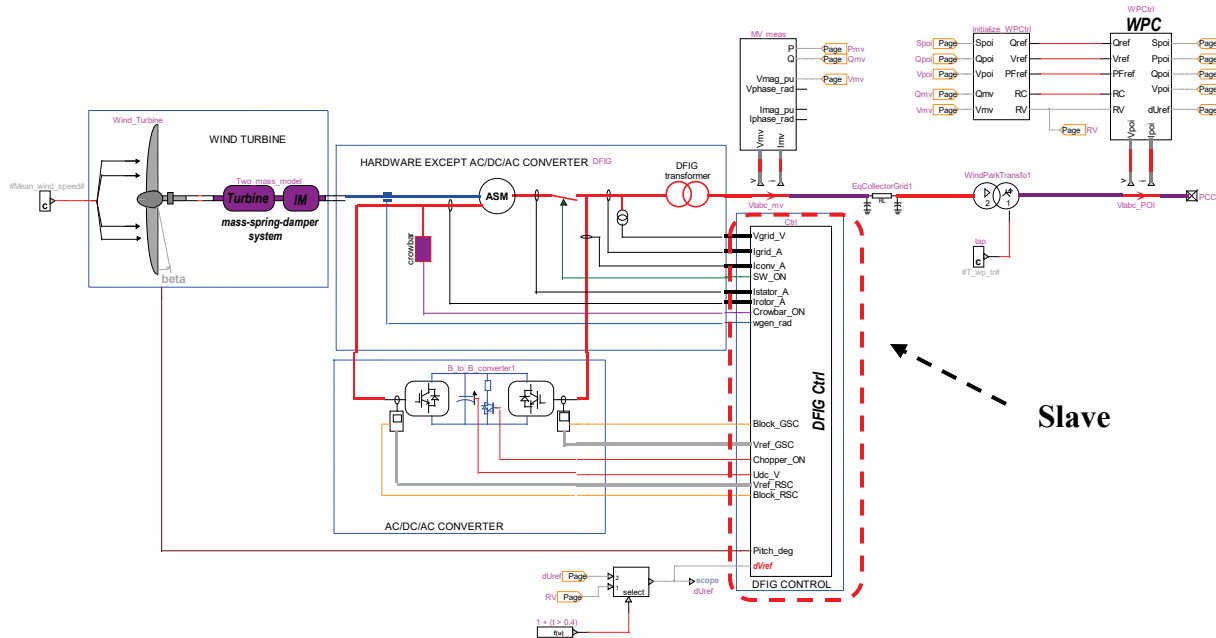


Figure 3.35 DFIG model with wind generator control systems modelled using the detailed model

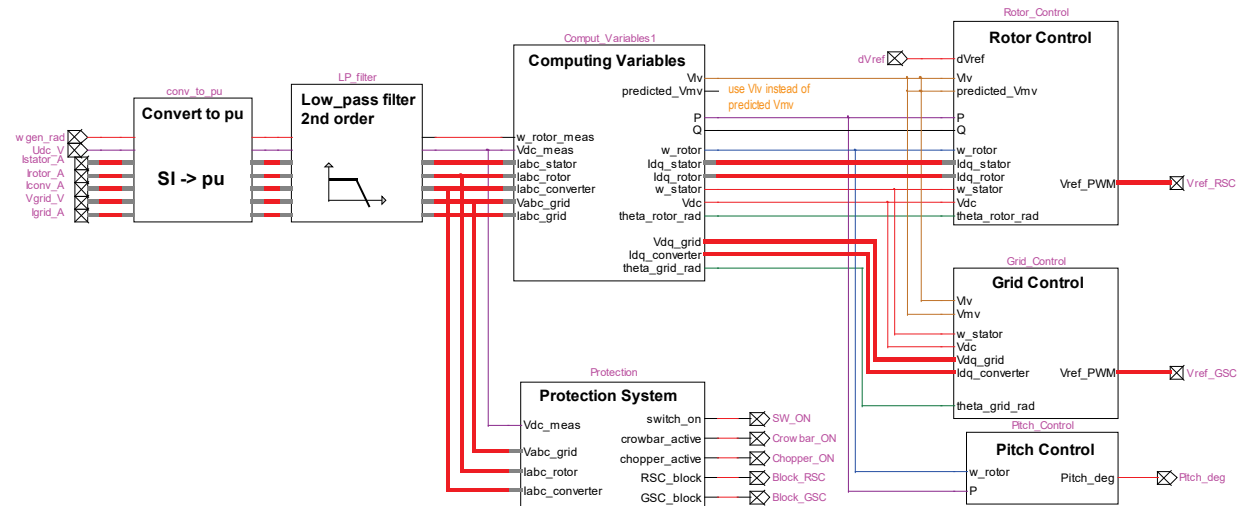


Figure 3.36 Control system design for the DFIG model in the wind generator

The average active power from the three feeders and the phase-a voltage at the 34.5 kV bus, are presented in Figure 3.37 and Figure 3.38 with clear observation in amplified intervals.

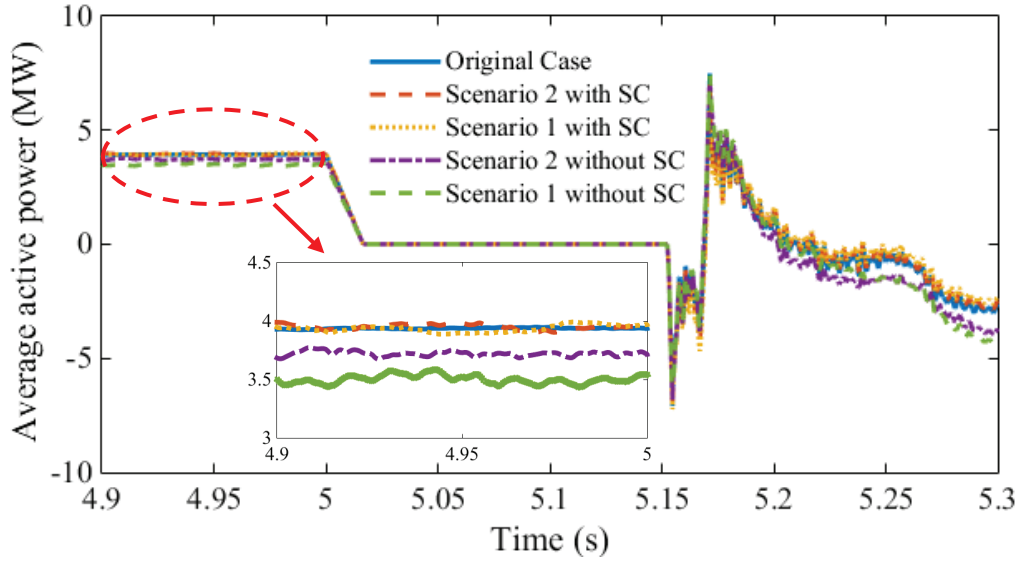


Figure 3.37 Average active power from wind park feeders

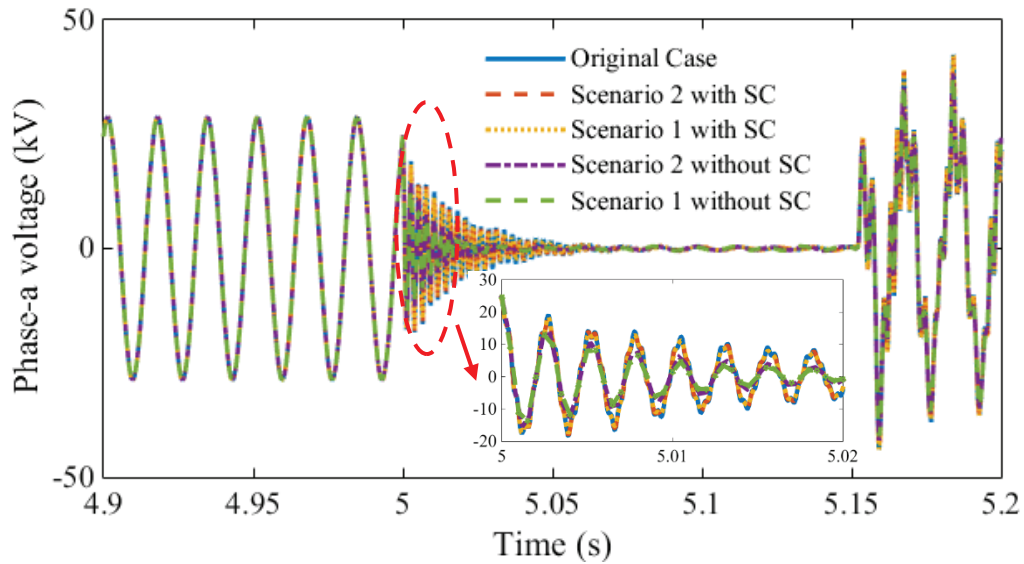


Figure 3.38 Phase-a voltage at the 34.5 kV bus

It is observed the accuracy of both co-simulation modes can be greatly improved with the implementation of the linear extrapolation-based signal correction procedure on floating-point type control signals compared to the Original Case. Additionally, it can be observed that the parallel multistep asynchronous mode (Scenario 1) offers comparable accuracy with the sequential multistep synchronous mode (Scenario 2) proposed in the original methodology, further

demonstrating that the implementation of the parallel multistep asynchronous mode does not affect accuracy.

Overall, it can be concluded that better simulation accuracy can be achieved by the implementation of floating-point type signal extrapolation in a multistep simulation environment.

3.3.3 Computation time gains

The objective of this section is to demonstrate the numerical performance advantages of the parallel multistep asynchronous mode proposed in the improved methodology on practical power system networks of different levels of complexity by comparing results obtained from benchmarks using the FMI-based co-simulation approach (including the parallel multistep asynchronous mode) with those using the single-core benchmarks on EMTP without co-simulation. Network-1, -2 and -4, shown in Figure 3.10, Figure 3.18, and Figure 3.33 respectively, are used as test benchmarks. Once again, a multiphase unbalanced load-flow solution is performed prior to time-domain simulation where a fault condition is simulated. A simulation interval of 2 s is used in all test cases.

3.3.3.1 Test case Network-1

The tests performed in this section include the Original Case without co-simulation and 5 co-simulation scenarios (see Table 3.15) implementing the parallel multistep asynchronous mode (Scenarios 2 and 3) on benchmark Network-1 (see Figure 3.10), where different co-simulation modes and numerical integration time-steps are employed in the control systems (offshore wind generator controls and line relays). The tests are performed on the 16-core PC whose hardware and software configurations are given in Table 3.5.

Table 3.15 Co-simulation mode scenarios, computation time gains with the parallel multistep asynchronous mode, Network-1 and -2

	Co-simulation mode	Main network Δt (μ s)	WP control Δt (μ s)	Relay Δt (μ s)
Original Case	N/A	50	50	50
Scenario 1	Asynchronous	50	50	50
Scenario 2	Asynchronous	50	200	200
Scenario 3	Asynchronous	50	200	400
Scenario 4	Synchronous	50	200	200
Scenario 5	Synchronous	50	200	400

Similar to the tests on computation time gains of the original methodology, different numbers of line relays, presented in Figure 3.16, are added into this benchmark (see Table 3.16). The slave subsystems in this test case thus consist of all 4 offshore aggregated wind generator control systems (average model, see [1]) and line relays implemented using co-simulation in all scenarios.

The total solution time for Network-1 in different scenarios compared to the Original Case are presented in Table 3.16, and the solution time speedup in comparison to the Original Case with respect to the number of slaves (offshore wind generator controls and line relays) is shown in Figure 3.39.

Table 3.16 Comparison of solution times (s) on the PC with 16 cores for different scenarios, Network-1, improved methodology

Number of WP controls	4						
Number of relays	11	16	22	28	34	40	46
Total solution time (s)							
Original Case	824	1292	1963	2415	2989	3901	4644
Scenario 1	220	248	299	345	375	424	476
Scenario 2	223	246	284	313	348	382	414
Scenario 3	213	239	270	303	330	359	387
Scenario 4	256	296	346	400	461	519	576
Scenario 5	242	275	313	355	400	444	480

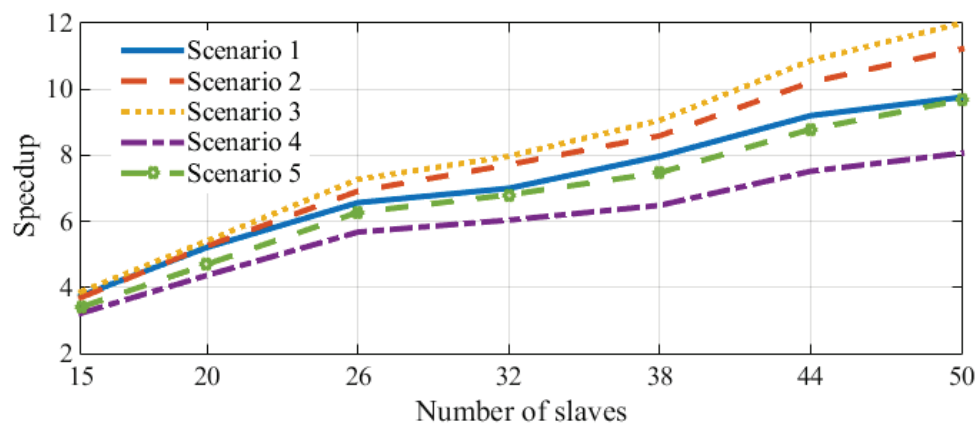


Figure 3.39 Total solution time speedup with respect to the number of slaves in Network-1 on a 16-core PC, improved methodology

3.3.3.2 Test case Network-2

The second batch of tests are performed on Network-2 benchmark shown in Figure 3.18 with different numbers of line relays used in each test (see Table 3.17) using the same 16-core PC (Table 3.5). They also include the Original Case without co-simulation and 5 co-simulation scenarios as shown in Table 3.15. Offshore wind parks are not present in this test case. Therefore, the slave subsystems in this test case are only comprised of line relays in all scenarios.

The total solution times for Network-2 in different scenarios compared to the Original Case are presented in Table 3.17, and the solution time speedup in comparison to the Original Case with respect to the number of slaves (line relays) is shown in Figure 3.40.

Table 3.17 Comparison of solution times (s) on the PC with 16 cores for different scenarios, Network-2, improved methodology

Number of relays	15	20	40	60	80	100	116
Total solution time (s)							
Original case	2419	2840	5494	9288	14426	19249	24351
Scenario 1	1237	1287	1471	1778	1939	2184	2563
Scenario 2	1214	1271	1461	1665	1860	2073	2259
Scenario 3	1191	1228	1409	1617	1804	1998	2174
Scenario 4	1259	1328	1492	1795	2091	2288	2601
Scenario 5	1246	1294	1418	1654	1862	2061	2263

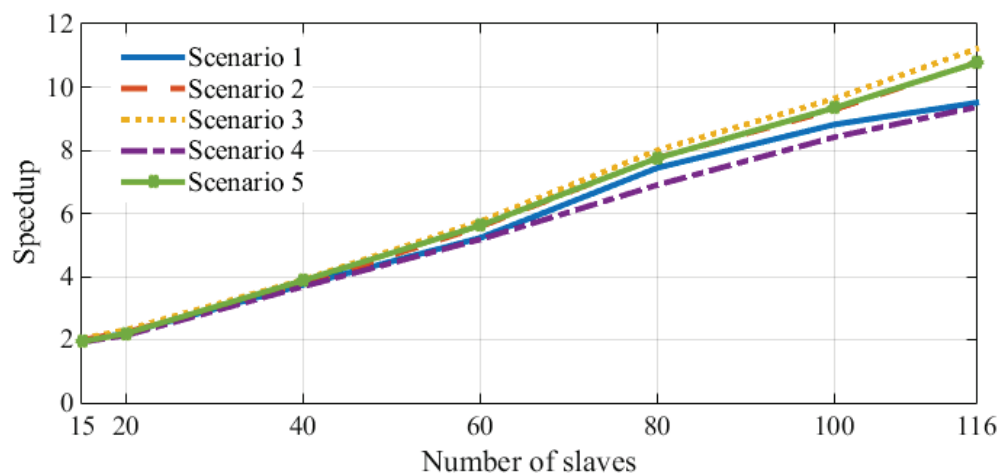


Figure 3.40 Total solution time speedup with respect to the number of slaves in Network-2 on a 16-core PC, improved methodology

3.3.3.3 Test case Network-4

The tests performed in this section use the Original Case without co-simulation and 5 new co-simulation scenarios, as is shown in Table 3.19, on the Network-4 benchmark presented in Figure 3.33. A numerical integration time-step of $5 \mu\text{s}$ is used in the electrical network for all shown examples. A 24-core PC is used in these tests, with its hardware and software configurations given in Table 3.18. It is noted that different numbers of wind generators (slaves) are used in the three feeders to test the numerical performance and scalability of parallel multistep asynchronous mode in the improved methodology. The exact numbers of wind generators used in this test case can be seen in Table 3.20.

The total solution times for Network-4 in different scenarios compared to the Original Case are presented in Table 3.20, and the solution time speedup in comparison to the Original Case with respect to the number of slaves (wind generator controls) is shown in Figure 3.41.

Table 3.18 Hardware and software configurations of the 24-core PC

Processor	Intel (R) Xeon (R) E5-2650 v4
Number of cores	24
Number of logical processors	48
Installed memory (RAM)	32.0 GB
Base speed	2.2 GHz
Operating system	Microsoft Windows 10 Pro

Table 3.19 Co-simulation mode scenarios, computation time gains with the parallel multistep asynchronous mode, Network-4

	Co-simulation mode	WP control Δt (μs)
Original Case	N/A	5
Scenario 1	Asynchronous	5
Scenario 2	Asynchronous	50
Scenario 3	Asynchronous	100
Scenario 4	Synchronous	50
Scenario 5	Synchronous	100

Table 3.20 Comparison of solution times (s) on the PC with 24 cores (48 logical processors) for different scenarios, Network-4, improved methodology

Number of WPs	10	15	20	25	30	35	40	45
Total solution time (s)								
Original Case	2162	3622	5070	7056	8809	11696	13856	17496
Scenario 1	443	677	855	1124	1310	1509	1719	1901
Scenario 2	432	574	745	989	1185	1371	1543	1704
Scenario 3	353	546	622	775	896	1157	1207	1394
Scenario 4	629	954	1305	1561	1846	2123	2422	2727
Scenario 5	517	734	986	1260	1480	1737	1985	2205

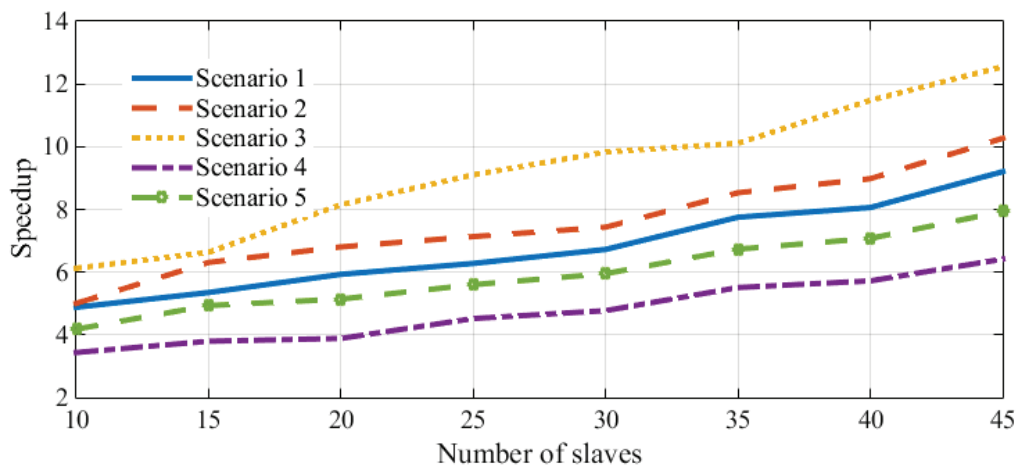


Figure 3.41 Total solution time speedup with respect to the number of slaves in Network-4 on a 24-core PC, improved methodology

3.3.3.4 Discussion

Overall, it can be observed in Figure 3.39, Figure 3.40, and Figure 3.41 that substantial solution time speedup is obtained in power system transient simulations in all scenarios implementing the FMI-based co-simulation approach. However, the parallel multistep asynchronous mode (Scenarios 2 and 3 in all test cases) renders the highest speedup, which was not observed in the scenarios implementing the co-simulation modes proposed in the original methodology, confirming that the parallel multistep asynchronous mode proposed in the improved methodology further enhances simulation efficiency.

Apart from the conclusions already drawn regarding the speedup achieved in different modes in the discussion of computation time gains of the original methodology as well as the discussion in Section 3.2.2.3, it is also worth mentioning that the results presented in Figure 3.39, Figure 3.40, and Figure 3.41 do not clearly show speedup saturation although the number of decoupled subsystems in some cases far exceeds the number of cores and logical processors of the PC. The high scalability in the FMI-based co-simulation approach indicates even higher speedup could be achieved in protection studies of large-scale networks using even more and different types of protection relays and control systems.

3.4 Conclusion

This chapter presented simulation results using the FMI-based parallel and multistep approach developed in this thesis. First of all, the design of the co-simulation interface on an EMT-type solver [2] using the FMI standard was validated, establishing and confirming the platform for further tests. Then both the original and the improved methodologies were validated for accuracy and computation time gains using realistic power system simulation benchmarks with complex control systems of different complexity, particularly wind generator controls and protection relays. The tests performed using the original methodology demonstrated the numerical advantage of the FMI-based co-simulation approach in terms of computation time gains while keeping the accuracy properly maintained. The validation of the improved methodology testified to the numerical performance enhancement in both accuracy control and solution speedup using the parallel multistep asynchronous mode and the linear extrapolation-based signal correction.

CHAPTER 4 A PARALLEL-IN-TIME APPROACH BASED ON PARALLEL-IN-TIME EQUATION GROUPING AND REORDERING (PEGR)

This chapter presents the development of a new parallel-in-time approach implementing the parallel-in-time equation grouping and reordering (PEGR) technique for power system EMT simulations. The PEGR technique, originally proposed in [30], is a novel numerical solution scheme that is formulated in state-space and could potentially formulate the solution by trapezoidal integration for $T = 2^\tau$ (τ is an integer) time-steps as a single problem involving $T \times n$ equations and $T \times n$ unknowns (n is the number of state variables). These equations can then be solved in $\tau = \log_2 T$ steps using $T/2$ parallel processors. This technique is based on the identification of truly independent tasks that could be performed simultaneously with an ideal number of parallel processors after preprocessing procedures that consist of the construction of a huge matrix containing the solutions at all time-points as well as a series of recursive row and column reordering. Since building a huge matrix containing solutions at all time-points and solving it using $T/2$ parallel processors are unrealistic in practice due to the limitations of PC memory and the number of available processors, this chapter focuses on adapting the PEGR technique to practical power system EMT simulations on the base of modified-augmented-nodal analysis (MANA) [96] which demonstrates several advantages over the conventional nodal analysis as well as state-space formulation.

This chapter starts with a brief introduction of the modified-augmented-nodal analysis (MANA), which is the theoretical base for the new parallel-in-time approach. It then explains in detail the adaptation of the PEGR technique originally formulated in state-space to MANA through the expansion of network equations. Subsequently, the equation grouping and recursive row and column reordering scheme used in this new parallel-in-time approach as well as in the original PEGR technique is demonstrated for a case in which the solutions of 16 time-steps are grouped. Detailed programming considerations in the implementation of the PEGR-based parallel-in-time approach are also discussed, including parallelization using OpenMP, measures adopted to minimize concurrency issues in a parallel programming environment, block elements extraction from **L** and **U** factors for backward and forward substitutions, , as well as treatment of arbitrary points of topological changes and numbers of simulation points.

4.1 Formulation of the PEGR technique in MANA

In this section, a brief introduction of the modified-augmented-nodal analysis (MANA) used as the theoretical base for the EMT-type solver [2] and the new parallel-in-time approach proposed in this chapter is presented, followed by the adaptation of the original PEGR technique in MANA through network equation expansion. The equation grouping and recursive row and column reordering technique used in the new parallel-in-time approach as well as the original PEGR technique is then explained in detail, using an example that groups the solutions of 16 time-steps. This is due to the fact that the original PEGR technique formulates the solution by trapezoidal integration for $T = 2^\tau$ (τ is an integer) time-steps as a single problem, an example where 16 ($16 = 2^4$) time-steps are grouped, reordered and solved would make it easy to understand this approach. The meaning of “grouping” used in this chapter will become self-evident in the following discussions. The newly proposed parallel-in-time approach and the original PEGR technique are general and can be used to group the solutions of different numbers of time-steps, based on the computation capacity of the PC.

4.1.1 Modified-augmented-nodal analysis (MANA)

The formulation of the parallel-in-time approach is based on the modified-augmented-nodal analysis (MANA) formulation method [2]. This method offers several advantages [96] over classical nodal analysis and state-space formulation. In MANA, the system of equations is generic and can include different types of unknowns in addition to voltage. A brief introduction of the MANA formulation is provided in this section. More details can be found in [2]. The general MANA equations of a power system network can be written as,

$$\mathbf{Ax} = \mathbf{b} \quad (4.1)$$

and can be expanded to,

$$\overbrace{\begin{bmatrix} \mathbf{Y}_n & \mathbf{V}_{\text{adj}}^t & \mathbf{D}_{\text{bdepr}} & \mathbf{S}_{\text{adj}}^t \\ \mathbf{V}_{\text{adj}} & \mathbf{0} & \mathbf{0} & \mathbf{0} \\ \mathbf{D}_{\text{bdepr}} & \mathbf{0} & \mathbf{0} & \mathbf{0} \\ \mathbf{S}_{\text{adj}} & \mathbf{0} & \mathbf{0} & \mathbf{S}_z \end{bmatrix}}^{\mathbf{A}} \overbrace{\begin{bmatrix} \mathbf{V}_n \\ \mathbf{I}_{V_s} \\ \mathbf{I}_{V_d} \\ \mathbf{I}_{\text{SW}} \end{bmatrix}}^{\mathbf{x}} = \overbrace{\begin{bmatrix} \mathbf{I}_n \\ \mathbf{V}_s \\ \mathbf{0} \\ \mathbf{0} \end{bmatrix}}^{\mathbf{b}} \quad (4.2)$$

where,

- \mathbf{Y}_n is the linear network admittance matrix, non-symmetric
- \mathbf{V}_{adj} is the voltage source adjacency matrix, for all voltage source types
- \mathbf{D}_{bdepr} and \mathbf{D}_{bdepc} (row and column contribution matrices) are used for holding branch dependent relations
- \mathbf{S}_{adj} is the adjacency matrix of closed switch type devices
- \mathbf{S}_z is a diagonal and unitary matrix for open switch type devices
- \mathbf{V}_n is the vector of unknown node voltages
- \mathbf{I}_{V_s} is the vector of unknown voltage source currents
- \mathbf{I}_{Vd} is the vector of unknown currents in dependent voltage source branches
- \mathbf{I}_{SW} is the vector of unknown switch currents
- \mathbf{I}_n is the vector of known nodal current injections
- \mathbf{V}_s is the vector of known voltage sources

Such an equation (4.1) can be used for both time-domain and steady-state solutions. It is a real system in time-domain and a complex system in steady-state. Considering equation (4.1) is non-symmetric and can also accommodate generic equations such as,

$$k_1 v_k + k_2 v_m + k_3 i_x + k_4 i_y + \dots = b_z \quad (4.3)$$

where the terms on the left contribute coefficients into the \mathbf{x} matrix for voltage and current unknowns, and b_z is a cell in the \mathbf{b} vector, other models, such as ideal transformers with tap control can be also easily accommodated.

The sparse system of equations (4.1) is solved for \mathbf{x} at each time-point in time-domain, after updating the non-symmetric coefficient matrix \mathbf{A} and the vector of known variables \mathbf{b} . The vector \mathbf{b} contains history terms and other independent functions.

4.1.2 Expansion of network MANA equations

The n^{th} order differential equation given in state-space,

$$\dot{x} = Ax + f(t) \quad (4.4)$$

with the initial condition,

$$x(0) = x_0 \quad (4.5)$$

can be rearranged using the trapezoidal rule into the form of,

$$x(t + \Delta t) = x(t) + \frac{\Delta t}{2} [Ax(t) + f(t) + Ax(t + \Delta t) + f(t + \Delta t)] \quad (4.6)$$

Equation (4.6) incorporates the solutions at two consecutive time-points $x(t)$ and $x(t + \Delta t)$, demonstrating the interdependency between solutions of consecutive time-points. This is the foundation based on which the PEGR technique can be implemented [30].

In the original MANA formulation for power system networks, the history terms of energy storage components such as inductors, capacitors and multiphase pi-sections are contained in \mathbf{b} and are updated for the calculation of each time-point in time-domain simulations, as was previously explained. The solutions at two consecutive time-points are not directly formulated into one single equation in MANA as in (4.6). Therefore, the PEGR technique cannot be directly implemented into MANA formulation. The following discussion presents how to rearrange the network MANA equations (4.1) into the form of (4.6) such that the equation grouping and recursive row and column reordering scheme in PEGR can be applied, using multiphase pi-section as an example. Other energy storage network components such as inductors and capacitors can be treated in a similar fashion.

The generalized multiphase pi-section component connected between a set of left-hand nodes \mathbf{k} and right-hand nodes \mathbf{m} is shown in Figure 4.1.

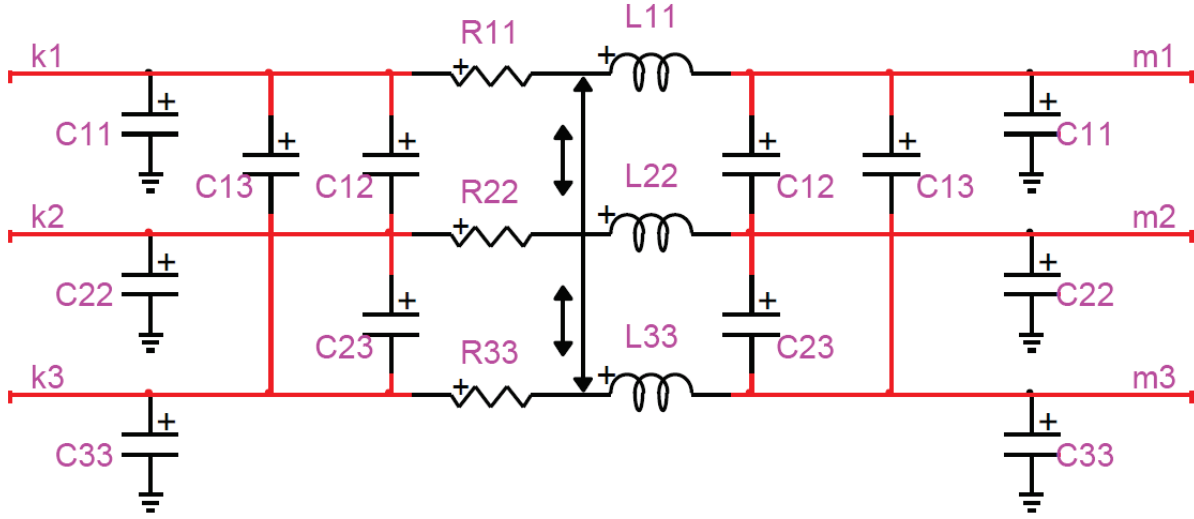


Figure 4.1 Multiphase pi-section schematic

Here nodes **k** and **m** are in bold because they represent a set of nodes. This component has history terms in its time-domain formulation in MANA. Using the trapezoidal rule, the equations for the RL section and the C section of the multiphase pi-section can be given by,

$$\mathbf{S}\mathbf{v}_{\mathbf{km},t} - \mathbf{i}_{RL,t} = -\mathbf{S}\mathbf{v}_{\mathbf{km},t-\Delta t} - \mathbf{S}\mathbf{H}\mathbf{i}_{RL,t-\Delta t} \quad (4.7)$$

$$\frac{2}{\Delta t}\mathbf{C}\mathbf{v}_t - \mathbf{i}_{C,t} = \frac{2}{\Delta t}\mathbf{C}\mathbf{v}_{t-\Delta t} + \mathbf{i}_{C,t-\Delta t} \quad (4.8)$$

where,

$$\mathbf{S} = \left[\mathbf{R} + \frac{2}{\Delta t}\mathbf{L} \right]^{-1}$$

$$\mathbf{H} = \left[\frac{2}{\Delta t}\mathbf{L} - \mathbf{R} \right]$$

and, once again, the subscripts **k** and **m** represent the nodes between which the multiphase pi-section is connected.

Suppose the system (4.1) has n multiphase pi-sections. Incorporating these RL current terms \mathbf{I}_{RL1} , ..., \mathbf{I}_{RLn} and shunt current terms \mathbf{I}_{C1} , ..., \mathbf{I}_{Cn} into the vector of unknowns \mathbf{x} , the network equations (4.1) can be expanded, as is shown in Figure 4.2.

For **N** :

$$\mathbf{N}(\mathbf{k}, \mathbf{cindex}) = \mathbf{N}(\mathbf{m}, \mathbf{cindex}) = \mathbf{I} \quad (4.13)$$

$$\mathbf{N}(\mathbf{k}, \mathbf{rlindex}) = -\mathbf{SH} \quad (4.14)$$

$$\mathbf{N}(\mathbf{m}, \mathbf{rlindex}) = \mathbf{SH} \quad (4.15)$$

For **P** :

$$\mathbf{P}(\mathbf{cindex}, \mathbf{cindex}) = \mathbf{I} \quad (4.16)$$

$$\mathbf{P}(\mathbf{rlindex}, \mathbf{rlindex}) = -\mathbf{SH} \quad (4.17)$$

The other elements in submatrices **M**, **N** and **P** are zeros. Indicating interdependency between solutions at two consecutive time-points, The equation shown in Figure 4.2 corresponds to the base equation of PEGR (4.6) and therefore allows the equation grouping and recursive row and column reordering scheme of the PEGR technique to be applied in a MANA-based formulation. As was mentioned previously, the expanded network MANA equations with individual inductors and capacitors can be derived similarly. Now let,

$$\mathbf{D} = \begin{bmatrix} & & & \mathbf{A} & & & \mathbf{0} \\ \cdots & \mathbf{S}_1 & \cdots & -\mathbf{S}_1 & \cdots & \cdots & \cdots \\ \cdots & \frac{2}{\Delta t} \mathbf{C}_{1,1} & \cdots & \cdots & \cdots & \cdots & \cdots \\ \cdots & \cdots & \cdots & \frac{2}{\Delta t} \mathbf{C}_{1,2} & \cdots & \cdots & \cdots \\ \cdots & \cdots & \cdots & \cdots & \cdots & \cdots & \cdots \\ \cdots & \cdots & \mathbf{S}_n & \cdots & -\mathbf{S}_n & \cdots & \cdots \\ \cdots & \cdots & \frac{2}{\Delta t} \mathbf{C}_{n,1} & \cdots & \cdots & \cdots & \cdots \\ \cdots & \cdots & \cdots & \cdots & \frac{2}{\Delta t} \mathbf{C}_{n,2} & \cdots & \cdots \end{bmatrix} \quad \mathbf{-I}$$

$$\begin{aligned}
 \mathbf{O} = & \left[\begin{array}{ccccccc} & & & \mathbf{M} & & \mathbf{N} & \\ & & & & & & \\ \dots & -\mathbf{S}_1 & \dots & \mathbf{S}_1 & \dots & \dots & \dots \\ & \frac{2}{\Delta t} \mathbf{C}_{1,1} & \dots & \dots & \dots & \dots & \dots \\ \dots & \dots & \dots & \frac{2}{\Delta t} \mathbf{C}_{1,2} & \dots & \dots & \dots \\ \dots & \dots & \dots & \dots & \dots & \dots & \mathbf{P} \\ \dots & \dots & -\mathbf{S}_n & \dots & \mathbf{S}_n & \dots & \dots \\ \dots & \dots & \frac{2}{\Delta t} \mathbf{C}_{n,1} & \dots & \dots & \dots & \dots \\ \dots & \dots & \dots & \dots & \frac{2}{\Delta t} \mathbf{C}_{n,2} & \dots & \dots \end{array} \right] \\
 \underline{\mathbf{x}}_t = & \left[\begin{array}{c} \mathbf{x}_t \\ \mathbf{I}_{RL1,t} \\ \mathbf{I}_{C1,1,t} \\ \mathbf{I}_{C1,2,t} \\ \vdots \\ \mathbf{I}_{RLn,t} \\ \mathbf{I}_{Cn,1,t} \\ \mathbf{I}_{Cn,2,t} \end{array} \right] \\
 \underline{\mathbf{b}}_t = & \left[\begin{array}{c} \mathbf{b}_t \\ \mathbf{0} \\ \vdots \\ \vdots \\ \vdots \\ \vdots \\ \vdots \\ \vdots \\ \mathbf{0} \end{array} \right]
 \end{aligned}$$

The equation in Figure 4.2 can be rewritten in a more compact form,

$$\mathbf{D}\underline{\mathbf{x}}_t - \mathbf{O}\underline{\mathbf{x}}_{t-\Delta t} = \underline{\mathbf{b}}_t \quad (4.18)$$

In the expanded network equations (4.18), vectors $\underline{\mathbf{x}}_t$ and $\underline{\mathbf{b}}_t$ denote the expanded unknown and right-hand-side vectors respectively. It is worth noting that vector $\underline{\mathbf{b}}_t$ is now devoid of history terms.

4.1.3 Grouping and recursive row and column reordering of expanded network MANA equations

The original proposition of the PEGR technique restricts the total number of solution steps T of a simulation to be $T = 2^\tau$ (τ is an integer) in order that the simulation can be solved in $\log_2 T$ steps using $T/2$ processors [30]. This is unrealistic in practice in that a simulation can require any number of solution steps and $T/2$ processors are usually unavailable to the user if the simulation takes too many solution steps (T is an extremely large number). Hence, instead of grouping the solutions at all T time-steps to construct a huge matrix as in the original PEGR proposition [30], this section presents the implementation of the equation grouping and recursive row and column reordering scheme used in the original PEGR technique in the new MANA-based parallel-in-time approach with an example showing the grouping of network equations of 16 ($16 = 2^4$) time-steps, thus solving the network for every 16 time-steps. It is noted that the algorithm can also accommodate the solution of every 32, 64, 128, ..., T time-steps ($T = 2^\tau$, τ is an integer), depending on the number of parallel processors available to the user. It is also worth mentioning that the treatment of arbitrary topological changes and numbers of solution time-steps will be presented in a later section.

Given equation (4.18), we can define,

$$\hat{\underline{\mathbf{x}}} = [\underline{\mathbf{x}}_1, \underline{\mathbf{x}}_2, \dots, \underline{\mathbf{x}}_{16}]$$

$$\hat{\underline{\mathbf{b}}} = [\underline{\mathbf{b}}_1, \underline{\mathbf{b}}_2, \dots, \underline{\mathbf{b}}_{16}]$$

Clearly, vectors $\hat{\underline{\mathbf{x}}}$ and $\hat{\underline{\mathbf{b}}}$ contain the solutions and the right-hand-side values of the first 16 time-steps. A set of 16 vector equations of (4.18) can thus be written as,

$$\begin{aligned}
\mathbf{D}\underline{\mathbf{x}}_1 &= \mathbf{O}\underline{\mathbf{x}}_0 + \underline{\mathbf{b}}_1 \\
\mathbf{D}\underline{\mathbf{x}}_2 - \mathbf{O}\underline{\mathbf{x}}_1 &= \underline{\mathbf{b}}_2 \\
\mathbf{D}\underline{\mathbf{x}}_3 - \mathbf{O}\underline{\mathbf{x}}_2 &= \underline{\mathbf{b}}_3 \\
&\vdots \\
\mathbf{D}\underline{\mathbf{x}}_{16} - \mathbf{O}\underline{\mathbf{x}}_{15} &= \underline{\mathbf{b}}_{16}
\end{aligned} \tag{4.19}$$

For the ease of discussion, initialization data from steady-state $\mathbf{O}\underline{\mathbf{x}}_0$ are assimilated in $\underline{\mathbf{b}}_1$. A large set of the expanded network equations consisting of solutions of the first 16 time-steps can thus be constructed, as is shown in (4.20).

$$\begin{array}{c} \hat{\mathbf{A}} \end{array} \left[\begin{array}{cccccccccccccccc} \mathbf{D} & & & & & & & & & & & & & & & \\ \mathbf{O} & \mathbf{D} & & & & & & & & & & & & & & \\ & \mathbf{O} & \mathbf{D} & & & & & & & & & & & & & \\ & & \mathbf{O} & \mathbf{D} & & & & & & & & & & & & \\ & & & \mathbf{O} & \mathbf{D} & & & & & & & & & & & \\ & & & & \mathbf{O} & \mathbf{D} & & & & & & & & & & \\ & & & & & \mathbf{O} & \mathbf{D} & & & & & & & & & \\ & & & & & & \mathbf{O} & \mathbf{D} & & & & & & & & \\ & & & & & & & \mathbf{O} & \mathbf{D} & & & & & & & \\ & & & & & & & & \mathbf{O} & \mathbf{D} & & & & & & \\ & & & & & & & & & \mathbf{O} & \mathbf{D} & & & & & \\ & & & & & & & & & & \mathbf{O} & \mathbf{D} & & & & \\ & & & & & & & & & & & \mathbf{O} & \mathbf{D} & & & \\ & & & & & & & & & & & & \mathbf{O} & \mathbf{D} & & \\ & & & & & & & & & & & & & \mathbf{O} & \mathbf{D} \end{array} \right] \begin{array}{c} \hat{\mathbf{x}} \\ \underline{\mathbf{x}}_1 \\ \underline{\mathbf{x}}_2 \\ \underline{\mathbf{x}}_3 \\ \underline{\mathbf{x}}_4 \\ \underline{\mathbf{x}}_5 \\ \underline{\mathbf{x}}_6 \\ \underline{\mathbf{x}}_7 \\ \underline{\mathbf{x}}_8 \\ \underline{\mathbf{x}}_9 \\ \underline{\mathbf{x}}_{10} \\ \underline{\mathbf{x}}_{11} \\ \underline{\mathbf{x}}_{12} \\ \underline{\mathbf{x}}_{13} \\ \underline{\mathbf{x}}_{14} \\ \underline{\mathbf{x}}_{15} \\ \underline{\mathbf{x}}_{16} \end{array} = \begin{array}{c} \hat{\mathbf{b}} \\ \underline{\mathbf{b}}_1 \\ \underline{\mathbf{b}}_2 \\ \underline{\mathbf{b}}_3 \\ \underline{\mathbf{b}}_4 \\ \underline{\mathbf{b}}_5 \\ \underline{\mathbf{b}}_6 \\ \underline{\mathbf{b}}_7 \\ \underline{\mathbf{b}}_8 \\ \underline{\mathbf{b}}_9 \\ \underline{\mathbf{b}}_{10} \\ \underline{\mathbf{b}}_{11} \\ \underline{\mathbf{b}}_{12} \\ \underline{\mathbf{b}}_{13} \\ \underline{\mathbf{b}}_{14} \\ \underline{\mathbf{b}}_{15} \\ \underline{\mathbf{b}}_{16} \end{array} \tag{4.20}$$

The network equations given in (4.20) can also be written in a compact manner,

$$\hat{\mathbf{A}}\hat{\mathbf{x}} = \hat{\mathbf{b}} \tag{4.21}$$

Interdependency between the solution of two consecutive time-steps is observed in the network equations given in (4.20). Therefore, the solution of these equations needs to be executed sequentially. Performing a one-time row and column reordering on the entire matrix $\hat{\mathbf{A}}$ by renumbering all odd rows and columns ahead of all even rows and columns and applying an LU factorization technique as in (4.22) and (4.23),

$$\hat{\underline{\mathbf{A}}} \underline{\hat{\mathbf{x}}} = \underline{\mathbf{L}} \underline{\mathbf{U}} \underline{\hat{\mathbf{x}}} = \underline{\mathbf{L}} \left[\overbrace{\underline{\mathbf{U}} \underline{\hat{\mathbf{x}}}}^{\underline{\hat{\mathbf{y}}}} \right] = \underline{\mathbf{L}} \underline{\hat{\mathbf{y}}} = \underline{\hat{\mathbf{b}}} \quad (4.22)$$

$$\underline{\mathbf{U}} \underline{\hat{\mathbf{x}}} = \underline{\hat{\mathbf{y}}} \quad (4.23)$$

the lower triangular equations in the L side for forward substitution to solve for $\underline{\hat{\mathbf{y}}}$ thus are,

$$\begin{bmatrix} \mathbf{d} & & & & & & & & & & & & & & & & \\ & \mathbf{d} & & & & & & & & & & & & & & & \\ & & \mathbf{d} & & & & & & & & & & & & & & \\ & & & \mathbf{d} & & & & & & & & & & & & & \\ & & & & \mathbf{d} & & & & & & & & & & & & \\ & & & & & \mathbf{d} & & & & & & & & & & & \\ & & & & & & \mathbf{d} & & & & & & & & & & \\ & & & & & & & \mathbf{d} & & & & & & & & & \\ & \mathbf{o} & & & & & & & \mathbf{d} & & & & & & & & \\ & & \mathbf{o} & & & & & & & \mathbf{f} & \mathbf{d} & & & & & & \\ & & & \mathbf{o} & & & & & & & \mathbf{f} & \mathbf{d} & & & & & \\ & & & & \mathbf{o} & & & & & & & \mathbf{f} & \mathbf{d} & & & & \\ & & & & & \mathbf{o} & & & & & & & \mathbf{f} & \mathbf{d} & & & \\ & & & & & & \mathbf{o} & & & & & & & \mathbf{f} & \mathbf{d} & & \\ & & & & & & & \mathbf{o} & & & & & & & \mathbf{f} & \mathbf{d} & \\ & & & & & & & & \mathbf{o} & & & & & & & \mathbf{f} & \mathbf{d} \end{bmatrix} \begin{bmatrix} \underline{\mathbf{y}}_1 \\ \underline{\mathbf{y}}_3 \\ \underline{\mathbf{y}}_5 \\ \underline{\mathbf{y}}_7 \\ \underline{\mathbf{y}}_9 \\ \underline{\mathbf{y}}_{11} \\ \underline{\mathbf{y}}_{13} \\ \underline{\mathbf{y}}_{15} \\ \underline{\mathbf{y}}_2 \\ \underline{\mathbf{y}}_4 \\ \underline{\mathbf{y}}_6 \\ \underline{\mathbf{y}}_8 \\ \underline{\mathbf{y}}_{10} \\ \underline{\mathbf{y}}_{12} \\ \underline{\mathbf{y}}_{14} \\ \underline{\mathbf{y}}_{16} \end{bmatrix} = \begin{bmatrix} \underline{\mathbf{b}}_1 \\ \underline{\mathbf{b}}_3 \\ \underline{\mathbf{b}}_5 \\ \underline{\mathbf{b}}_7 \\ \underline{\mathbf{b}}_9 \\ \underline{\mathbf{b}}_{11} \\ \underline{\mathbf{b}}_{13} \\ \underline{\mathbf{b}}_{15} \\ \underline{\mathbf{b}}_2 \\ \underline{\mathbf{b}}_4 \\ \underline{\mathbf{b}}_6 \\ \underline{\mathbf{b}}_8 \\ \underline{\mathbf{b}}_{10} \\ \underline{\mathbf{b}}_{12} \\ \underline{\mathbf{b}}_{14} \\ \underline{\mathbf{b}}_{16} \end{bmatrix} \quad (4.24)$$

In (4.24), \mathbf{d} , \mathbf{o} and \mathbf{f} represent diagonal, off-diagonal and fill-in blocks respectively. The vectors $[\underline{\mathbf{y}}_1, \underline{\mathbf{y}}_3, \dots, \underline{\mathbf{y}}_{16}]$ and $[\underline{\mathbf{b}}_1, \underline{\mathbf{b}}_3, \dots, \underline{\mathbf{b}}_{16}]$ are vectors $\underline{\hat{\mathbf{y}}}$ and $\underline{\hat{\mathbf{b}}}$ after the one-time row and column reordering. It is noted that \mathbf{d} and \mathbf{o} differ from \mathbf{D} and \mathbf{O} in equations (4.20) in that the former two are the diagonal and off-diagonal elements after LU factorization, and the numerical values of each \mathbf{d} , \mathbf{o} and \mathbf{f} block are not necessarily identical. A remarkable fact is that this one-time row and column reordering process results in fill-ins (represented by \mathbf{f} blocks) where there was previously none in equations (4.20). Nevertheless, the overall computational effort will be lower thanks to parallelism, as will be seen hereinafter.

It is easy to observe that the first 8 equations in the forward substitution procedure have become independent. Therefore, their solutions can be obtained in parallel with the help of 8 parallel processors. However, the remaining 8 equations, due to the interdependency with previously solved solutions reflected by the **o** and **f** blocks, still need to be solved sequentially, as is shown in Figure 4.3.

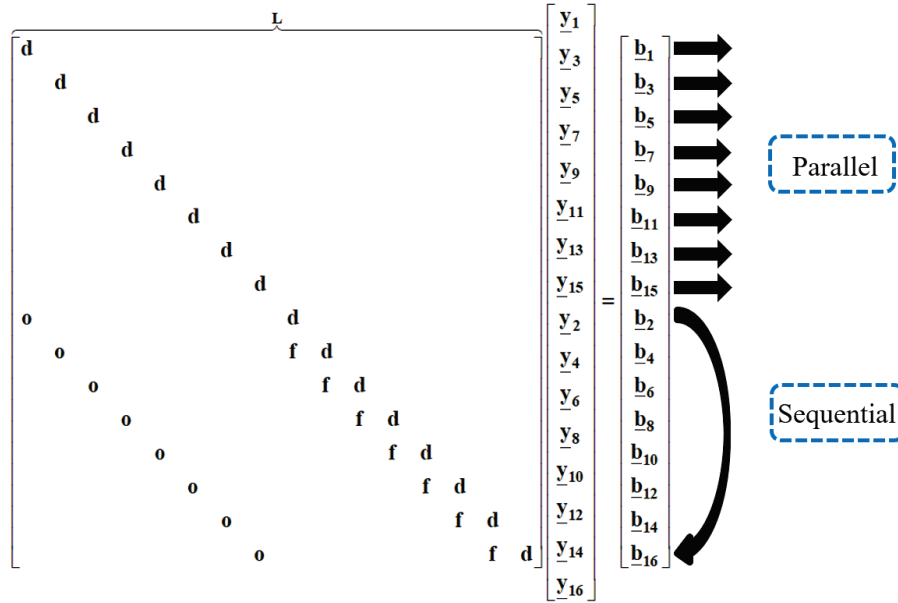


Figure 4.3 Lower triangular matrix L and equations for forward substitution after a one-time row and column reordering and LU factorization

Performing the same row and column reordering scheme recursively first on the entire 16 equations, then on the remaining 8, 4 and 2 equations, after LU factorization, the lower triangular equations for forward substitution are presented in Figure 4.4. It is observed that in the first step using 8 parallel processors, the solutions for y_1 , y_3 , y_5 , y_7 , y_9 , y_{11} , y_{13} and y_{15} can be obtained; in the second step after the solution of the 8 variables in the first step, the solutions for y_2 , y_6 , y_{10} and y_{14} can be obtained; similarly, the solutions for y_4 and y_{12} can be obtained in the third step whereas y_8 and y_{16} are obtained in the fifth and the sixth steps sequentially. The entire forward substitution solution steps are presented in (4.25), in which it is observed that the original 16-step forward substitution procedure can now be achieved in 3 parallel steps and 2 sequential steps (in total 5) with a total of 8 parallel processors.

$$\begin{aligned}
\text{step 1: } & \begin{cases} \underline{d}\underline{y}_1 = \underline{b}_1 \\ \underline{d}\underline{y}_3 = \underline{b}_3 \\ \underline{d}\underline{y}_5 = \underline{b}_5 \\ \underline{d}\underline{y}_7 = \underline{b}_7 \\ \underline{d}\underline{y}_9 = \underline{b}_9 \\ \underline{d}\underline{y}_{11} = \underline{b}_{11} \\ \underline{d}\underline{y}_{13} = \underline{b}_{13} \\ \underline{d}\underline{y}_{15} = \underline{b}_{15} \end{cases} \\
\text{step 2: } & \begin{cases} \underline{d}\underline{y}_2 = \underline{b}_2 - \underline{o}_{21}\underline{y}_1 \\ \underline{d}\underline{y}_6 = \underline{b}_6 - \underline{o}_{63}\underline{y}_3 \\ \underline{d}\underline{y}_{10} = \underline{b}_{10} - \underline{o}_{10,5}\underline{y}_5 \\ \underline{d}\underline{y}_{14} = \underline{b}_{14} - \underline{o}_{14,7}\underline{y}_7 \end{cases} \\
\text{step 3: } & \begin{cases} \underline{d}\underline{y}_4 = \underline{b}_4 - \underline{o}_{43}\underline{y}_3 - \underline{f}_{42}\underline{y}_2 \\ \underline{d}\underline{y}_{12} = \underline{b}_{12} - \underline{o}_{12,11}\underline{y}_{11} - \underline{f}_{12,10}\underline{y}_{10} \end{cases} \\
\text{step 4: } & \underline{d}\underline{y}_8 = \underline{b}_8 - \underline{o}_{87}\underline{y}_7 - \underline{f}_{86}\underline{y}_6 - \underline{f}_{84}\underline{y}_4 \\
\text{step 5: } & \underline{d}\underline{y}_{16} = \underline{b}_{16} - \underline{o}_{16,15}\underline{y}_{15} - \underline{f}_{16,14}\underline{y}_{14} - \underline{f}_{16,12}\underline{y}_{12} - \underline{f}_{16,8}\underline{y}_8
\end{aligned} \tag{4.25}$$

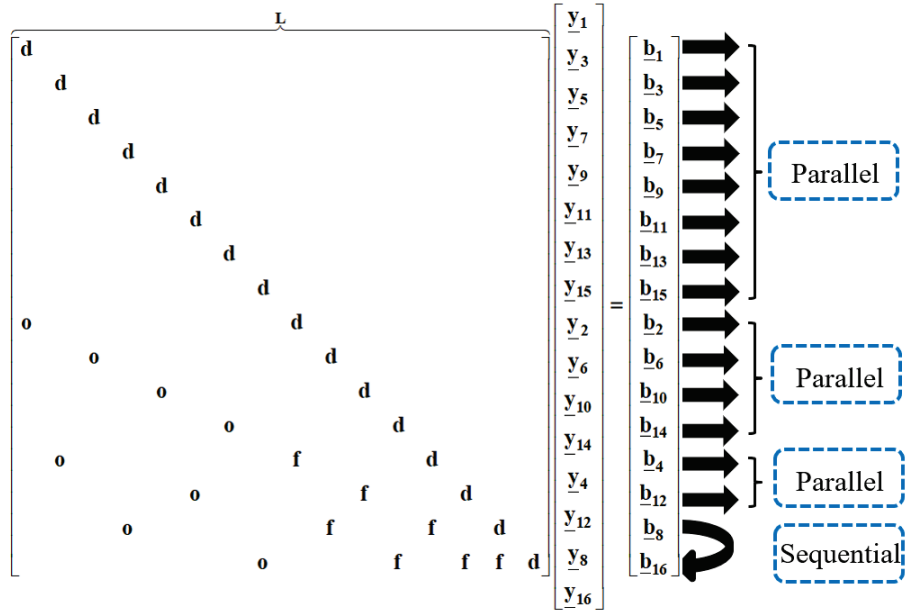


Figure 4.4 Lower triangular matrix L and equations for forward substitution after recursive row and column reordering and LU factorization

Similarly, the upper triangular \mathbf{U} matrix and equations for backward substitution are shown in (4.26), where \mathbf{I} , \mathbf{o} and \mathbf{f} represent identity, off-diagonal and fill-in blocks respectively. The entire backward substitution steps are presented in (4.27). Therefore, the original 16-step backward substitution procedure can be achieved in 4 steps, using 8 parallel processors.

In summary, using the original step-by-step methodology, the total number of steps in the forward and backward substitutions is 32 for the solutions of 16 time-steps. However, applying the equation grouping and recursive row and column reordering scheme, the total number of steps in the forward and backward substitutions is reduced to 9 (5 steps in the forward substitution and 4 steps in the backward substitution). Without topological changes in the network or iterations, the \mathbf{L} and \mathbf{U} matrices remain unchanged. Therefore, major computation time gains can be expected in a time-domain simulation scenario by grouping the MANA-based expanded network equations of every 16 time-steps together, as in (4.20), and solving them with the help of the recursive row and column reordering scheme.

The equation grouping and recursive row and column reordering scheme can be applied on the solutions of every 32, 64, 128, ... time-steps in the exact same fashion. Higher computation time gains are expected. Once again, the number of solutions to be grouped depends on the available parallel processors in order to achieve maximal efficiency brought by the equation grouping and recursive row and column reordering scheme (8, 16, 32, ... processors are needed to group the solutions of 16, 32, 64, ...time-steps).

[illegible]

$$\begin{aligned}
&\text{step 1: } \begin{cases} \underline{\mathbf{I}}_{\underline{\mathbf{x}}_{16}} = \underline{\mathbf{y}}_{16} \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_8} = \underline{\mathbf{y}}_8 \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_4} = \underline{\mathbf{y}}_4 \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_2} = \underline{\mathbf{y}}_2 \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_1} = \underline{\mathbf{y}}_1 \end{cases} \\
&\text{step 2: } \begin{cases} \underline{\mathbf{I}}_{\underline{\mathbf{x}}_{12}} = \underline{\mathbf{y}}_{12} - \underline{\mathbf{f}}_{12,8} \underline{\mathbf{x}}_8 \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_{10}} = \underline{\mathbf{y}}_{10} - \underline{\mathbf{f}}_{10,8} \underline{\mathbf{x}}_8 \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_6} = \underline{\mathbf{y}}_6 - \underline{\mathbf{f}}_{64} \underline{\mathbf{x}}_4 \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_3} = \underline{\mathbf{y}}_3 - \underline{\mathbf{f}}_{32} \underline{\mathbf{x}}_2 \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_9} = \underline{\mathbf{y}}_9 - \underline{\mathbf{f}}_{9,8} \underline{\mathbf{x}}_8 \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_5} = \underline{\mathbf{y}}_5 - \underline{\mathbf{f}}_{54} \underline{\mathbf{x}}_4 \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_3} = \underline{\mathbf{y}}_3 - \underline{\mathbf{f}}_{32} \underline{\mathbf{x}}_2 \end{cases} \\
&\text{step 3: } \begin{cases} \underline{\mathbf{I}}_{\underline{\mathbf{x}}_{14}} = \underline{\mathbf{y}}_{14} - \underline{\mathbf{f}}_{14,12} \underline{\mathbf{x}}_{12} \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_{13}} = \underline{\mathbf{y}}_{13} - \underline{\mathbf{f}}_{13,12} \underline{\mathbf{x}}_{12} \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_{11}} = \underline{\mathbf{y}}_{11} - \underline{\mathbf{f}}_{11,10} \underline{\mathbf{x}}_{10} \\ \underline{\mathbf{I}}_{\underline{\mathbf{x}}_7} = \underline{\mathbf{y}}_7 - \underline{\mathbf{f}}_{76} \underline{\mathbf{x}}_6 \end{cases} \\
&\text{step 4: } \underline{\mathbf{I}}_{\underline{\mathbf{x}}_{15}} = \underline{\mathbf{y}}_{15} - \underline{\mathbf{f}}_{15,14} \underline{\mathbf{x}}_{14}
\end{aligned} \tag{4.27}$$

4.2 Implementation of the PEGR-based parallel-in-time approach

This section tackles issues encountered in the implementation of the PEGR-based parallel-in-time approach. It first elucidates the platform of implementation, which is C++ using API OpenMP Multithreading. Then it proceeds to elaborate on measures taken to minimize false sharing, data race as well as inconsistency between temporary view and memory in a parallel computing environment. An important procedure in this parallel-in-time approach is the extraction of off-diagonal block elements from the \mathbf{L} and \mathbf{U} factors after factorization for forward and backward substitution, whose details are presented. Last but not least, since a discontinuity caused by switching or any other topological change in the network can happen at an arbitrary time-point and a time-domain simulation can consist of any number of simulation points whereas the PEGR-based parallel-in-time approach solves the network equations for every 16, 32, 64, 128, ... time-steps, the treatment of arbitrary points of topological changes and numbers of simulation points is discussed.

4.2.1 Parallelization using OpenMP

It is noted the PEGR-based parallel-in-time approach proposed in this thesis is formulated using sparse matrix techniques [97], which indicates that all the formulated matrices are represented as sparse matrices, and all algorithms used in matrix manipulations cater to sparse matrices. Furthermore, the KLU sparse linear solver [11]-[13] is adopted as the sparse matrix solver. The entire algorithm is programmed in C++ language using API OpenMP Multithreading [98] for parallelization during the forward and backward substitution processes. A snippet of the parallelization process is shown as follows:

```
omp_set_dynamic(0);
start = std::chrono::system_clock::now();

// parallelization
for (int itr = 0; itr < tot_time_points / 64; itr++)
{
#pragma omp parallel num_threads(32)
{
#pragma omp sections
{
////////// thread 1
#pragma omp section
{
if (itr > 0)
{
sp_solve_thread0.init_first_element(sp_matL, b, itr);
}
sp_solve_thread0.solve_for_z(b, z_for_x, itr, 0);
sp_solve_thread0.mat_vec_mult_subt_for_z(L1_p, L1_i, L1_x,
b, z_for_x, itr, 0, 32 * size_sub_mat);
sp_solve_thread0.solve_for_z(b, x, itr, 32 * size_sub_mat);
sp_solve_thread0.mat_vec_mult_subt_for_z(L2_p, L2_i, L2_x,
b, z_for_x, itr, 32 * size_sub_mat, 48 * size_sub_mat);
sp_solve_thread0.solve_for_z(b, z_for_x, itr, 48 *
size_sub_mat);
sp_solve_thread0.mat_vec_mult_subt_for_z(L3_p, L3_i, L3_x,
b, z_for_x, itr, 48 * size_sub_mat, 56 * size_sub_mat);
sp_solve_thread0.solve_for_z(b, z_for_x, itr, 56 *
size_sub_mat);
sp_solve_thread0.mat_vec_mult_subt_for_z(L4_p, L4_i, L4_x,
b, z_for_x, itr, 56 * size_sub_mat, 60 * size_sub_mat);
sp_solve_thread0.solve_for_z(b, z_for_x, itr, 60 *
size_sub_mat);
sp_solve_thread0.mat_vec_mult_subt_for_z(L5_p, L5_i, L5_x,
b, z_for_x, itr, 60 * size_sub_mat, 62 * size_sub_mat);
sp_solve_thread0.solve_for_z(b, z_for_x, itr, 62 *
size_sub_mat);
sp_solve_thread0.mat_vec_mult_subt_for_z(L6_p, L6_i, L6_x,
b, z_for_x, itr, 62 * size_sub_mat, 63 * size_sub_mat);
sp_solve_thread0.solve_for_z(b, z_for_x, itr, 63 *
size_sub_mat);
}
```

```

                                sp_solve_thread0.solve_for_x(x, z_for_x, itr, 63 *
size_sub_mat);
                                }
                                // thread 2
#pragma omp section
                                {
                                    sp_solve_thread1.solve_for_z(b, z_for_x, itr, size_sub_mat);
                                    sp_solve_thread1.mat_vec_mult_subt_for_z(L1_p, L1_i, L1_x,
b, z_for_x, itr, size_sub_mat, 48 * size_sub_mat);
                                    sp_solve_thread1.solve_for_x(x, z_for_x, itr, 62 *
size_sub_mat);
                                    sp_solve_thread1.mat_vec_mult_subt_for_x(U2_p, U2_i, U2_x,
b, x, z_for_x, itr, 62 * size_sub_mat, 40 * size_sub_mat);
                                    sp_solve_thread1.solve_for_x(x, z_for_x, itr, 40 *
size_sub_mat);
                                    sp_solve_thread1.mat_vec_mult_subt_for_x(U1_p, U1_i, U1_x,
b, x, z_for_x, itr, 40 * size_sub_mat, 17 * size_sub_mat);
                                    sp_solve_thread1.solve_for_x(x, z_for_x, itr, 17 *
size_sub_mat);
                                }

```

The **parallel** construct is the fundamental construct in OpenMP to start parallel execution. When execution encounters a **parallel** directive, the number of threads to use in this region can be determined by the value indicated in the **num_threads** clause. In the code snippet shown above, this value is set to 32, which means a team of 32 threads is used in the parallel region (31 new threads are created, with the thread encountering the **parallel** construct being the master thread of the team). If 32 threads are used in the parallelization of forward and backward substitutions, the algorithm solves the network equations for every 64 time-steps. Before explicitly setting the number of threads in the subsequent parallel region to 32, dynamic thread setting adjusted by the run time is deactivated.

In the PEGR-based parallel-in-time approach, for the solution of every T time-steps ($T = 2^\tau$, τ is an integer), several forward and backward substitution procedures can be grouped together and performed in parallel. Based on this particular nature of the approach, the **sections** construct is adopted. The **sections** construct is a non-iterative worksharing construct that contains a set of structured blocks that are to be distributed among and executed by the threads in a team. Each independent structured block is executed once by one of the threads in the team, using the syntax **#pragma omp section** as is shown in the above code snippet. During the entire solution process, the whole parallelization region is encapsulated within a for loop that executes n iterations, with n defined by,

$$n = \frac{\text{Total number of time-steps}}{T} \quad (4.41)$$

where T is the number of time-steps the algorithm solves in a single iteration using a total number of $T/2$ threads. It is to be modified depending on the actual number of threads used in the algorithm. Moreover, the discussion here solely aims to shed light on the realization of parallelization in the algorithm for the PEGR-based parallel-in-time approach without considering discontinuity or topological changes in the network. More details on the constructs and clauses used in the algorithm, as presented in the code snippet, can be found in [98] and [99].

4.2.2 Measures against false sharing, data race and inconsistency between temporary view and memory

4.2.2.1 Minimizing false sharing

False sharing occurs when processors in a shared-memory parallel system make references to different data objects within the same coherence block (cache line or page), thereby inducing “unnecessary” coherence operations. Specifically, a thread attempts to periodically access data that is not altered by another party but share the same cache block with data that are altered. The caching protocol therefore forces the thread to reload the whole unit despite a lack of logical necessity. It is widely believed to be a performance-degrading usage pattern for parallel programs [100]. Several techniques are employed in this algorithm in order to minimize false sharing. The goal is to ensure that variables causing false sharing are spaced far enough apart in memory that they cannot reside on the same cache line.

- **`__declspec (align(n))`**

This directive can force alignment of individual variables on the cache line boundaries. In modern-day CPUs, the size of the cache line is usually 64 bytes. Therefore, the alignment value is set to 64, as is shown in the following code snippet:

```
__declspec(align(64)) int tot_time_points = 8192;
__declspec(align(64)) int size_sub_mat = 50;
__declspec(align(64)) double dt = 0.00005;
```

- **`_aligned_malloc()`**

This function allows dynamic allocation of memory on a specific alignment boundary, providing similar functionalities as the previous one, as is presented in the following code snippet:

```
D1 = (double**) _aligned_malloc(sizeof(double*)*size_sub_mat, 64);
D2 = (double**) _aligned_malloc(sizeof(double*)*size_sub_mat, 64);
D3 = (double**) _aligned_malloc(sizeof(double*)*size_sub_mat, 64);
```

4.2.2.2 Minimizing data race

Different threads concurrently accessing data stored on the same cache line can potentially cause data corruption leading to irreproducible incorrect results, which is another common problem in parallel programming named “data race”. Due to the algorithm complexity and data sharing nature of the PEGR-based parallel-in-time approach between threads, the following measure has been effectuated in order to minimize data race problems.

- **Thread-local parallel solution objects**

Equations (4.25) and (4.27) show that during the parallel solution process, solutions at certain time-points depend on those at previous time-points calculated on different threads, which requires different threads reading from or writing into shared memory concurrently. Therefore, the goal is to make the object of the solver class (a class in the algorithm that contains functions of factorization, forward and backward substitution) thread-local, namely, to instantiate an independent solver class object for each thread in order to construct independent **L** and **U** factors and related solution functions. These independent objects are used exclusively in their respective threads, greatly reducing the contention caused by the data race problem from using the same solver class object in the solution process.

4.2.2.3 Enforcing consistency between the temporary view and memory

In OpenMP, a thread’s temporary view of memory is not required to be consistent with memory at all times. A value written to a variable can remain in the thread’s temporary view until it is forced to memory at a later time. Likewise, a read from a variable may retrieve the value from the thread’s temporary view, unless it is forced to read from memory. The OpenMP flush operation enforces consistency between the temporary view and memory [98].

In the algorithm, due to the need of constantly exchanging shared information between different threads, a series of variables named “flags” are utilized to indicate the completion of writing into a set of shared variables and the authorization of the start of reading from a set of shared variables. The “flag” variables are used together with the **flush** construct which makes a thread’s temporary view of memory consistent with memory and enforces an order on the memory operations of the variables explicitly specified or implied [98]. The **flush** construct is a low-level synchronization scheme that is used to impose order constraints and to protect access to shared data. The operations performed in such circumstances are presented in the following code snippet:

```

        z1ready = true;
#pragma omp flush(z1ready)
        sp_solve_thread2.mat_vec_mult_subt_for_z(L1_p, L1_i, L1_x, b, z_for_x,
itr, 0, 8 * size_sub_mat);
        sp_solve_thread2.solve_for_z(b, x, itr, 8 * size_sub_mat);
        while (flag_thread_1 != true)
        {
#pragma omp flush(flag_thread_1)
        }
        sp_solve_thread2.mat_vec_mult_subt_for_z(L2_p, L2_i, L2_x, b, z_for_x,
itr, 8 * size_sub_mat, 12 * size_sub_mat);
        sp_solve_thread2.solve_for_z(b, z_for_x, itr, 12 * size_sub_mat);

```

In the above code snippet, the variable “z1ready” with value “true” is “flushed” into memory indicating the previous task is complete. After the execution of the subsequent 2 operations using functions “sp_solve_thread2.mat_vec_mult_subt_for_z” and “sp_solve_thread2.solve_for_z”, the value of the variable “flag_thread_1” keeps being checked in a loop, and the following operations will not be executed until its value becomes “true”.

It is noted the use of the **flush** construct is restricted in the algorithm in that abusing it will severely affect the algorithmic efficiency.

4.2.3 Block element extraction for forward and backward substitutions

After factorization of the expanded and grouped network equations of multiple time-steps using the KLU sparse linear solver, certain off-diagonal block elements need to be extracted from the **L** and **U** factors to be used in the forward and backward substitutions. The structure of the **L** and **U** factors for the case of grouping the network solutions of 16 time-steps is presented in Figure 4.5. It is noted that the block elements with the same subscript in each factor is identical, and the row and column numbers indicate the numbers of the rows and columns in the symbolic matrix shown

in Figure 4.5 but not the actual row and column numbers since the size of the actual expanded and grouped matrix in reality would be much larger than 16×16 . Therefore, for the case of grouping the network solutions of 16 time-steps, the \mathbf{L} factor has 4 distinctive block elements and the \mathbf{U} factor consists of 3 distinctive block elements.

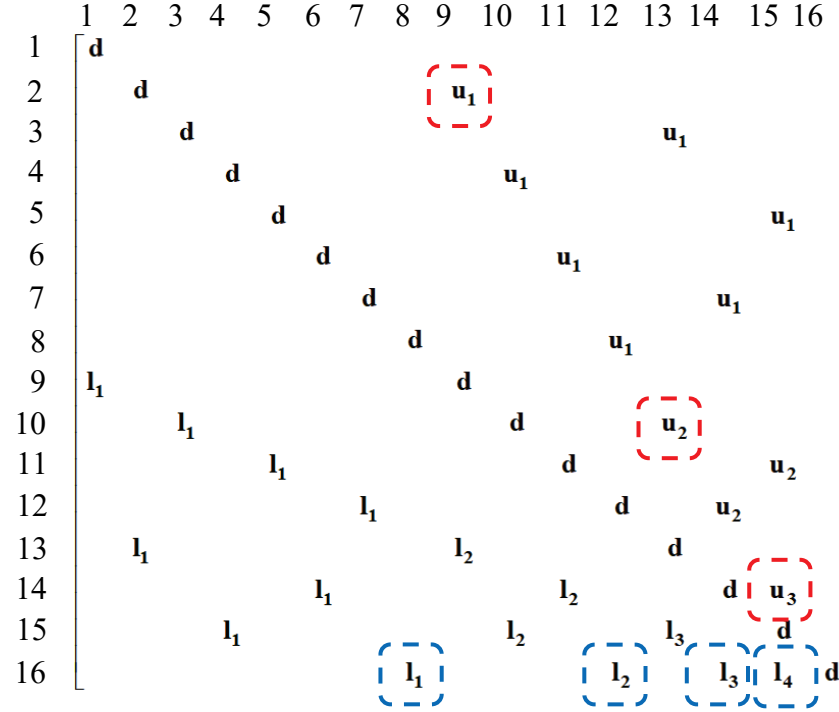


Figure 4.5 : Structure of the \mathbf{L} and \mathbf{U} factors when grouping the network equations of 16 time-steps

It is worth mentioning that the non-zeros in the \mathbf{L} and \mathbf{U} factors after factorization using KLU are stored in a column-wise manner from left to right. For the case of grouping the network solutions of 16 time-steps, it is the elements circled by rounded-cornered rectangles in Figure 4.5 that are extracted to be used in further forward and backward substitutions (blue rounded-cornered rectangles indicate extracted block elements in the \mathbf{L} factor, and red rounded-cornered rectangles denote extracted block elements in the \mathbf{U} factor). In the extraction of block elements, both factors are traversed in the column major order from left to right. For the example shown in Figure 4.5, the extraction of block elements in the \mathbf{L} and \mathbf{U} factors are explained as follows:

For the extraction of block elements in the **L** factor:

- 1) When the traverse hits symbolic row 16, extraction of \mathbf{l}_1 starts
- 2) When the traverse hits symbolic row 9 (the diagonal block element in the next symbolic column), mark the completion of \mathbf{l}_1 extraction
- 3) When the traverse hits symbolic row 16 and \mathbf{l}_1 extraction is completed, extraction of \mathbf{l}_2 starts
- 4) When the traverse hits symbolic row 13 (the diagonal block element in the next symbolic column), mark the completion of \mathbf{l}_2 extraction
- 5) When the traverse hits symbolic row 16 and \mathbf{l}_2 extraction is completed, extraction of \mathbf{l}_3 starts
- 6) When the traverse hits symbolic row 15 (the diagonal block element in the next symbolic column), mark the completion of \mathbf{l}_3 extraction
- 7) When the traverse hits symbolic row 15, \mathbf{l}_3 extraction is completed and the traverse element count $i \leq nz_L - nz_{diag} - 1$, extraction of \mathbf{l}_4 starts

For the extraction of block elements in the **U** factor:

- 1) When the traverse hits symbolic row 8 (the diagonal element in the 8th symbolic row), mark the extraction of \mathbf{u}_1 ready (it is the first block element in the next symbolic column)
- 2) When the traverse hits symbolic row 2 and the extraction of \mathbf{u}_1 is ready, extraction of \mathbf{u}_1 starts
- 3) When the traverse hits symbolic row 4 (the first block element in the next symbolic column), mark the completion of \mathbf{u}_1 extraction

- 4) When the traverse hits symbolic row 12 (the diagonal element in the 12th symbolic row), mark the extraction of \mathbf{u}_2 ready
- 5) When the traverse hits symbolic row 10 (the second block element in this symbolic column) and the extraction of \mathbf{u}_2 is ready, extraction of \mathbf{u}_2 starts
- 6) When the traverse hits symbolic row 7 (the first block element in the next symbolic column), mark the completion of \mathbf{u}_2 extraction
- 7) When the traverse hits symbolic row 5 (the 15th symbolic column), mark the extraction of \mathbf{u}_3 ready
- 8) When the traverse hits symbolic row 14 and the extraction of \mathbf{u}_3 is ready, extraction of \mathbf{u}_3 starts
- 9) When the traverse hits symbolic row 16, mark the completion of \mathbf{u}_3 extraction

It is noted in step 7) of the extraction of block elements in the \mathbf{L} factor, nz_L represents the total number of non-zeros in the \mathbf{L} factor, and nz_{diag} denotes the total number of non-zeros in the diagonal block element. Hence, the condition $i \leq nz_L - nz_{diag} - 1$ guarantees that it is the block element \mathbf{l}_4 being extracted, not the last diagonal block element which is also in the 16th symbolic row. The generalized algorithm for the extraction of block elements in the \mathbf{L} and \mathbf{U} factors for cases where network equations of different number of solution time-steps are grouped is given in Appendix C.

4.2.4 Treatment of arbitrary points of topological changes and numbers of simulation points

Since the PEGR-based parallel-in-time approach groups the expanded network equations of every 16, 32, 64, 128, ... time-steps and solves them together, network topological change time-points need to be known before time-domain simulations and a conventional sequential step-wise approach also employing the KLU solver to solve the original network MANA equations is hence

incorporated in the algorithm to handle topological changes that happen at arbitrary time-points. Using the grouping of solutions at 16 time-steps as an example, as is illustrated in Figure 4.6, as the simulation proceeds, the PEGR-based parallel-in-time approach solves equations between $t = T$ and $t = T + 16\Delta t$ together (by grouping the expanded network equations of 16 time-steps). Nonetheless, a topological change happens at $t = T + n\Delta t$ which is before $t = T + 32\Delta t$ (during the solution of the next 16 time-steps). Therefore, the PEGR-based parallel-in-time approach is no longer applicable here. The algorithm therefore switches to the conventional sequential step-wise approach at $t = T + 16\Delta t$ to solve the time-points between $t = T + 16\Delta t$ and $t = T + n\Delta t$, and switches back to the PEGR-based approach afterwards.

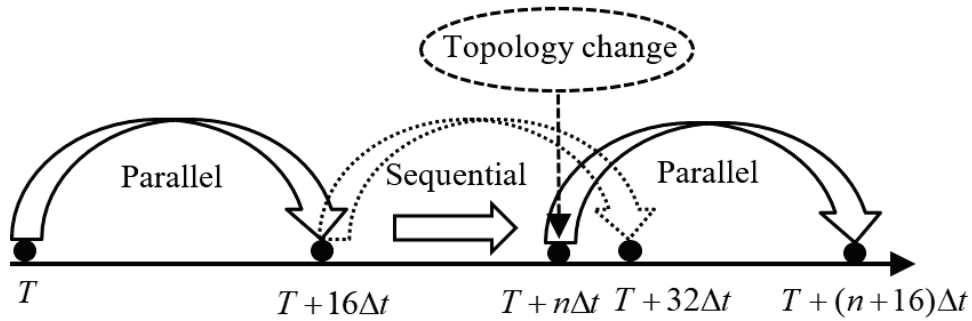


Figure 4.6 Switching between the PEGR-based approach and the conventional approach in the case of topological changes in the network

The complete algorithm for time-domain simulations is summarized as follows:

1) Initialization

- a) Analysis of topological change time-points
- b) Generation of expanded network equations $\mathbf{D}\underline{\mathbf{x}}_t - \mathbf{O}\underline{\mathbf{x}}_{t-\Delta t} = \underline{\mathbf{b}}_t$ from the original network MANA equations
- c) Grouping the expanded network equations to construct $\hat{\mathbf{A}}\hat{\mathbf{x}} = \hat{\mathbf{b}}$, based on the user-chosen number of threads to be launched
- d) Recursive block-wise row and column reordering on $\hat{\mathbf{A}}$

2) Factorization

- a) Symbolic and numeric factorization of the reordered matrix $\hat{\mathbf{A}}$ using KLU
 - b) \mathbf{L} and \mathbf{U} factor block element extraction from the results of numeric factorization for the use in forward and backward substitutions in the PEGR-based parallel-in-time approach
 - c) Symbolic and numeric factorization of the original network MANA coefficient matrix using KLU
- 3) Solution
- a) Solution using the PEGR-based parallel-in-time approach for every n time-steps (n is twice the user-chosen number of launched threads)
 - b) Sequential step-wise solution based on the conventional approach before reaching points of topological changes or end of simulation
 - c) If it is the end of the simulation, go to step 4); if not, modify and update the \mathbf{D} and \mathbf{O} blocks in the reordered matrix as well as the original network MANA coefficient matrix, then go back to step 2)
- 4) End of time-domain simulation

4.3 Conclusion

This chapter presented the development of a new parallel-in-time approach implementing the parallel-in-time equation grouping and reordering (PEGR) scheme for power system EMT simulations. It first discussed the adaption of the PEGR technique originally formulated in state-space to the modified-augmented-nodal analysis (MANA) through the expansion of the original network equations and explicitly manifesting interdependency between consecutive solution time-steps. It then demonstrated the PEGR technique used in the new parallel-in-time approach as well as the original state-space proposition through an example in which the solutions of 16 time-steps are grouped. Subsequently, detailed programming issues encountered in the implementation of the PEGR-based parallel-in-time approach in C++ were discussed, which include the OpenMP platform for parallelization, particular measures and programming considerations to minimize false sharing, data race as well as maintaining consistency between the temporary view and memory in

OpenMP, block elements extraction from the \mathbf{L} and \mathbf{U} factors after factorization to be used in the forward and backward substitutions, and the treatment of arbitrary points of topological changes and numbers of simulation points.

CHAPTER 5 TEST CASES AND SIMULATION RESULTS OF THE PEGR-BASED PARALLEL-IN-TIME APPROACH

This chapter presents simulation results of the PEGR-based parallel-in-time approach developed in this thesis. As in the FMI-based co-simulation approach proposed in early chapters of this thesis, the tests on the PEGR-based parallel-in-time approach aim to validate this approach for accuracy control and solution speedup in power system time-domain simulations. New power system simulation benchmarks are used for the validation of each aspect. The validation of accuracy is based on the comparison of various waveforms using the PEGR-based parallel-in-time approach with those obtained from an EMT-type solver [2], and solution speedup is validated by comparing solution times using the proposed approach to those using the conventional step-wise solution scheme. The advantages of the proposed PEGR-based parallel-in-time approach for large-scale power system time-domain simulations are demonstrated, and a discussion on the obtained results is presented. All tests are executed on the PC with 24 cores and 48 logical processors whose hardware and software configurations are given in Table 3.18. In all test cases, a simulation interval of 3 s is chosen, and a phase-a-to-ground fault scenario occurring at $t = 2s$ and cleared at $t = 2.1s$ is simulated in time-domain.

5.1 Accuracy validation

In this section, the accuracy of the PEGR-based parallel-in-time approach proposed in this thesis is validated by examining voltages on various buses and the fault current during a fault scenario. The reference waveforms are those found from an EMT-type solver [2] with ultimate accuracy. The Network-5 benchmark, as is shown in Figure 5.1, is used in the test, with the fault location clearly indicated. It is based on the original IEEE-34 bus test system [101], [102]. In this benchmark, the generator is represented by a three-phase voltage source (AC at bus 800); the transformers and regulators (T832_888, TAP852_832 and TAP814_850) are modelled by three-phase two coupled windings; all the transmission lines are modelled by coupled pi-sections; and the loads (both spot loads and distributed loads) are represented by three-phase resistors, inductors and capacitors. Therefore, the network illustrated in Figure 5.1 is completely linear.

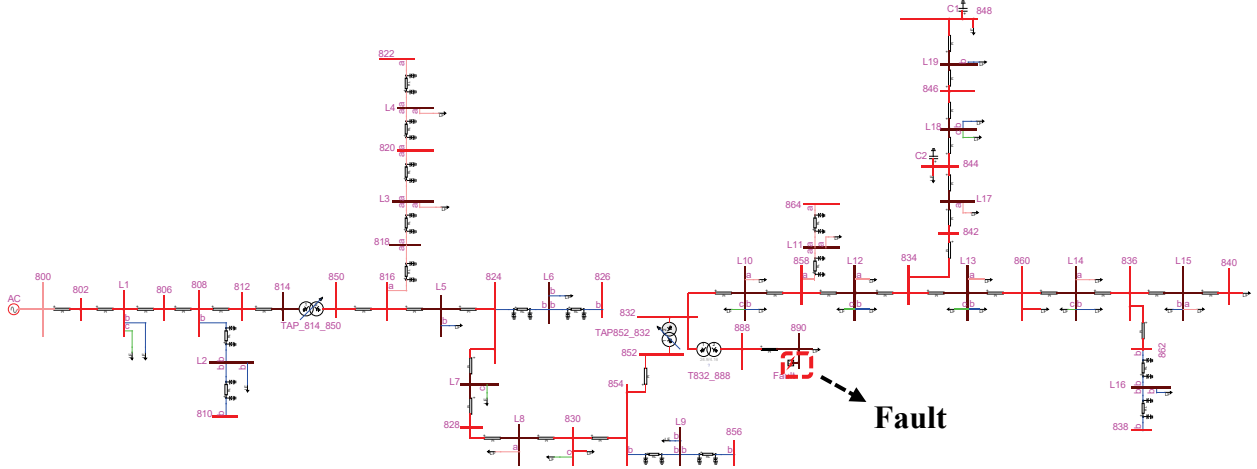
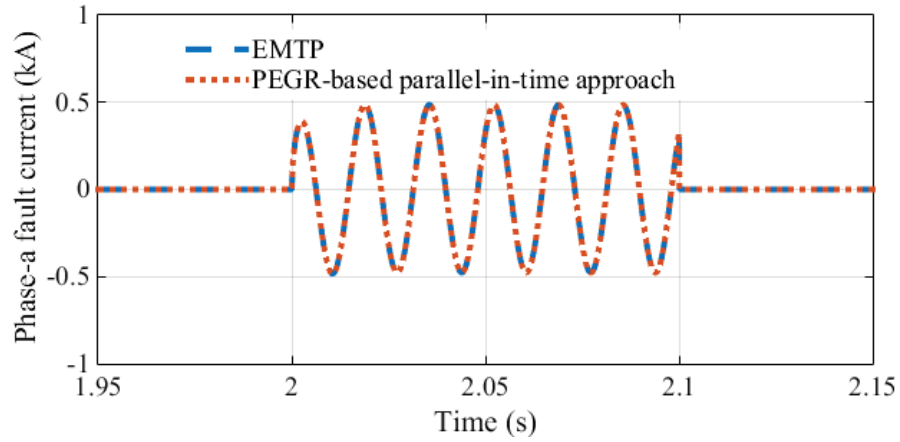


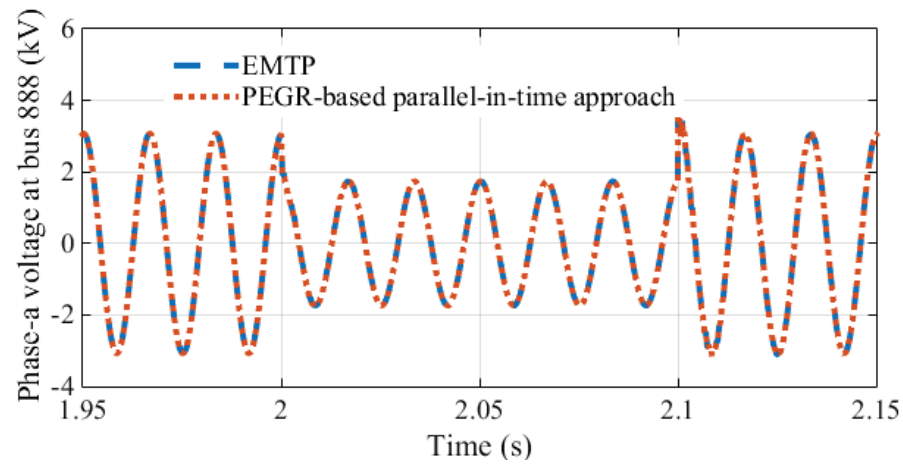
Figure 5.1 Network-5 benchmark

In this test, a numerical integration time-step of $\Delta t = 50\mu s$ is used for the time-domain simulation, and the phase-a-to-ground fault occurs at bus 890. The tested PEGR-based parallel-in-time approach groups the expanded network equations at 16 time-points with the help of 8 parallel threads. This is adequate for the test of accuracy control of the PEGR-based approach in that the formulation of the expanded network equations before grouping and reordering remains the same despite the number of launched threads and the number of solution time-points at which the expanded network equations are to be grouped.

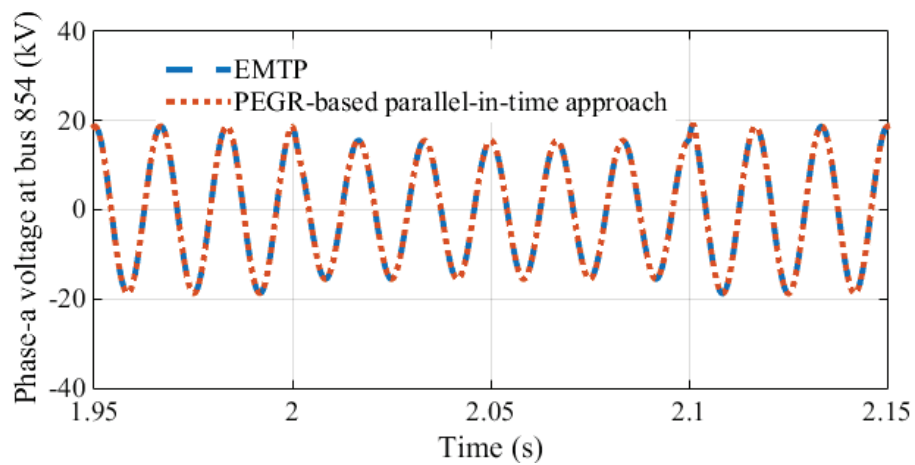
The waveforms of the phase-a fault current, as well as phase-a voltages at buses 888 and 854 during the fault, are presented in Figure 5.2. For clear observation of the waveforms, only the interval between $t = 1.95s$ and $t = 2.15s$ is shown in all plots. It can be observed that the various waveforms obtained from the proposed PEGR-based parallel-in-time approach are identical to those from the EMT-type solver [2], demonstrating that time-domain simulations using the proposed approach remain accurate.



(a) Phase-a fault current



(b) Phase-a voltage at bus 888



(c) Phase-a voltage at bus 854

Figure 5.2 Phase-a fault current, phase-a voltages at buses 888 and 854

5.2 Computation time gains

The validation of computation time gains is based on the comparison of the solution times of the benchmarks simulated with the PEGR-based parallel-in-time approach proposed in this thesis and using the conventional sequential step-wise scheme. Its objective is to demonstrate the numerical performance advantages of the proposed parallel-in-time approach on power system networks of different levels of complexity. The 24-core PC, whose hardware and software configurations are given in Table 3.18, is used in the tests of computation time gains.

It is worth noting that the purpose here is not to compare the performance of the PEGR-based parallel-in-time approach developed in this thesis with that of any highly optimized commercial power system simulation packages, but to demonstrate the numerical advantages of the new methodology brought by grouping and recursive row and column reordering of expanded network equations over the conventional sequential step-wise solution scheme. Hence, the solution times obtained from the PEGR-based parallel-in-time approach are compared with a conventional sequential step-wise solution scheme also implemented in C++ and employing the KLU sparse linear solver as the matrix solver.

5.2.1 Test case 1

The first test case is based on the Network-5 benchmark shown in Figure 5.1. It is performed on the original Network-5 benchmark which is a 34-bus network as well as on a 170-bus network constructed by replicating the Network-5 benchmark in Figure 5.1 five times. Similarly, all transmission lines in the 170-bus network are modelled by coupled pi-sections, and all loads are modelled by three-phase resistors, inductors and capacitors. A simulation interval of 3 s is chosen, with a numerical integration time-step of 50 μ s. In all tests, a phase-a-to-ground fault occurring at $t = 2$ s and cleared at $t = 2.1$ s is simulated in time-domain. It is noted that different numbers of threads 8, 16 and 32 are used in the tests with the PEGR-based parallel-in-time approach on the 24-core PC (Table 3.18), corresponding to the numbers of cores used 4, 8 and 16 respectively. The expanded network equations are thus grouped for every 16, 32, and 64 time-steps. The performance timings of both the conventional sequential step-wise approach and the PEGR-based parallel-in-time approach are presented in Table 5.1, in which solution times refer to the time consumed in the factorization of network equations and their solutions throughout the entire simulation interval in

time-domain for the conventional sequential step-wise approach, and to the time consumed in initialization, factorization and solution, as shown in Section 4.2.4, for the PEGR-based parallel-in-time approach. It is noted that the number of launched threads for the conventional step-wise approach is marked as “N/A”, indicating that this approach does not involve parallelization.

Table 5.1 Performance timings of test case 1

	Step-wise approach	PEGR-based approach		
Number of launched threads	N/A	8	16	32
34-bus network solution times (s)	83.72	50.43	39.03	28.20
170-bus network solution times (s)	444.53	233.16	170.45	125.53

Based on the data presented in Table 5.1, the solution time speedup using the PEGR-based approach compared to the conventional sequential step-wise approach with respect to the number of launched threads in the simulation can be calculated and is shown in Figure 5.3 for both test networks.

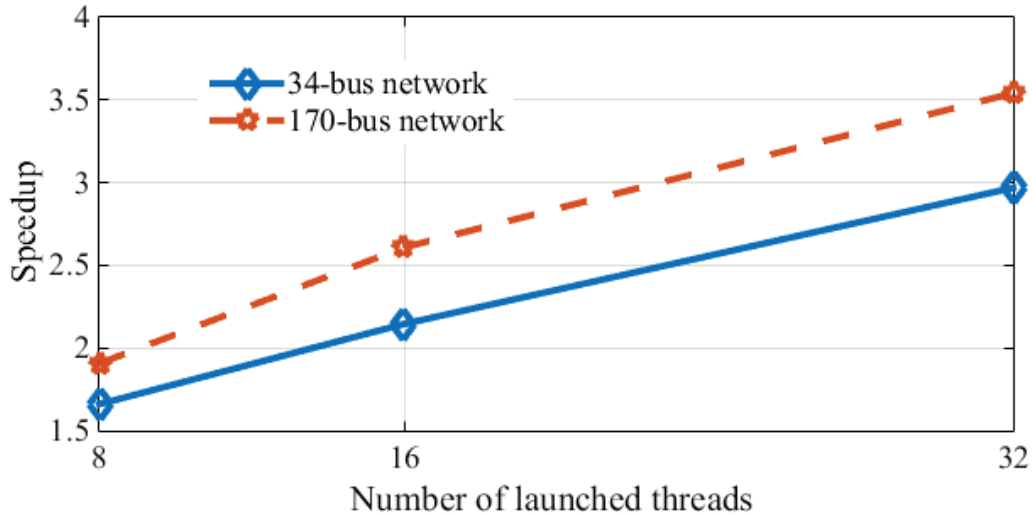


Figure 5.3 Solution time speedup using the PEGR-based parallel-in-time approach with respect to the number of launched threads in the simulation for test case 1

5.2.2 Test case 2

The second test case is based on the Network-6 benchmark shown in Figure 5.4. It is the EMT version of the original IEEE-14 test system. The parameters for this benchmark can be found in [103]. Similar to the first test case presented in the previous section, in this test, the generators are

represented by three-phase voltage sources; the transformers and regulators are modeled by three-phase two coupled windings; all the transmission lines are modeled using coupled pi-sections; and the loads are represented by three-phase resistors, inductors and capacitors. By replicating the Network-6 benchmark 5, 10, 15 and 20 times, larger systems of 70, 140, 210 and 280 buses can therefore be constructed.

Once again, different numbers of threads 8, 16 and 32 are used in the tests using the PEGR-based parallel-in-time approach on the 24-core PC (Table 3.18), corresponding to the numbers of cores used 4, 8 and 16 respectively. The expanded network equations are thus grouped for every 16, 32, and 64 time-steps as in the previous test case. The simulation interval is 3 s, with a numerical integration time-step of $\Delta t = 50\mu s$. The phase-a-to-ground fault with a duration of 0.1 s occurs at $t = 2s$ at the location indicated in Figure 5.4.

Table 5.2 presents the performance timings of all test networks using both the conventional sequential step-wise method and the proposed PEGR-based parallel-in-time approach.

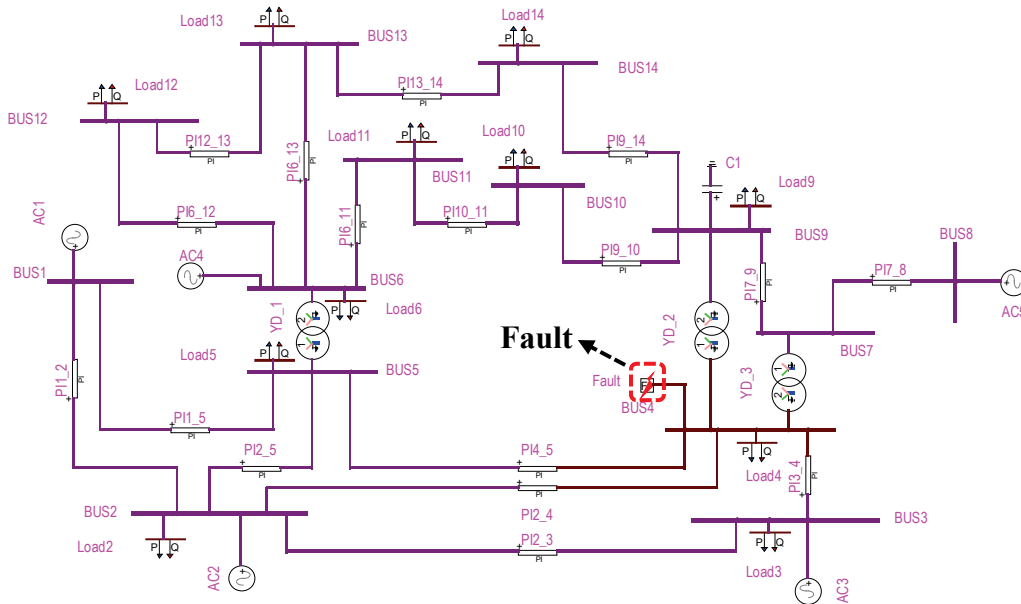


Figure 5.4 Network-6 benchmark

Table 5.2 Performance timings of test case 2

	Step-wise approach	PEGR-based approach		
Number of launched threads	N/A	8	16	32
14-bus network solution times (s)	9.56	12.23	8.65	5.39
70-bus network solution times (s)	35.24	21.83	17.40	12.70
140-bus network solution times (s)	85.61	42.44	34.98	27.23
210-bus network solution times (s)	137.98	59.62	47.89	37.41
280-bus network solution times (s)	185.01	72.94	57.89	46.17

Based on the data presented in Table 5.2, the solution time speedup using the PEGR-based approach compared to the conventional sequential step-wise approach with respect to the number of launched threads for different test networks can be calculated and is shown in Figure 5.5.

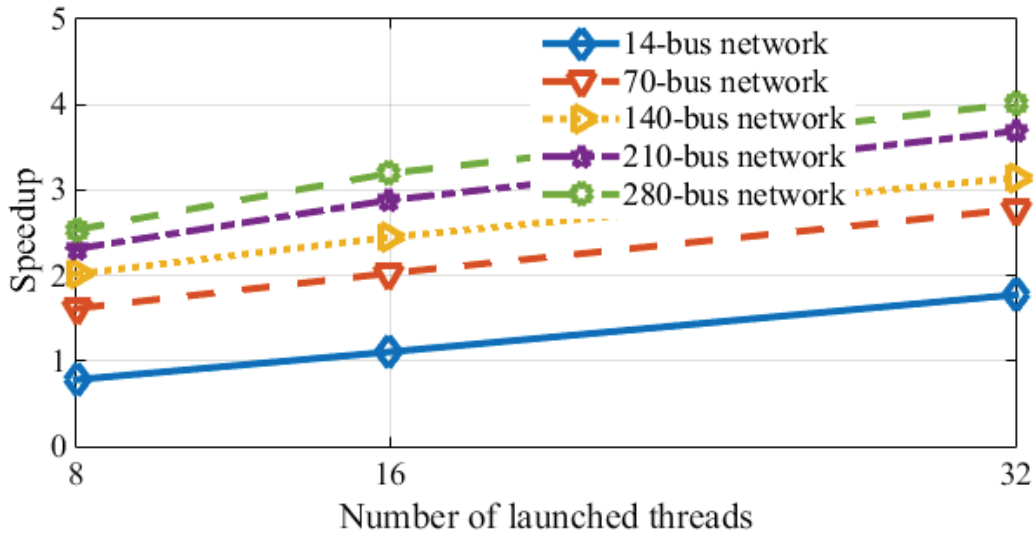


Figure 5.5 Solution time speedup using the PEGR-based parallel-in-time approach with respect to the number of launched threads in the simulation for test case 2

5.2.3 Discussion

As was mentioned earlier, the goal of the tests performed in this section is to demonstrate the enhanced simulation efficiency using the PEGR-based parallel-in-time approach thanks to the equation grouping and recursive row and column reordering scheme it adopts, compared to the conventional sequential step-by-step network solution scheme. Therefore, in both compared approaches, the KLU sparse linear solver is employed as the sparse matrix solver for network equations, and both approaches are implemented in C++ and share the same set of function routines

in network matrix factorization, solution, forward and backward substitutions. This aims to create a common platform to compare the two solution schemes which are the parallel-in-time solution scheme and the conventional sequential step-wise solution scheme in order to demonstrate the numerical advantages of the former, whereas not to compare any two specific solvers, . This is why the performance timings using the PEGR-based parallel-in-time approach for both test cases presented in Table 5.1 and Table 5.2 could still be inferior to commercialized power system network solver with highly optimized routines [2] for the particular test cases presented here.

The performance timings presented in Table 5.1 and Table 5.2 include all necessary procedures in time-domain simulations. Particularly for the case using the PEGR-based parallel-in-time approach, it includes all three stages shown in Section 4.2.4, that is to say, the equation grouping and recursive row and column reordering, factorization and refactorization of the grouped expanded network equations whose size is much larger than that of the original network equations, as well as extraction of block elements for forward and backward substitutions, are also included. This proves, through the results presented in Figure 5.3 and Figure 5.5, that despite the need of operating on a much larger sparse matrix comprised of solutions at 16, 32 and even 64 time-steps, the proposed parallel-in-time approach can still be more efficient than the conventional sequential step-wise solution scheme.

From Figure 5.3 and Figure 5.5, a few conclusions can be drawn:

- Higher gains in a simulation environment with more threads

This is fairly easy to understand from the original proposition of the PEGR scheme, as the recursive row and column reordering scheme was designed such that the total solution steps needed to solve a power system network can be theoretically logarithmically reduced [30]. Therefore, the more threads the simulation launches, the more forward and backward substitution steps that can be parallelized, which leads to an even more reduced number of total solution steps needed compared to the conventional step-wise solution scheme.

Nonetheless, it is also observed that the actual computation time gains obtained in both test cases are lower than those expected in the original proposition [30]. For instance, the original proposition suggests that the total number of forward substitution steps will be reduced to 4, 5 and 6, as compared to 16, 32 and 64, if 8, 16, and 32 parallel processors are available. However, firstly the original proposition only refers to the numbers that forward substitution steps can be theoretically

reduced to with sufficient parallel processors whilst not considering other procedures needed in time-domain simulations using the PEGR-based parallel-in-time approach such as equation grouping, recursive row and column reordering of grouped network equations, factorization and refactorization of the reordered and grouped network equations as well as block element extraction. Secondly, certain measures adopted in the implementation of such an approach in C++ with OpenMP Multithreading, such as enforcing consistency between the temporary view and memory by using the **flush** construct, have negative impacts on the total solution times. Thirdly, overhead caused by the inner mechanism of OpenMP also impacts the performance of the proposed PEGR-based approach negatively.

- Higher gains on larger networks

It can be observed from Figure 5.3 and Figure 5.5 that for the same number of launched threads in a simulation, as the network becomes larger, higher computation time gains can be achieved. This is because for a given number of launched threads, the computational efficiency enhancement brought by the PEGR-based approach for smaller networks is dwarfed by the time consumed in the supplementary operations in the PEGR-based approach that are not necessary in the conventional step-wise solution scheme, considering that the adopted sparse linear solver KLU is sufficiently optimized thus highly efficient for the solution of smaller networks; however, as the network grows larger, the time consumed in the supplementary operations in the PEGR-based approach becomes less significant, the computational efficiency enhancement brought by the PEGR-based approach therefore becomes more significant.

Despite the potential performance enhancement brought by the PEGR-based approach, there exist certain limitations in the implementation of the PEGR-based approach on realistic large-scale power system time-domain simulations:

- Higher formulation and solution complexity

Apart from formulating the original network MANA coefficient matrix, the PEGR-based approach requires special treatment of the history terms in energy storage components such that interdependency between network solutions at two consecutive time-points can be obtained. Subsequently, the equation grouping and recursive row and column reordering procedures with respect to the number of launched threads need to be performed in order to achieve solution parallelism. After network matrix factorization, an extra block element extraction procedure is

required in the forward and backward substitution processes. All of the aboved-mentioned procedures, which do not exist in the original MANA formulation method, add to the formulation and solution complexity.

- More challenges on the PC memory capacity

The PEGR-approach requires operations on a huge network matrix consisting of solutions at $T = 2^\tau$ (τ is an integer) time-steps, where T could be 16, 32, 64, 128, etc. Moreover, the inclusion of history terms of energy storage components further increases the size of the network matrix for the solution of a single time-step, compared to that of the original network MANA coefficient matrix. Therefore, the implementation of the PEGR-based approach on large-scale networks would pose great challenges on the PC memory capacity in terms of storage of certain vectors and matrices used in the factorization and solution procedures.

- Knowledge of network topological change time-points prior to simulation start

Due to the nature of the PEGR-based approach of solving the network at every 16, 32, 64, ..., T time-steps ($T = 2^\tau$, τ is an integer), the network topological change time-points (fault, switching, etc.) need to be known and analyzed prior to time-domain simulation start in order that a step-wise conventional approach can be used to handle the topological changes. This is because the network solution of every 16, 32, 64, ... T time-steps ($T = 2^\tau$, τ is an integer) using the PEGR-approach is a complete process and cannot be broken or stopped in the middle due to an arbitrary topological change. This differs from the conventional EMT approach [2] and would introduce supplementary burden into the pre-processing and solution procedures. Furthermore, although computation time gains are observed in the test cases presented in this thesis, performance deterioration could happen due to extensive switching between two types of solvers in test cases with multiple topological changes.

- Difficulty in handling non-linearity

In the original PEGR technique, it was proposed to isolate nonlinear equations, group them and solve them independently from linear equations using the Newton method [30]. The problems with this proposition are: firstly, obtaining good initial estimates for a set of nonlinear equations comprised of the solutions at multiple time-steps in order to lower the number of iterations and achieve convergence is not straightforward; secondly, the interaction and data exchange between

linear and nonlinear equations in a realistic complex power system network are not answered for. A numerical example of solving one single isolated nonlinear equation using the PEGR technique was presented in [30], demonstrating the feasibility of the PEGR technique in handling non-linearity, there is, however, not sufficient proof that the above-mentioned two problems could be accounted for by this proposition. EMTP [2] linearizes nonlinear components by the method of “piecewise linearization”, formulates both linear and nonlinear components into the MANA coefficient matrix and performs iterations within the solution of each time-step to reach convergence. Since the PEGR-based approach proposed in this thesis is based on the original MANA formulation in EMTP [2], it would require that iterations be performed within the solution of each time-step, which is impractical considering the PEGR-based approach solves the entire network at multiple time-steps together as a complete procedure. Therefore, the PEGR-based approach can only be implemented on linear networks if the MANA formulation method is kept as a base method.

5.3 Conclusion

Test results using the PEGR-based parallel-in-time approach developed in this thesis were presented in this chapter. The tests performed in this chapter aimed to validate the accuracy of the approach and the computation time gains that can be achieved in power system time-domain simulations. Two new benchmarks, as well as those constructed using their multiple replicas representing power systems of different levels of complexities, were used in the tests. Accuracy of the PEGR-based parallel-in-time approach was validated by comparing various waveforms using the PEGR-based approach with those obtained from an EMT-type solver, whereas computation time gains were demonstrated by comparing time-domain solution times using the developed approach to those using the conventional step-wise solution scheme. The advantages of the proposed PEGR-based parallel-in-time approach for large-scale power system time-domain simulations were demonstrated, the achieved computation time gains for different numbers of launched threads on power system benchmarks of different sizes were analyzed, and the limitations of the implementation of the PEGR-based parallel-in-time approach on realistic large-scale power system time-domain simulations were also explained.

CHAPTER 6 CONCLUSION AND RECOMMENDATIONS

6.1 Summary of the thesis

The main objective of this thesis is to develop new numerical techniques for fast and flexible power system electromagnetic transient simulations with the help of modern-day multicore computers. It is triggered by the increasing complexity of modern power systems brought by state-of-the-art developments in HVDC and wind generation, and considerable solution times rendered by relatively small numerical integration time-steps in conventional EMT-type solution schemes when solving large sets of differential and algebraic equations. The important remarks of this thesis are summarized in the following categories.

- Establishment of compatibility between the Functional Mock-up Interface (FMI) standard and EMTP

The Functional Mock-up Interface (FMI) standard is a tool independent interface standard that provides a means of integrating dynamic models developed in different simulation environments into a common simulation platform. Following the requirements and protocols (in the form of xml files and C-code) defined in the standard, importing and exporting a dynamic model from one simulation tool to another can be greatly facilitated. The FMI standard targets a wide range of dynamic simulation needs in various domains. In this thesis, full compatibility is established between the co-simulation form of the FMI standard and EMTP, significantly enhancing the flexibility of conventional EMT-type solution schemes by allowing network decoupling in memory, parallel and multistep simulation on the conventional EMTP platform, thus responding to the demand of computing time reduction in large-scale power system EMT simulations.

- Development of a co-simulation based off-line simulation approach using the FMI standard for power system parallel and multistep electromagnetic transients simulations with complex control systems

A co-simulation based off-line simulation approach using the FMI standard for power system parallel and multistep EMT simulations with complex control systems is developed in this thesis. Taking advantage of the full compatibility established between the co-simulation form of the FMI standard and EMTP, this approach creates a co-simulation interface between the power network

and various decoupled control systems (e.g., wind generator controls and protection relays), greatly reducing the total computation burden by distributing the solution of the power network and control systems among different processors.

The power network is designated as the “master” whereas the various decoupled control systems as “slaves” during the simulation, and they communicate with each other through the co-simulation interface with the help of synchronization functions implementing the low-level synchronization primitive semaphore. The communication between the master and slaves is achieved and organized within the context of the FMI standard. This approach offers two co-simulation modes, which are the asynchronous mode and the synchronous mode that allow the decoupled control systems to be simulated either in parallel or in series with the master respectively. There exist no constraints on the numerical integration time-steps used in the master or slaves in either mode, indicating that the master and slaves can use either the same or different numerical integration time-steps in either mode, hence allowing for higher simulation flexibility. Additionally, a linear extrapolation-based signal correction procedure is also developed in this approach, aiming to improve accuracy of floating-point type control signals in multistep simulations.

The developed co-simulation approach is tested for accuracy and simulation efficiency on realistic power system benchmarks of different levels of complexity.

The accuracy tests focus on comparing different variables obtained from power system benchmarks implementing the FMI-based co-simulation approach with those from the same benchmarks without co-simulation. Satisfactory results are obtained, demonstrating that the implementation of the FMI-based co-simulation approach developed in this thesis does not affect simulation accuracy and the linear extrapolation-based signal correction procedure can indeed improve accuracy of floating-point type control signals in a multistep simulation environment. A detailed error analysis of the obtained results is also provided.

The tests for computation time gains are based on the comparison of solution times for benchmarks with and without co-simulation. Several PCs with different numbers of processors are used in these tests, with both co-simulation modes and various numerical integration time-steps implemented in the control systems. Considerable computation time gains are achieved in the cases using the FMI-based co-simulation approach developed in this thesis, demonstrating the numerical advantage of the developed approach in the aspect of simulation efficiency enhancement.

- Development of an off-line parallel-in-time approach based on the parallel-in-time equation grouping and reordering (PEGR) technique for power system electromagnetic transients simulations

Apart from the FMI-based co-simulation approach developed in this thesis, a parallel-in-time approach based on the parallel-in-time equation grouping and reordering (PEGR) technique for power system electromagnetic transients simulations is also developed. Extending the original proposition of the PEGR technique from state-space to the modified-augmented-nodal analysis (MANA), this parallel-in-time approach groups the expanded MANA equations at several solution time-points to construct a large DAE system then applies a recursive row and column reordering scheme on the constructed large network matrix, thus revealing the independency between the solutions at certain time-points. This approach thereby takes advantage of the independency between solutions at these time-points to logarithmically reduce the total number of steps in the forward and backward substitution processes. The developed approach can group the expanded MANA equations at $16, 32, 64, \dots, 2^{n+4}$ ($n = 0, 1, 2, \dots$) consecutive time-points, requiring therefore $8, 16, 32, \dots, 2^{n+3}$ ($n = 0, 1, 2, \dots$) parallel processors to realize its full potential.

The parallel-in-time approach developed in this thesis is implemented in C++ with API OpenMP Multithreading, using the highly efficient sparse linear solver KLU as the network matrix solver. Measures to minimize false sharing and data race, enforce consistency between the temporary view and memory, as well as treatments of arbitrary points of topological changes are discussed in the thesis.

This newly developed approach is tested for accuracy and efficiency using different power system benchmarks and those constructed with their multiple replicas (representing networks of much larger scales).

Once again, the accuracy tests are based on the comparison of various variables obtained from a benchmark implementing the parallel-in-time approach and an EMT-type solver, with results showing an exact match between variables obtained from these two approaches.

The tests for computation time gains focus on comparing the developed parallel-in-time approach with the conventional EMT-type step-wise solution schemes that employ the same functions in matrix factorization as well as forward and backward substitutions, not with any highly optimized

commercial power system EMT simulation software. This aims to create a common platform to compare the two solution schemes in order to demonstrate the numerical advantages brought by the proposed approach. Results from benchmarks of different levels of complexity with different numbers of launched threads (within the capacity of the PC used in the tests) indicate an enhanced simulation efficiency using the parallel-in-time approach developed in this thesis compared to the conventional EMT-type step-wise solution scheme, therefore showing great prospects for fast power system electromagnetic transients simulations from a different perspective.

The numerical techniques developed in this PhD project are readily applicable to parallel and multistep power system electromagnetic transients simulations using the off-the-shelf multicore PCs. The methodology followed in the tests, together with the various benchmarks used, can serve as a reference and be adopted in the development of other numerical techniques for large power system EMT simulations. Finally, the discussions on the observed computation time gains in both approaches developed in this thesis can be useful insights in analyzing the numerical behaviour of other parallel and multistep EMT approaches. A list of journal and conference publications derived from this PhD project can be found in Appendix D.

6.2 Future work

Considering the general scope of this thesis, which is the numerical techniques for parallel and multistep power system EMT simulations, possible developments are as follows.

- Expanding the implementation of the FMI-standard to arbitrary decoupling of power signals

The FMI standard currently has been implemented for control signals (integer, floating-point or Boolean) that can be exchanged by means of the co-simulation interface. Therefore, a possible future development is to generalize the implementation of the FMI standard to include decoupling of power signals, which could further improve simulation flexibility and efficiency. To this end, iterative co-simulation between the main and decoupled networks must be established. The current implementation of the FMI standard does not accommodate iteration between the decoupled control systems and the power network, an artificial delay of one time-step is thus manually added during the master-slave data exchange process for simulation stability and robustness. There exists

an attribute named `canHandleVariableCommunicationStepSize` in the FMI standard for co-simulation (please see page 110 in [19]). If this attribute is set to “true”, iterative co-simulation between the main and decoupled networks could be achieved with a time-step of zero. This would not only allow the generalization of the implementation of the FMI standard on arbitrary decoupling of power signals, but also improve simulation accuracy by possibly removing the added one time-step delay. Of course, details on co-simulation management in an iterative co-simulation environment would also need to be studied.

- Investigation on the application of the PEGR-based parallel-in-time approach on networks with non-linear elements

As was explained in the previous chapter, the current implementation of the PEGR-based parallel-in-time approach can only handle networks with linear components due to its special feature of combining the solutions at several time-points and solving them together. Therefore, no iteration is performed at the solution of any time-point. The possible application of the PEGR technique in non-linear networks was briefly discussed in [30], in which it was proposed that nonlinear equations are isolated, grouped, and solved independently from linear equations using the Newton method with a simple example of solving one single nonlinear equation using the PEGR technique demonstrating its feasibility in the solution of nonlinear systems. However, this does not resolve the problems of obtaining good initial estimates for the solutions at multiple time-steps in order to decrease the number of iterations and achieve convergence as well as interaction and data exchange with the linear networks as is the case in realistic complex power systems. It was also explained in the previous chapter that it would be impractical for the current implementation of the PEGR-based parallel-in-time approach to handle iterations within each time-step while keeping the original MANA formulation in EMTP as a base method. Therefore, another possible future development is to investigate the application of the PEGR-based parallel-in-time approach on networks with non-linear elements. The discussion on the application of the PEGR technique in non-linear networks should be revisited and further studied. The combination of the formulations for both linear and non-linear networks needs to be investigated. If a new formulation suitable for both linear and non-linear network components can indeed be found, its formulation complexity will need to be evaluated and the computation time gains it can possibly bring will need to be studied.

- Investigation on the possibility of a hybrid approach integrating the PEGR-based parallel-in-time approach with another approach

Another possible aspect to look into is to combine the PEGR-based parallel-in-time approach with another approach to form a hybrid approach. Since the current PEGR-based parallel-in-time approach can only be used to solve linear networks, the entire network to be solved could thus be decoupled, with the linear subnetworks being solved using the PEGR-based parallel-in-time approach while the other approach used to solve the rest. This is theoretically achievable as long as the decoupling interface can be well defined. Multistep simulation could also be possible with different numerical integration time-steps used in the subnetworks solved by the two approaches (e.g., the PEGR-based approach can adopt a smaller numerical integration time-step to solve the subnetwork that involves fast transients or requires detailed and accurate waveforms whereas the rest of the network is solved using the other approach with a larger numerical integration time-step). However, considering the PEGR-based approach combines the network equations at every 16, 32, 64, ... time-points, unless the numerical integration time-step used in solving the rest of the network is 16, 32, 64, ... times of that used by the PEGR-based approach, coordinating different numerical integration time-steps used in the two approaches would need to be investigated, which could involve the use of the compensation method [29] and/or signal interpolation or extrapolation.

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APPENDIX A – EXAMPLE XML FILE IN AN FUNCTIONAL MOCK-UP UNIT (FMU)

The contents of the xml file in the FMU generated from a wind generator control system are presented as follows.

```
<fmiModelDescription
  fmiVersion="2.0"
  modelName="WG_CONTROL200"
  guid="{aa787663-1998-43ad-828d-ec831fd998db}"
  generationTool="EMTP-RV"
  description="FMU Generated with EMTP-RV. Parameters emtpopt.exe and emtpstate.ini must exist and the
corresponding  EMTP must be opened (see EMTP FMI documentation)."
```

generationDateAndTime="2017-7-13 15:46:53"

variableNamingConvention="flat"

numberOfEventIndicators="0">


```
<CoSimulation
  modelIdentifier="EMTP"
  needsExecutionTool="false"
  canHandleVariableCommunicationStepSize="false"
  canInterpolateInputs="false"
  maxOutputDerivativeOrder="0"
  canRunAsynchronously="false"
  canBeInstantiatedOnlyOncePerProcess="false"
  canNotUseMemoryManagementFunctions="false"
  canGetAndSetFMUstate="false"
  canSerializeFMUstate="false"
  providesDirectionalDerivative="false">
</CoSimulation>
```



```

<TypeDefinitions>

  <SimpleType

    name="CoSimulationModeType" description="Co-simulation mode">

      <Enumeration>

        <Item name="Mode1" description="Synchronous mode" value="1"/>

        <Item name="Mode2" description="Asynchronous mode, same time step" value="2"/>

      </Enumeration>

    </SimpleType>
  </TypeDefinitions>

  <LogCategories>

    <Category name="logEvents"/>

    <Category name="logStatusDiscard"/>

    <Category name="logStatusWarning"/>

    <Category name="logStatusError"/>

    <Category name="logStatusFatal"/>

    <Category name="logStatusPending"/>

    <Category name="logAll"/>

    <Category name="logSingularLinearSystems"/>

    <Category name="logNonLinearSystems"/>

    <Category name="logDynamicStateSelection"/>

  </LogCategories>

  <DefaultExperiment startTime="0" stopTime="0.032" tolerance="1e-04" stepSize="0.00005"/>

  <VendorAnnotations>

    <Tool

      name="EMTPRV">

        <customEMTPRVData netlistName="DFIG Control.net"/>

      </Tool>

```

</VendorAnnotations>

<ModelVariables>

<ScalarVariable name="w_rotor_rads" valueReference="0" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Vdc_V" valueReference="1" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_converter_A_Ia" valueReference="2" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_converter_A_Ib" valueReference="3" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_converter_A_Ic" valueReference="4" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_Stator_A_Ia" valueReference="5" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_Stator_A_Ib" valueReference="6" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_Stator_A_Ic" valueReference="7" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_Rotor_A_Ia" valueReference="8" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_Rotor_A_Ib" valueReference="9" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_Rotor_A_Ic" valueReference="10" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_Grid_A_Ia" valueReference="11" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_Grid_A_Ib" valueReference="12" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Iabc_Grid_A_Ic" valueReference="13" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

<ScalarVariable name="Vabc_Grid_A_Va" valueReference="14" description="Description..." initial="approx" variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>

```
<ScalarVariable name="Vabc_Grid_A_Vb" valueReference="15" description="Description..." initial="approx"
variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="Vabc_Grid_A_Vc" valueReference="16" description="Description..." initial="approx"
variability="continuous" causality="input"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="pitch" valueReference="17" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="Vref_grid_a" valueReference="18" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="Vref_grid_b" valueReference="19" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="Vref_grid_c" valueReference="20" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="block_grid" valueReference="21" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="Vref_rotor_a" valueReference="22" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="Vref_rotor_b" valueReference="23" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="Vref_rotor_c" valueReference="24" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="block_rotor" valueReference="25" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="chopper_active" valueReference="26" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="crowbar_active" valueReference="27" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="switch_on" valueReference="28" description="Description..." initial="approx"
variability="continuous" causality="output"><Real start="0.0"/></ScalarVariable>
```

```
<ScalarVariable name="emptopt.exe location" valueReference="20000" description="Location of the emptopt.exe file"
variability="tunable" causality="parameter"><String start="C:\Program Files (x86)\EMTPWorks
3.4\EMTP\emptopt.exe"/></ScalarVariable>
```

```
<ScalarVariable name="emptstate.ini location" valueReference="20001" description="Location of the emptstate.ini file"
variability="tunable" causality="parameter"><String
start="C:\Users\A.M\AppData\Roaming\EMTP\C_Program_Files_x86_EMTPWorks_3_4\emptstate.ini"/></ScalarVariable>
```

```
<ScalarVariable name="Co-simulation mode" valueReference="20002" description="List of available Co-simulation
modes" variability="tunable" causality="parameter"><Enumeration
start="1"declaredType="CoSimulationModeType"/></ScalarVariable>
```

```
<ScalarVariable name="Time Out (ms)" valueReference="20003" description="Maximal time for synchronisation of the
co-simulation" variability="tunable" causality="parameter"><Integer start="5000" min="2000"/></ScalarVariable>
```

```
</ModelVariables>
```

```
<ModelStructure>
```

```
</ModelStructure>
```

```
</fmiModelDescription>
```

APPENDIX B – DETAILED IMPLEMENTATION OF FUNCITONS

COORDINATING MASTER-SLAVE CO-SIMULATION

The detailed implementation of functions `fmi2DoStep` (in file `FMI2_Link_Master`) and `stepFunc` (in file `FMI2_Device_Master`) is presented in this appendix. It is noted that these two functions are directly involved in coordinating the co-simulation between master and slaves.

- `fmi2DoStep`

```
fmi2Status fmi2DoStep(fmi2Component c, fmi2Real currentCommunicationPoint,
    fmi2Real communicationStepSize, fmi2Boolean noSetFMUStatePriorToCurrentPoint) {
    ModelInstance *comp = (ModelInstance *)c;
    if (invalidState(comp, "fmi2DoStep", modelInitialized | modelStepping))
        return fmi2Error;
    comp->state = modelStepping;

    int nbIter = 0;
    double time_slave;
    double step_slave;

    FILTERED_LOG(comp, fmi2OK, logAll, "fmi2DoStep: "
        "time out = %i,"
        "mode slave = %i,"
        "currentCommunicationPoint = %g, "
        "communicationStepSize = %g, "
        "noSetFMUStatePriorToCurrentPoint = fmi%s",
        comp->timeOut, comp->SlaveMode, currentCommunicationPoint,
        communicationStepSize, noSetFMUStatePriorToCurrentPoint ? "True" : "False")

    if (communicationStepSize <= 0) {
        FILTERED_LOG(comp, fmi2Error, logStatusError,
            "fmi2DoStep: communication step size must be > 0. Found %g.",
            communicationStepSize)
        return fmi2Error;
    }

    setValueInInOutVector(tab_index_MasterTime, currentCommunicationPoint, comp);
    setValueInInOutVector(tab_index_MasterStep, communicationStepSize, comp);

    time_slave = getValueFromInOutVector(tab_index_SlaveTime, comp);
    step_slave = getValueFromInOutVector(tab_index_SlaveStep, comp);

    FILTERED_LOG(comp, fmi2OK, logAll, "fmi2DoStep: "
        "time slave = %g,"
        "time master = %g,",
        getValueFromInOutVector(tab_index_SlaveTime, comp),
        getValueFromInOutVector(tab_index_MasterTime, comp));

    if ((comp->SlaveMode == SlaveSynchron))
    {
        while(time_slave < currentCommunicationPoint + 0.95 * communicationStepSize)
        {
```

```

        if (ReleaseSemaphore(comp, SemSlave, 1) == false)
        {
            FILTERED_LOG(comp, fmi2Error, logAll, "fmi2DoStep : Error while
releasing slave semaphore (SEM3)")
            return(fmi2Error);
        }

        if (WaitSemaphoreTimeOut(comp, SemMaster, 1, comp->timeOut) == false)
        {
            stringstream ssTimeOut;
            ssTimeOut << comp->timeOut;
            string strTimeOut = ssTimeOut.str();
            FILTERED_LOG(comp, fmi2Error, logAll, "fmi2DoStep : Error while
waiting master semaphore (SEM2), waiting time out: %s ms (b)", strTimeOut.c_str())
            return(fmi2Error);
        }
        time_slave = getValueFromInOutVector(tab_index_SlaveTime, comp);
        FILTERED_LOG(comp, fmi2OK, logAll, "fmi2DoStep finished: "
            "time slave = %g,"
            "time master = %g,",
            time_slave, currentCommunicationPoint);
    }
}

else
{
    if (currentCommunicationPoint == 0)
    {
        if (ReleaseSemaphore(comp, SemSlave, 1) == false)
        {
            FILTERED_LOG(comp, fmi2Error, logAll, "fmi2DoStep : Error
while releasing slave semaphore (SEM3)")
            return(fmi2Error);
        }
        FILTERED_LOG(comp, fmi2OK, logAll, "fmi2DoStep Slave is released
for the first step: "
            "time slave = %g,"
            "time master = %g,",
            getValueFromInOutVector(tab_index_SlaveTime, comp),
            getValueFromInOutVector(tab_index_MasterTime, comp));
    }
    else
    {
        if (WaitSemaphoreTimeOut(comp, SemMaster, 1, comp->timeOut) ==
false)
        {
            stringstream ssTimeOut;
            ssTimeOut << comp->timeOut;
            string strTimeOut = ssTimeOut.str();
            FILTERED_LOG(comp, fmi2Error, logAll, "fmi2DoStep : Error
while waiting master semaphore (SEM2), waiting time out: %s ms (d)", strTimeOut)
            return(fmi2Error);
        }
        time_slave = getValueFromInOutVector(tab_index_SlaveTime, comp);
        FILTERED_LOG(comp, fmi2OK, logAll, "Slave calculation finished: "
            "time slave = %g,"
            "time master = %g,",
            time_slave, currentCommunicationPoint);
    }
}

```

```

        while(time_slave < currentCommunicationPoint -
0.05*communicationStepSize)
        {
            if (ReleaseSemaphore(comp, SemSlave, 1) == false)
            {
                FILTERED_LOG(comp, fmi2Error, logAll, "fmi2DoStep :
Error while releasing slave semaphore (SEM3)")
                return(fmi2Error);
            }
            FILTERED_LOG(comp, fmi2OK, logAll, "Slave has to do another
calculation ("
            "time slave = %g,"
            "time master = %g,)",
            time_slave, currentCommunicationPoint);
            if (WaitSemaphoreTimeOut(comp, SemMaster, 1, comp->timeOut)
== false)
            {
                stringstream ssTimeOut;
                ssTimeOut << comp->timeOut;
                string strTimeOut = ssTimeOut.str();
                FILTERED_LOG(comp, fmi2Error, logAll, "fmi2DoStep :
Error while waiting master semaphore (SEM2), waiting time out: %s ms (d)", strTimeOut)
                return(fmi2Error);
            }
            time_slave =
getValueFromInOutVector(tab_index_SlaveTime, comp);
            FILTERED_LOG(comp, fmi2OK, logAll, "Slave calculation
finished: "
            "time slave = %g,"
            "time master = %g,",
            time_slave, currentCommunicationPoint);
        }
        FILTERED_LOG(comp, fmi2OK, logAll, "fmi2DoStep finished: "
        "time slave = %g,"
        "time master = %g,",
        time_slave, currentCommunicationPoint);
        if (ReleaseSemaphore(comp, SemSlave, 1) == false)
        {
            FILTERED_LOG(comp, fmi2Error, logAll, "fmi2DoStep : Error
while releasing slave semaphore (SEM3)")
            return(fmi2Error);
        }
    }

    if (comp->eventInfo.nextEventTimeDefined && ((comp->time -
comp->eventInfo.nextEventTime) > -0.0000000001)) {
        FILTERED_LOG(comp, fmi2OK, logAll, "fmi2DoStep: time event detected
at %g", comp->time);
    }
    if (comp->eventInfo.terminateSimulation) {
        FILTERED_LOG(comp, fmi2Discard, logAll, "fmi2DoStep: model
requested termination at t=%g", comp->time)
        return fmi2Discard;
    }
    return fmi2OK;
}

```

- stepFunc

```

void EMTPDevice::stepFunc(double *simulation_Time_Point, int *procedure_type, double
*inputs_u, double *outputs_y)
{
    double timeNow = this->EMTPData.simTime->t;
    int mainCtr, realCtr, intCtr, stringCtr, boolCtr;
    mainCtr = realCtr = intCtr = stringCtr = boolCtr = 0;
    this->isNewTime = (this->oldTime < timeNow);
    fmi2Status status;
    string errorMsg;

    if (timeNow == 0)
    {
        if (this->isNewTime)
        {
            status = this->fmi2EnterInitializationMode(this->s1);
            if (status == fmi2Error){
                errorMsg = Def_Error_Msg15b;
                this->sendError(errorMsg, true);
                return;
            }
            this->flagInit = true;
            status = this->setIN();
            if (status == fmi2Error) return;

            status = this->setINIE();
            if (status == fmi2Error) return;

            if (!(this->noInitializationOfInputOutput))
            {
                status = this->setFMUInput(inputs_u);
                if (status == fmi2Error)
                {
                    return;
                }
            }
            for(int i = 0 ; i< this->nInputs; i++)
                this->prev_inputs[i] = inputs_u[i];

            if (!(this->noInitializationOfInputOutput))
            {
                status = this->getFMUOutputs();
                if (status == fmi2Error)
                {
                    return;
                }
            }
        }
        else
        {
            if (!(this->noInitializationOfInputOutput))
            {
                status = this->setFMUInput(inputs_u);
                if (status == fmi2Error)
                {
                    return;
                }
            }
        }
    }
}

```



```

        }
        for(int i = 0 ; i< this->nInputs; i++)
this->prev_inputs[i] = inputs_u[i];

        if (!(this->noInitializationOfInputOutput))
        {
            status = this->getFMUOutputs();
            if (status == fmi2Error)
            {
                return;
            }
        }
    }
}
else
{
    if ((timeNow > 0) && this->flagInit)
    {
        status = this->fmi2ExitInitializationMode(this->s1);
        if (status == fmi2Error){
            errorMsg = Def_Error_Msg15c;
            this->sendError(errorMsg, true);
            return;
        }
        this->flagInit = false;
    }

    if (this->getAndSetFMUstate)
    {
        if (this->isNewTime)
        {
            status = this->fmi2GetFMUstate(this->s1,&this->FMUstate);
            if (status == fmi2Error) {
                std::ostringstream oss;
                oss << this->EMTPData.simTime->t;
                string stringtime = oss.str();
                errorMsg = Def_Error_Msg23;
                this->sendError(errorMsg, true);
                return;
            }
        }
        else
        {
            status = this->fmi2SetFMUstate(this->s1,this->FMUstate);
            if (status == fmi2Error) {
                std::ostringstream oss;
                oss << this->EMTPData.simTime->t;
                string stringtime = oss.str();
                errorMsg = Def_Error_Msg21;
                this->sendError(errorMsg, true);
                return;
            }
            status = this->fmi2GetFMUstate(this->s1,&this->FMUstate);
            if (status == fmi2Error) {
                std::ostringstream oss;
                oss << this->EMTPData.simTime->t;
                string stringtime = oss.str();

```

```

        errorMsg = Def_Error_Msg23;
        this->sendError(errorMsg, true);
        return;
    }
}
fmi2Real currentCommunicationPoint =
this->EMTPData.simTime->prev_t;
fmi2Real communicationStepSize = this->EMTPData.simTime->t -
this->EMTPData.simTime->prev_t;

    if (this->canHandleVariableCommunicationStepSize) {
        status =
this->doStep(currentCommunicationPoint,communicationStepSize,inputs_u);
        if (status == fmi2Error)
        {
            return;
        }
    }
    else if (fabs((this->EMTPData.simTime->t - this->slaveTime -
this->EMTPData.simTime->dt)/this->EMTPData.simTime->dt) < 0.00001)
    {
        status =
this->doStep(this->slaveTime,communicationStepSize,inputs_u);
        if (status == fmi2Error)
        {
            return;
        }
        this->slaveTime = this->slaveTime+communicationStepSize;
    }
}
else if (handleVariableCommunicationStepSize)
{
    fmi2Real currentCommunicationPoint = this->oldTime;
    fmi2Real communicationStepSize = timeNow - this->oldTime;
    status =
this->doStep(currentCommunicationPoint,communicationStepSize,inputs_u);
    if (status == fmi2Error)
    {
        return;
    }
}
else
{
    if (isNewTime){
        if (this->canHandleVariableCommunicationStepSize)
        {
            if(this->oldTime > 0)
            {
                fmi2Real currentCommunicationPoint =
this->previousCommunicationPoint;
                fmi2Real communicationStepSize =
this->previousStepSize;
                status =
this->doStep(currentCommunicationPoint,communicationStepSize,prev_inputs);
                if (status == fmi2Error)
                {

```

```

        return;
    }
    this->previousCommunicationPoint =
currentCommunicationPoint+communicationStepSize;
    this->previousStepSize=timeNow-this->oldTime;

    for(int i = 0 ; i< this->nInputs; i++)
        this->prev_inputs[i] = inputs_u[i];
    }
    else
    {
        this->previousCommunicationPoint =
this->oldTime;
        this->previousStepSize=timeNow-this->oldTime;
    }
}
else
{
    if (this->oldTime > 0)
    {
        if (fabs((this->previousCommunicationPoint +
this->previousStepSize - this->slaveTime -
this->EMTPData.simTime->dt)/this->EMTPData.simTime->dt) < 0.00001)
        {
            fmi2Real currentCommunicationPoint =
this->slaveTime;
            fmi2Real communicationStepSize =
this->oldTime - this->slaveTime;
            status =
this->doStep(currentCommunicationPoint,communicationStepSize,prev_inputs);
            if (status == fmi2Error)
            {
                return;
            }
            this->previousCommunicationPoint =
currentCommunicationPoint+communicationStepSize;
            this->slaveTime =
currentCommunicationPoint+communicationStepSize;
            this->previousStepSize = timeNow -
this->oldTime;
            for(int i = 0 ; i< this->nInputs; i++)
                this->prev_inputs[i] = inputs_u[i];
        }
        else
        {
            this->previousCommunicationPoint =
this->previousCommunicationPoint + this->previousStepSize;
            this->previousStepSize = timeNow -
this->oldTime;
        }
    }
    else
    {
        this->slaveTime = this->oldTime;
        this->previousCommunicationPoint =
this->oldTime;
        this->previousStepSize = timeNow-this->oldTime;
    }
}

```

```
        }  
    }  
}  
this->oldTime = timeNow;  
for(int i=0;i<nOutputs;i++)  
    outputs_y[i] = this->outputs[i];  
}
```

APPENDIX C – ALGORITHM FOR THE EXTRACTION OF BLOCK ELEMENTS IN THE L AND U FACTORS

The algorithm for the extraction of block elements in the **L** and **U** factors for forward and backward substitution is presented below. This algorithm is designed for any given number of launched threads in the simulation (the total number of elements to be extracted from the **L** and **U** factors and their positions in the **L** and **U** factors are dependent on the particular formulation determined by the number of solution steps grouped together).

```

    int Lnz_max = _lnz; /* the total non-zeros in the L matrix */
    int Unz_max = _unz;
    Lsize = size_sub_mat * size_sub_mat; /* maximum size of each element in the L
matrix */
    Lpsize = size_sub_mat + 1; /* size of the column vector = size of full
matrix + 1 */
    Usize = size_sub_mat * size_sub_mat; /* maximum number of non-zeros in each
element */
    Upsize = size_sub_mat + 1; /* size of the column vector = size of full
matrix + 1 */

    int size_perm_mat = perm_mat.get_size_perm_mat(); // The size of the
permuted matrix determines how many elements to extract from the L and U factors

    row_count_L = new int[(int)round(log2(size_perm_mat)) + 1];
    col_count_L = new int[(int)round(log2(size_perm_mat)) + 1];
    for (i = 0; i < (int)round(log2(size_perm_mat)) + 1; i++)
    {
        row_count_L[i] = 0;
        col_count_L[i] = 0;
    }

    Li_extract = new int*[(int)round(log2(size_perm_mat)) + 1];
    Lx_extract = new double*[(int)round(log2(size_perm_mat)) + 1];
    Lp_extract = new int*[(int)round(log2(size_perm_mat)) + 1];
    int s;
    for (j = 0; j < (int)round(log2(size_perm_mat)) + 1; j++)
    {
        Li_extract[j] = new int[Lsize];
        Lx_extract[j] = new double[Lsize];
        Lp_extract[j] = new int[Lpsize];
    }
    for (j = 0; j < (int)round(log2(size_perm_mat)) + 1; j++)
    {
        for (s = 0; s < Lsize; s++)
        {
            Li_extract[j][s] = 0;
            Lx_extract[j][s] = 0.0;
        }
        for (s = 0; s < Lpsize; s++)
        {
            Lp_extract[j][s] = 0;
        }
    }

```

```

    }

    j = 0; int p = 0; bool flag = 0;
    for (i = 0; i < Lnz_max; i++) // Iterate through all the elements in the
    original Li and Lx arrays
    {
        /* Extraction of the diagonal element */
        if (Li[i] >= 0 && Li[i] < size_sub_mat)
        {
            Li_extract[j][row_count_L[j]] = Li[i];
            Lx_extract[j][row_count_L[j]] = Lx[i];
            if (i >= Lp[p + 1])
            {
                Lp_extract[j][col_count_L[j] + 1] = row_count_L[j];
                col_count_L[j] = col_count_L[j] + 1;
                p++;
            }
            row_count_L[j] = row_count_L[j] + 1;
            flag = 1;
        }
        /*
        When the iteration hits elements between row[size_sub_mat] and
        row[2*size_sub_mat], means the extraction of diagonal is complete.
        Set flag to 0;
        */
        if (Li[i] >= size_sub_mat && Li[i] < 2 * size_sub_mat && flag == 1)
        {
            Lp_extract[j][col_count_L[j] + 1] = row_count_L[j]; // Final
            element in the column vector stores the number of non-zeros
            j++;
            flag = 0;
            p = 0;
        }
        /*
        If the iteration hits the last row, but j is not large enough to be the
        last element to extract, keep extracting these elements
        */
        if (Li[i] >= (size_perm_mat - 1) * size_sub_mat && Li[i] < size_perm_mat *
        size_sub_mat && j <= (int)round(log2(size_perm_mat)) - 1) // All elements to be
        extracted are in the last row
        {
            Li_extract[j][row_count_L[j]] = Li[i] - (size_perm_mat - 1) *
            size_sub_mat;
            Lx_extract[j][row_count_L[j]] = Lx[i];
            if (i >= Lp[p + (int)round((size_perm_mat*(1 - 1 / pow(2, j)) -
            1)*size_sub_mat) + 1])
            {
                Lp_extract[j][col_count_L[j] + 1] = row_count_L[j];
                col_count_L[j] = col_count_L[j] + 1;
                p++;
            }
            row_count_L[j] = row_count_L[j] + 1;
            flag = 1;
        }
        /* Completion of extraction*/
        if (Li[i] >= (int)round((size_perm_mat*(1 - 1 / pow(2, j)))*size_sub_mat)
        && Li[i] < (int)round((size_perm_mat*(1 - 1 / pow(2, j)) + 1)*size_sub_mat) && flag ==
        1 && j > 0)
    }

```

```

        {
            Lp_extract[j][col_count_L[j] + 1] = row_count_L[j];
            j++;
            flag = 0;
            p = 0;
        }
        /* Extract the last element */
        if (Li[i] >= (size_perm_mat - 1) * size_sub_mat && Li[i] < size_perm_mat *
size_sub_mat && j > (int)round(log2(size_perm_mat)) - 1 && i <= Lnz_max - 1 -
row_count_L[0])
        {
            Li_extract[j][row_count_L[j]] = Li[i] - (size_perm_mat - 1) *
size_sub_mat;
            Lx_extract[j][row_count_L[j]] = Lx[i];
            if (i >= Lp[p + (int)round((size_perm_mat*(1 - 1 / pow(2, j)) -
1)*size_sub_mat) + 1])
            {
                Lp_extract[j][col_count_L[j] + 1] = row_count_L[j];
                col_count_L[j] = col_count_L[j] + 1;
                p++;
            }
            row_count_L[j] = row_count_L[j] + 1;
        }
        /* Completion of the extraction of the last element */
        if (i > Lnz_max - 1 - row_count_L[0])
        {
            Lp_extract[j][col_count_L[j] + 1] = row_count_L[j];
            p = 0;
        }
    }

    row_count_U = new int[(int)round(log2(size_perm_mat)) - 1];
    col_count_U = new int[(int)round(log2(size_perm_mat)) - 1];
    U_flag = new int[(int)round(log2(size_perm_mat)) - 1];
    for (i = 0; i < (int)round(log2(size_perm_mat)) - 1; i++)
    {
        row_count_U[i] = 0;
        col_count_U[i] = 0;
        U_flag[i] = 0;        // The flag is turned on (1) when the particular
element is ready to be extracted ( a right row has been hit)
    }

    Ui_extract = new int*[(int)round(log2(size_perm_mat)) - 1];
    Ux_extract = new double*[(int)round(log2(size_perm_mat)) - 1];
    Up_extract = new int*[(int)round(log2(size_perm_mat)) - 1];
    for (j = 0; j < (int)round(log2(size_perm_mat)) - 1; j++)
    {
        Ui_extract[j] = new int[Usize];
        Ux_extract[j] = new double[Usize];
        Up_extract[j] = new int[Usize];
    }
    for (j = 0; j < (int)round(log2(size_perm_mat)) - 1; j++)
    {
        for (s = 0; s < Usize; s++)
        {
            Ui_extract[j][s] = 0;
            Ux_extract[j][s] = 0.0;
        }
    }

```

```

        for (s = 0; s < Upsize; s++)
        {
            Up_extract[j][s] = 0;
        }
    }

    j = 0; int m = 3; p = 0;
    for (i = 0; i < Unz_max; i++)
    {
        if (Ui[i] >= (int)round((size_perm_mat*(1 - 1 / pow(2, j + 1)) -
1))*size_sub_mat && Ui[i] < (int)round((size_perm_mat*(1 - 1 / pow(2, j +
1))))*size_sub_mat && j < (int)round(log2(size_perm_mat)) - 2)
        {
            U_flag[j] = 1;
        }
        /* Extract the first element*/
        if (j == 0 && Ui[i] >= size_sub_mat && Ui[i] < 2 * size_sub_mat &&
U_flag[j] == 1)
        {
            Ui_extract[j][row_count_U[j]] = Ui[i] - size_sub_mat;
            Ux_extract[j][row_count_U[j]] = Ux[i];
            if (i >= Up[p + (int)round((size_perm_mat*(1 - 1 / pow(2, j +
1))))*size_sub_mat + 1])
            {
                Up_extract[j][col_count_U[j] + 1] = row_count_U[j];
                col_count_U[j] = col_count_U[j] + 1;
                p++;
            }
            row_count_U[j] = row_count_U[j] + 1;
        }
        /* Extract U2, U3, etc */
        if (Ui[i] >= (2 + (int)round(size_perm_mat*(1 - 1 / pow(2, j)))) -
1)*size_sub_mat && Ui[i] < (2 + (int)round(size_perm_mat*(1 - 1 / pow(2,
j))))*size_sub_mat && j > 0 && j < (int)round(log2(size_perm_mat)) - 2 && U_flag[j] ==
1)
        {
            Ui_extract[j][row_count_U[j]] = Ui[i] - (2 +
(int)round(size_perm_mat*(1 - 1 / pow(2, j)))) - 1)*size_sub_mat;
            Ux_extract[j][row_count_U[j]] = Ux[i];
            if (i >= Up[p + (int)round((size_perm_mat*(1 - 1 / pow(2, j +
1))))*size_sub_mat + 1])
            {
                Up_extract[j][col_count_U[j] + 1] = row_count_U[j];
                col_count_U[j] = col_count_U[j] + 1;
                p++;
            }
            row_count_U[j] = row_count_U[j] + 1;
        }
        /* Mark the completion of an extraction */
        if (U_flag[j] == 1 && Ui[i] >= m * size_sub_mat && Ui[i] < (m + 1) *
size_sub_mat)
        {
            Up_extract[j][col_count_U[j] + 1] = row_count_U[j];
            j++;
            m = 2 * m;
            p = 0;
        }
        /* Mark the extraction of the last element is ready */
    }

```



```

        if (j == (int)round(log2(size_perm_mat)) - 2 && Ui[i] >=
((int)size_perm_mat / 4 + 1 - 1) * size_sub_mat && Ui[i] < ((int)size_perm_mat / 4 + 1)
* size_sub_mat)
        {
            U_flag[j] = 1;
        }
        /* Extraction the last element */
        if (j == (int)round(log2(size_perm_mat)) - 2 && U_flag[j] == 1 && Ui[i] >=
(size_perm_mat - 2 - 1) * size_sub_mat && Ui[i] < (size_perm_mat - 2) * size_sub_mat)
        {
            Ui_extract[j][row_count_U[j]] = Ui[i] - (size_perm_mat - 2 - 1) *
size_sub_mat;
            Ux_extract[j][row_count_U[j]] = Ux[i];
            if (i >= Up[p + (int)round((size_perm_mat*(1 - 1 / pow(2, j +
1))))*size_sub_mat + 1])
            {
                Up_extract[j][col_count_U[j] + 1] = row_count_U[j];
                col_count_U[j] = col_count_U[j] + 1;
                p++;
            }
            row_count_U[j] = row_count_U[j] + 1;
        }
        /* Mark the completion of the extraction of the last element */
        if (j == (int)round(log2(size_perm_mat)) - 2 && Ui[i] >= (size_perm_mat -
1) * size_sub_mat && Ui[i] < size_perm_mat * size_sub_mat)
        {
            Up_extract[j][col_count_U[j] + 1] = row_count_U[j];
            j++;
        }
    }
}

```

APPENDIX D – LIST OF PUBLICATIONS

The following is a list of journal and conference papers that are derived from this Ph.D. project.

Journal publications

M. Cai, H. Gras, J. Mahseredjian, E. Rutovic, A. El-Akoum, “Functional Mock-up Interface-Based Approach for Parallel and Multistep Simulation of Electromagnetic Transients,” *IEEE Trans. on Power Delivery*, vol. 33, issue 6, pp. 2978-2988, Dec. 2018.

M. Cai, J. Mahseredjian, U. Karaagac, A. El-Akoum, X. Fu, “Functional Mock-up Interface Based Parallel Multistep Approach with Signal Correction for Electromagnetic Transients Simulations,” *IEEE Trans. on Power Systems*, vol. 34, issue 3, pp. 2482-2484, May 2019.

Conference publications

M. Cai, J. Mahseredjian, H. Gras, A. El-Akoum, X. Fu, “A Co-Simulation Based Parallel and Multistep Approach for Accelerating EMT Simulations,” International Conference on Power Systems Transients (*IPST*), Perpignan, France, June 18-21, 2019.

M. Cai, J. Mahseredjian, A. Haddadi, X. Fu, “A Parallelization-in-time Approach for Accelerating Power System EMT Simulations,” Power Systems Computation Conference (*PSCC*), Porto, Portugal, June 29-July 3, 2020, submitted.

A. Haddadi, **M. Cai**, U. Karaagac, J. Mahseredjian, “**Power System Test Cases for EMT-type Simulation Studies,**” *International Conference on Power Systems Transients (IPST)*, Perpignan, France, June 18-21, 2019.

X. Fu, S. M. Seye, J. Mahseredjian, **M. Cai**, C. Dufour, “**A Comparison of Numerical Integration Methods and Discontinuity Treatment for EMT Simulations,**” *Power System Computation Conference (PSCC)*, Dublin, Ireland, 2019.

M. Cai, J. Mahseredjian, “A Parallel Numerical Method Based on the PIMIT for Fast Power System EMT Simulations,” *The 11th International Symposium on EMC and Transients in Infrastructures, the 13th International Student Session (ISET/ISS)*, Kyoto, Japan, 2017.